

# All-Optical Processing of Novel Modulation Formats Using Semiconductor Optical Amplifiers

Tutorial FTuV1

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# What is this tutorial about?

**NRZ-OOK** has been the data format of choice for decades:

- Nonreturn-to-zero on/off keying (NRZ-OOK, low complexity)

**Bit rates  $\geq 40$  Gbit/s/ch** need novel **OOK** formats or **regeneration**:

- DB (CD tolerance  $\geq 120$  ps/nm)
- VSB-CSRZ (spectral efficiency  $\geq 0.8$  bit/s/Hz)

**Bit rates  $\geq 40$  Gbit/s/ch** with novel **phase** modulation formats:

- **RZ-DPSK** (nonlinear tolerance)
- **RZ-DQPSK** (CD tol.  $\geq 120$  ps/nm, DGD tolerance  $\geq 20$  ps, spectral efficiency  $\geq 0.8$  bit/s/Hz, nonlinear tolerance)

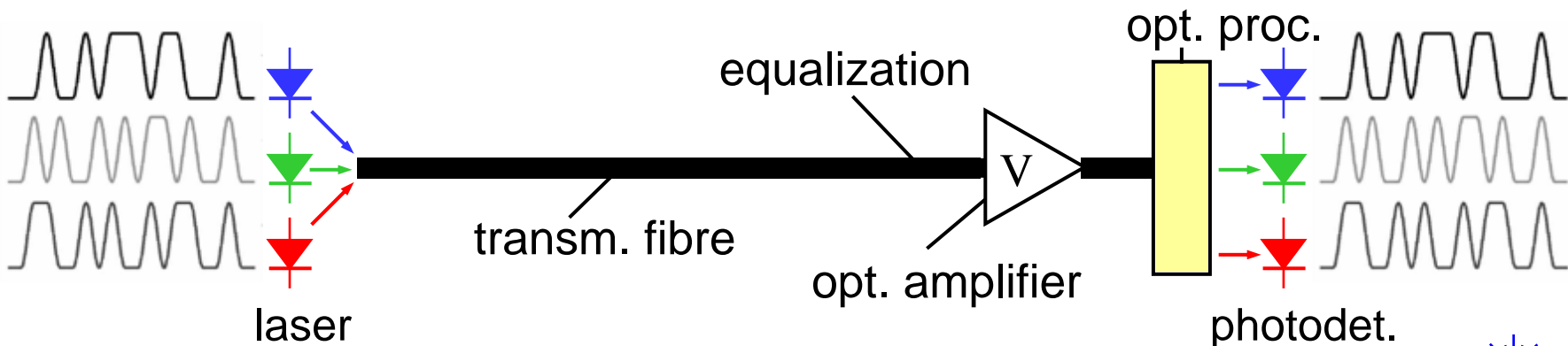
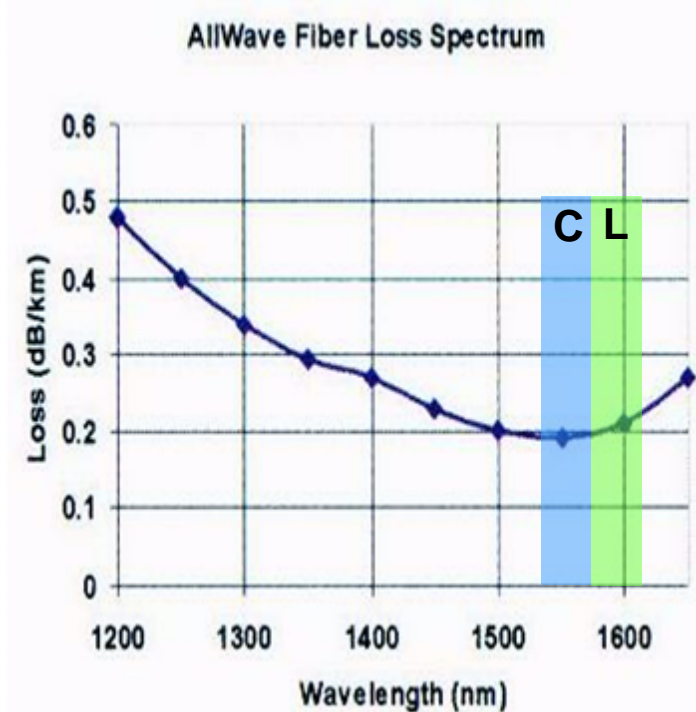
**SOA with fast XPM / XGM** (cross-phase / gain modulation) act as

- all-optical **logic gates**, and as
- all-optical **wavelength converters** and **regenerators**.



# Optical Wavelength Division Multiplexing (WDM)

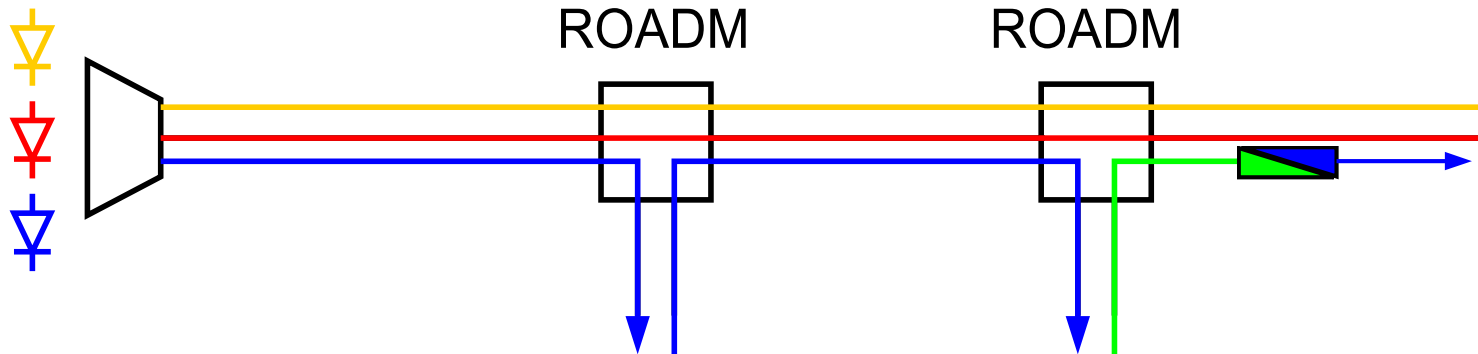
- Internet: Need for bandwidth  $B$
- Optical transmission systems
  - fibres:  $B \approx 65$  THz (450 nm)
  - amplifiers:  $B \approx 10$  THz ( 80 nm)
  - wavelength division multiplexing
  - channels:  $\Delta f \approx 5, 10, 25, 50, 100$  GHz
  - capacity:  $40$  Gbit/s  $\times 100$  ch =  $4$  Tbit/s





# The Need for All-Optical Processing — Wavelength Blocking

Optical transparent wavelength division multiplexing (WDM) networks need reconfigurable optical add-drop multiplexers (ROADM):

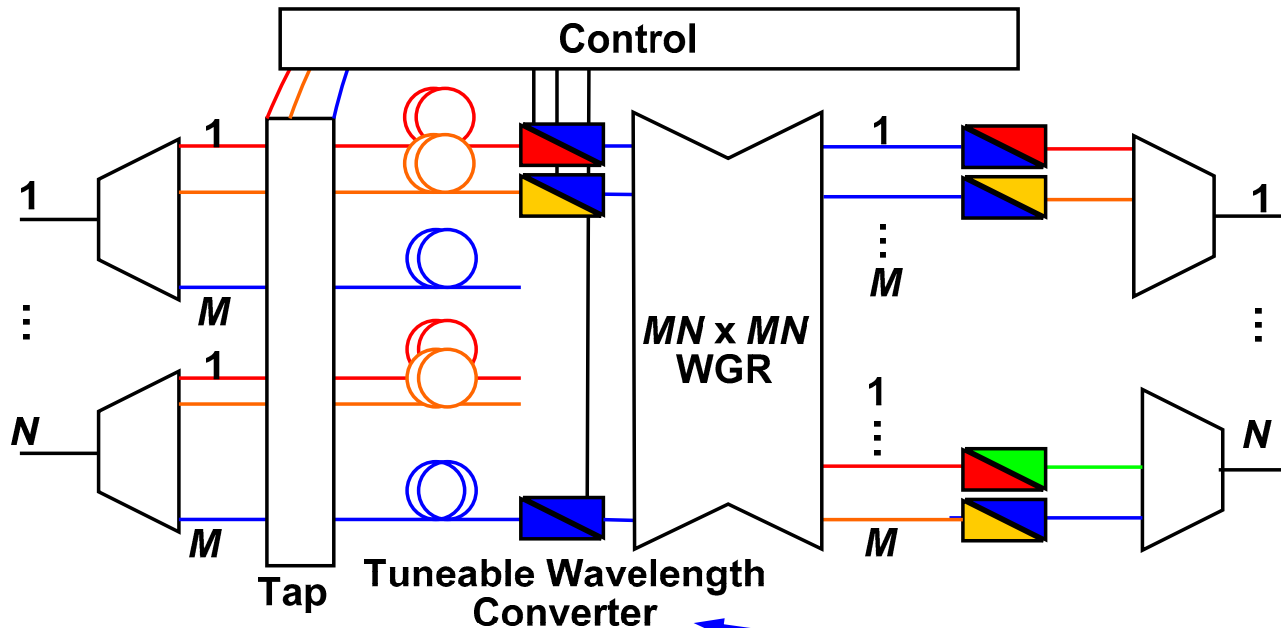


All-optical wavelength converters prevent wavelength blocking.



# The Need for All-Optical Processing — Space Switching

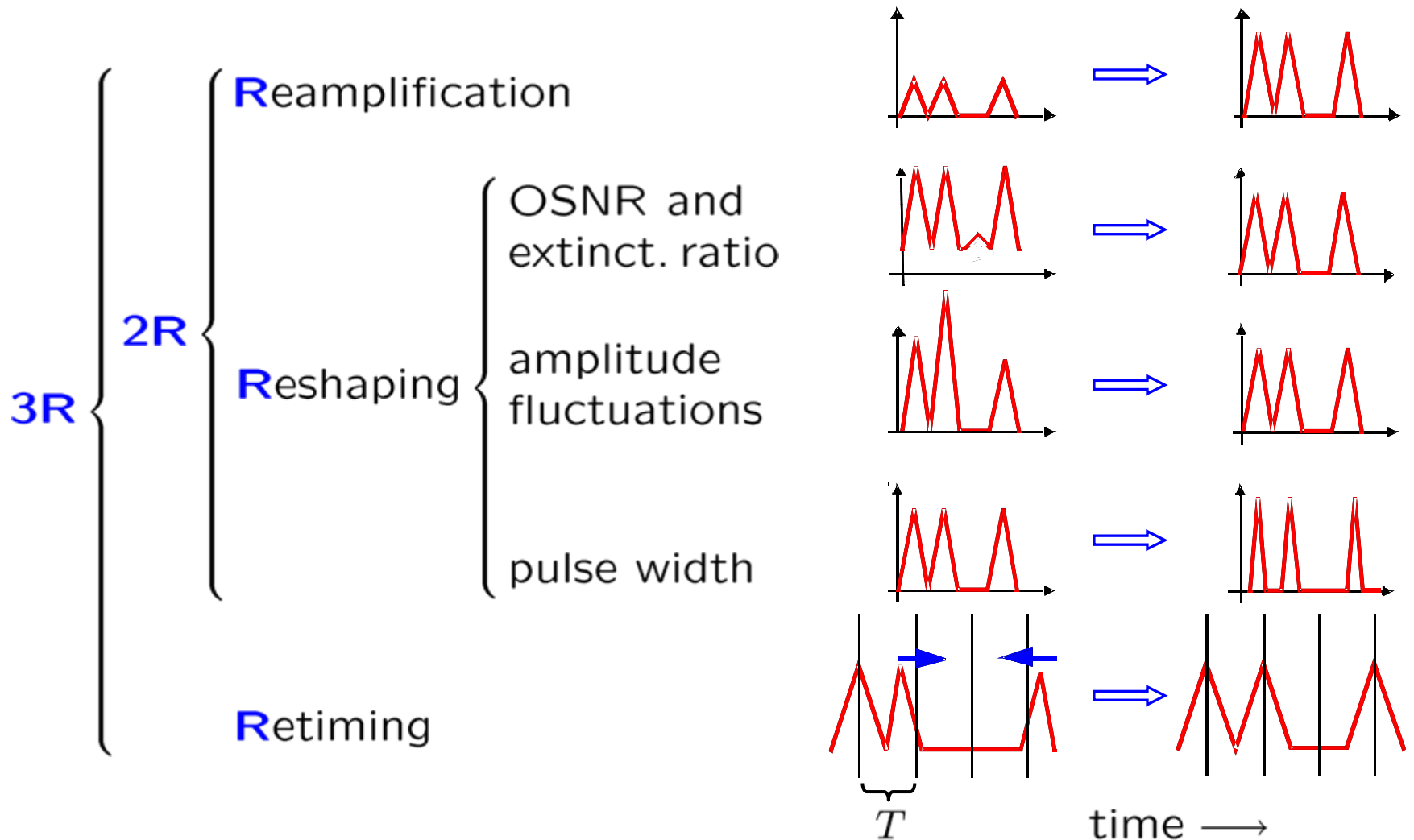
Optical cross-connect (OXC) in a meshed network:



- Sub-nanosecond guard time for reconfiguring a tunable laser
- Space-switching via fast wavelength switching and WGR
- Wavelength mapping



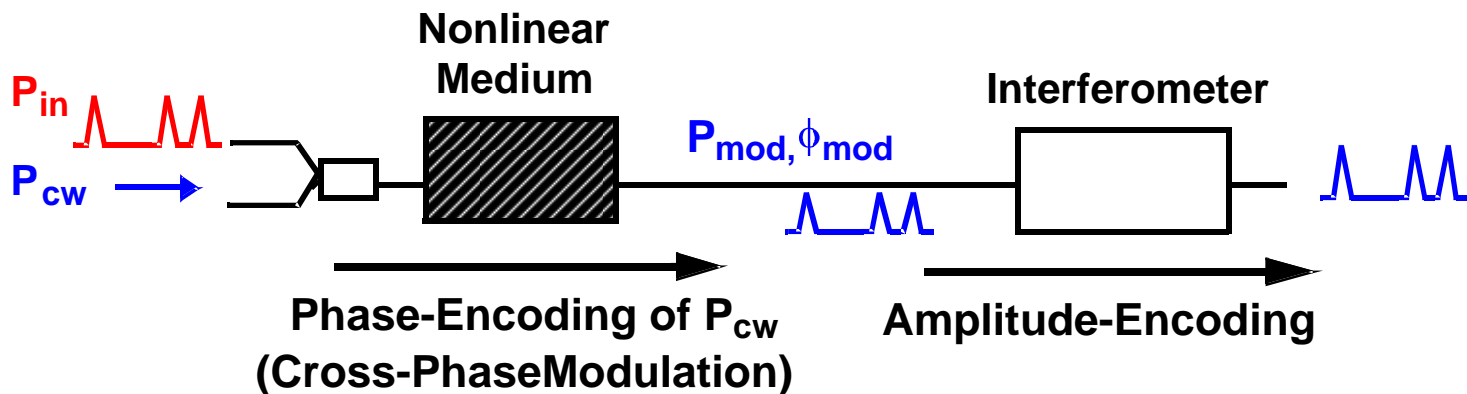
# The Need for All-Optical Processing — Regeneration



Ultimate criterion: Signal quality  $Q$



# Regeneration and Transfer Function



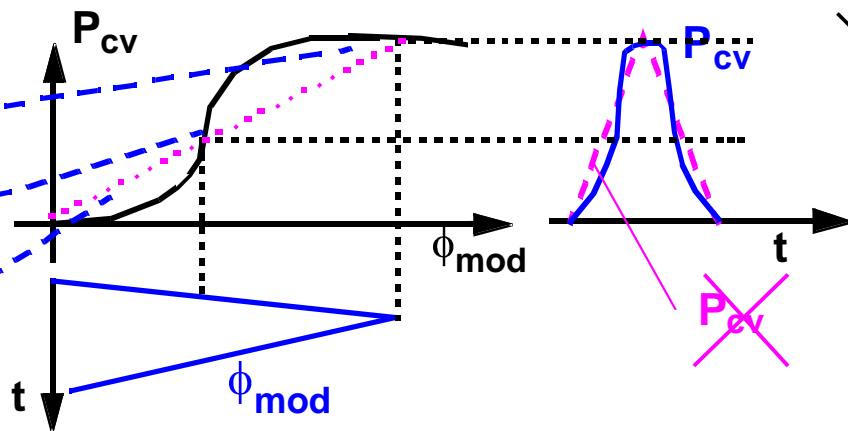
## Interferometer Transfer Fct

## Ultimate Test

Improves Dynamic  
Input Range

Reduces FWHM

Improves OSNR,  
Extinction Ratio,



Signal  
Quality  
&  
BER  
Sensitivity  
Improvement!



# Outline

- **Modulation techniques**
  - Analogue, digital, coding
  - Symbol diagrams, spectra
  - Benefits, transmission capacity
- **SOA gain and phase recovery**
  - Gain-phase coupling
  - Physical explanation
- **SOA signal processing**
  - Logic gate
  - OOK wavelength converter
  - DPSK wavelength converter
- **Summary**



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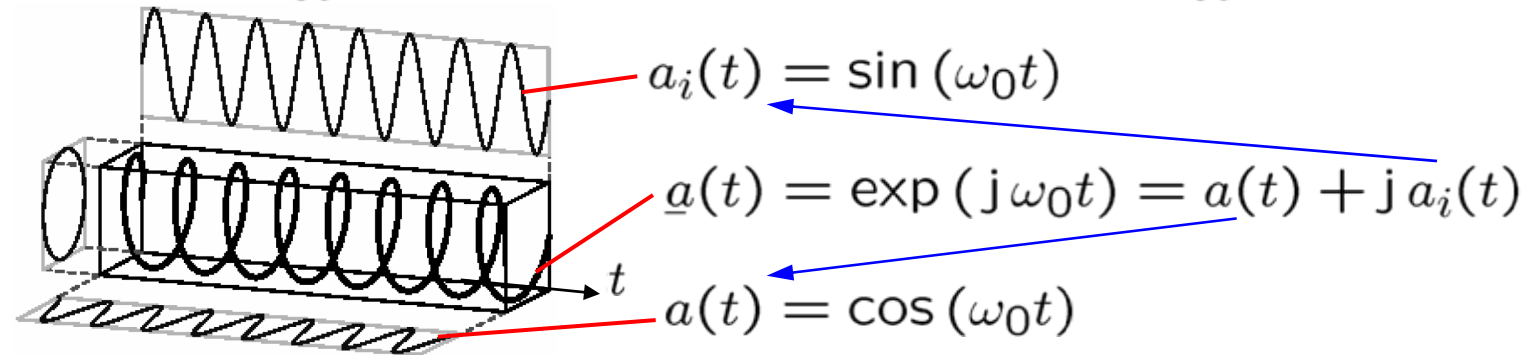
# Analytic Time Signals

Analytic signals are frequency-causal (only positive freq.), example:

$$\underline{a}(t) = A_0 e^{j\omega_0 t} \quad \circ \text{---} \bullet \quad \bar{a}(f) = \int_{-\infty}^{+\infty} \underline{a}(t) e^{-j2\pi ft} dt = A_0 \delta(f - f_0)$$

Analytic signal is generated from its real  $a(t) = \Re\{\underline{a}(t)\}$  or imaginary part  $a_i(t) = \Im\{\underline{a}(t)\}$  by Hilbert trafo, i. e. by convolution (\*):

$$a_i(t) = \frac{1}{\pi} \mathcal{P} \int_{-\infty}^{+\infty} \frac{a(t')}{t - t'} dt' = a(t) * \frac{1}{\pi t}, \quad a(t) = -\frac{1}{\pi} \mathcal{P} \int_{-\infty}^{+\infty} \frac{a_i(t')}{t - t'} dt' = -a_i(t) * \frac{1}{\pi t}$$



Quadrature (Hilbert) filter  $\bar{\mathcal{H}}(f)$  with impulse response  $\mathcal{H}(t) = \frac{1}{\pi t}$ :

$$\bar{\mathcal{H}}(f) = -j \operatorname{sgn}(f), \quad \operatorname{sgn}(f) = \begin{cases} +1, & f > 0 \\ -1, & f < 0 \end{cases}$$



# Analogue Amplitude Modulation

Analytic narrowband modulation signal:

$$\underline{s}(t) = s(t) + j s_i(t)$$

Narrowband-modulated analytic carrier:

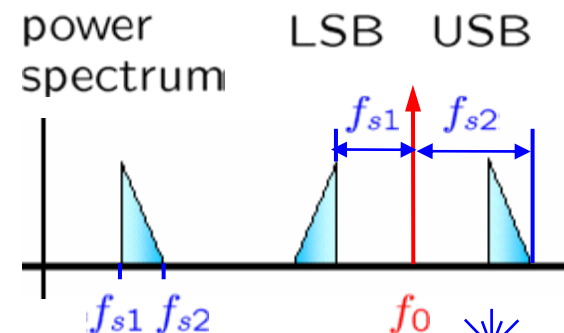
$$\begin{aligned} \underline{a}_s(t) &= a_s(t) + j a_{s_i}(t) = \underline{s}(t) A_0 \exp(j \omega_0 t) \\ &= A_0 [s(t) + j s_i(t)] [\cos(\omega_0 t) + j \sin(\omega_0 t)], \end{aligned}$$

real part:  $a_s(t) = A_0 [s(t) \cos(\omega_0 t) - s_i(t) \sin(\omega_0 t)],$

imag. part:  $a_{s_i}(t) = A_0 [s(t) \sin(\omega_0 t) + s_i(t) \cos(\omega_0 t)]$

Amplitude modulation (AM)  $s(t) = 1 + m \cos(\omega_s t)$ ,  $s_i(t) = 0$ , AM index  $m$ , symmetric sidebands for  $\underline{s}(t) = s(t)$  real:

$$\begin{aligned} a_{AM}(t) &= \{1 + m \cos(\omega_s t)\} A_0 \cos(\omega_0 t) \\ &= A_0 \cos(\omega_0 t) + \frac{m}{2} A_0 \cos[(\omega_0 - \omega_s) t] \\ &\quad + \frac{m}{2} A_0 \cos[(\omega_0 + \omega_s) t] \end{aligned}$$



Incoherent Rx with  $a_{AM}^2$ -operation



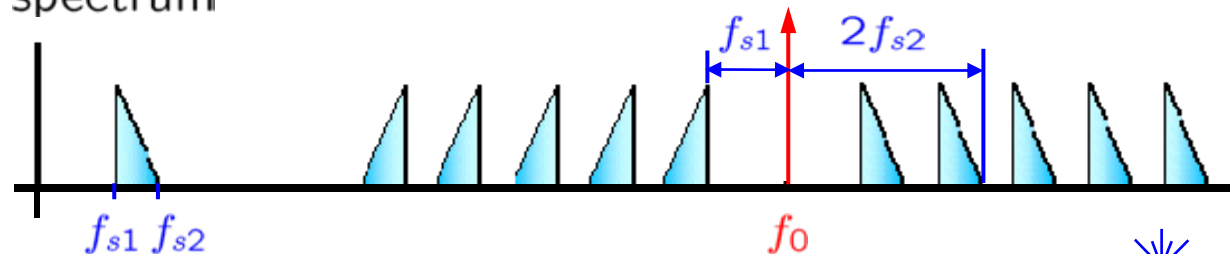


# Analogue Intensity Modulation

Intensity modulation (IM)  $s^2(t) = 1 + p \cos(\omega_s t)$ ,  $s_i(t) = 0$ , IM index  $p$  (here:  $p \ll 1$ ):

$$\begin{aligned}
 a_{\text{IM}}(t) &= \sqrt{1 + p \cos(\omega_s t)} A \cos(\omega_0 t) \\
 &\approx \left\{ 1 + \frac{p}{2} \cos(\omega_s t) - \frac{p^2}{8} \cos^2(\omega_s t) + \dots \right\} A \cos(\omega_0 t) \\
 &\approx \left\{ 1 - \frac{p^2}{16} + \dots \right\} A \cos(\omega_0 t) \\
 &\quad + \left\{ \frac{p}{4} + \frac{3p^3}{128} + \dots \right\} A \left\{ \cos[(\omega_0 - \omega_s)t] + \cos[(\omega_0 + \omega_s)t] \right\} \\
 &\quad + \left\{ -\frac{p^2}{32} + \dots \right\} A \left\{ \cos[(\omega_0 - 2\omega_s)t] + \cos[(\omega_0 + 2\omega_s)t] \right\} \\
 &\quad + \left\{ \frac{p^3}{128} + \dots \right\} A \left\{ \cos[(\omega_0 - 3\omega_s)t] + \cos[(\omega_0 + 3\omega_s)t] \right\}
 \end{aligned}$$

power  
spectrum



Incoherent Rx  
with  $a_{\text{IM}}^2$ -operation



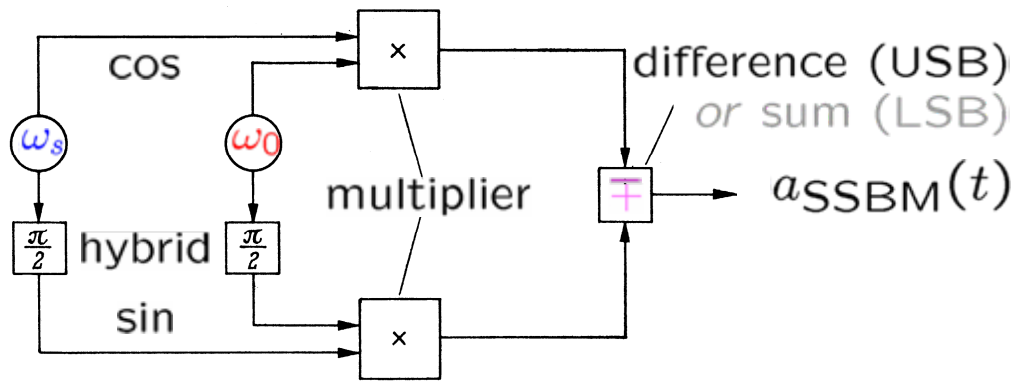
# Analogue Single and Vestigial Sideband Modulation

Single-sideband (SSB) modulation  $\underline{s}(t) = m \exp(j\omega_s t)$ , USB/LSB:

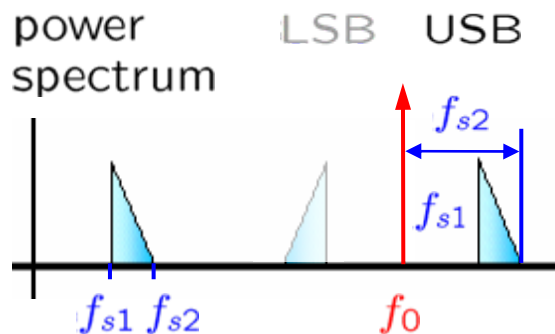
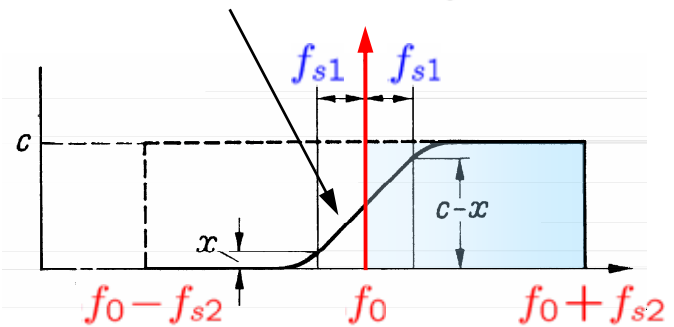
$$a_{SSBM}(t) = m \exp(\pm j\omega_s t) A_0 \exp(j\omega_0 t) = mA_0 \exp[j(\omega_0 \pm \omega_s)t]$$

$$a_{SSBM}(t) = mA_0 [\cos(\omega_s t) \times \cos(\omega_0 t) \mp \sin(\omega_s t) \times \sin(\omega_0 t)]$$

$$a_{SSBM,i}(t) = mA_0 [\cos(\omega_s t) \times \sin(\omega_0 t) \pm \sin(\omega_s t) \times \cos(\omega_0 t)]$$



Vestigial sideband mod.  
with finite-slope filter:



Coherent Rx

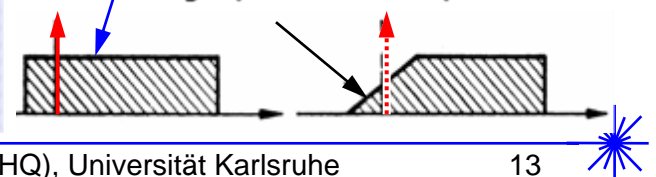
DSB:  $\varphi$ -synchr. LO

SSB: asynchr. LO

VSB: vestigial  $f_0$

At Tx partial filtering,

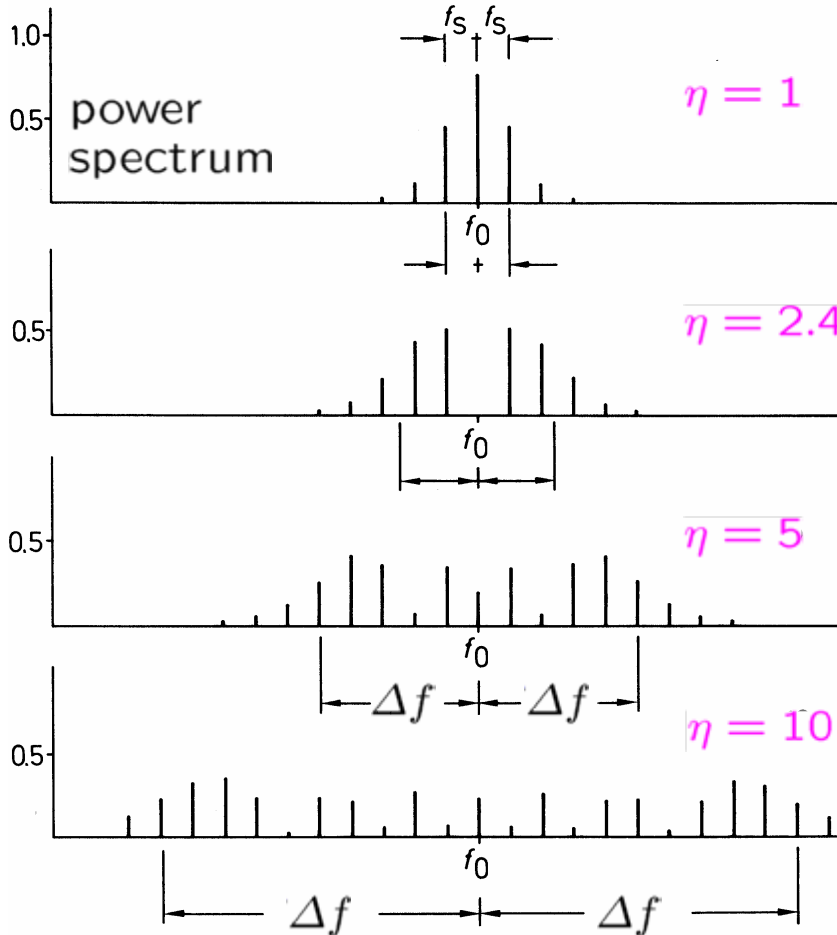
Nyquist slope at Rx:



# Analogue Angle Modulation

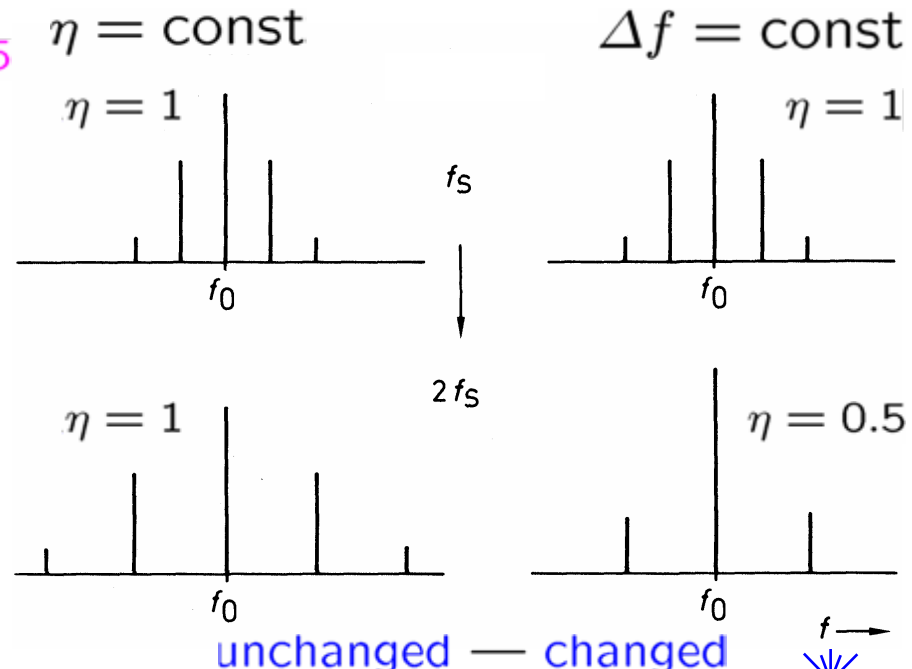
Phase/frequency modulation (PM/FM)  $s(t) = \eta \sin(\omega_s t)$ , index  $\eta$ :

$$\underline{a}_\eta(t) = A_0 \exp \left[ j \left[ \omega_0 t + \eta \sin(\omega_s t) \right] \right] = A_0 \sum_{n=-\infty}^{+\infty} J_n(\eta) \exp \left[ j (\omega_0 + n \omega_s) t \right]$$



Frequency deviation:  $\Delta f = \eta f_s$

PM — spectra — FM



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# Time Functions for a Binary-Modulated Carrier

Unipolar ASK: No phase change between mark and space

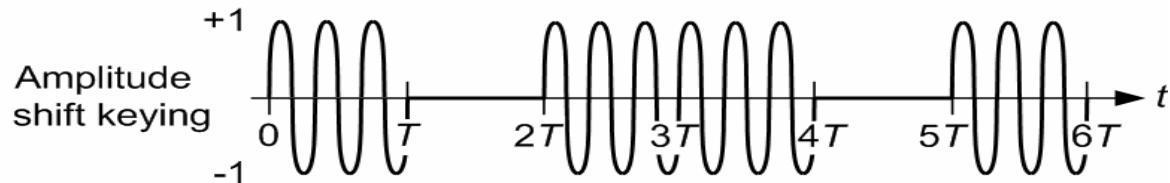
OOK: Unipolar ASK, high/zero power for mark/space

Bipolar ASK:  $\pi$ -phase change between alternating bit slots

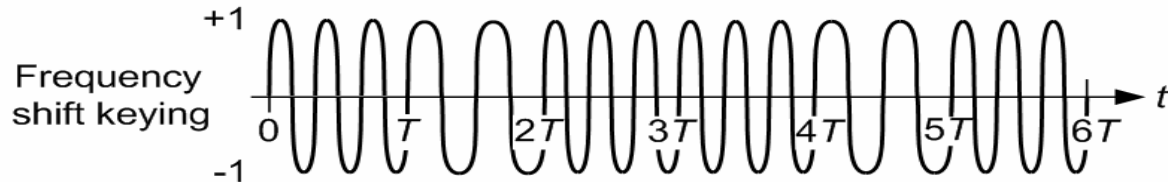
$\pi/2$ -ASK:  $\pi/2$ -phase change between alternating bit slots

Information                      1            0            1            1            0            1

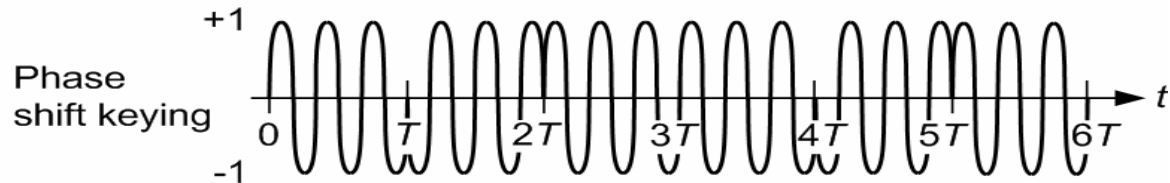
ASK OOK:



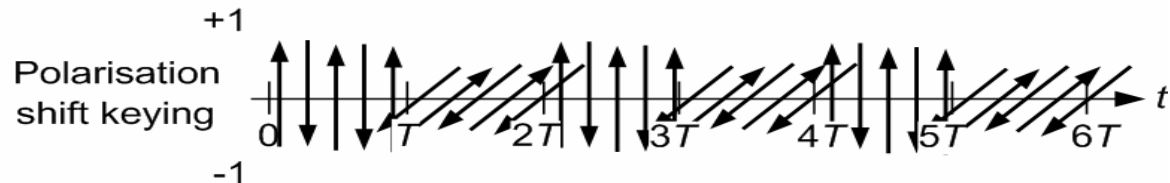
FSK:



PSK:



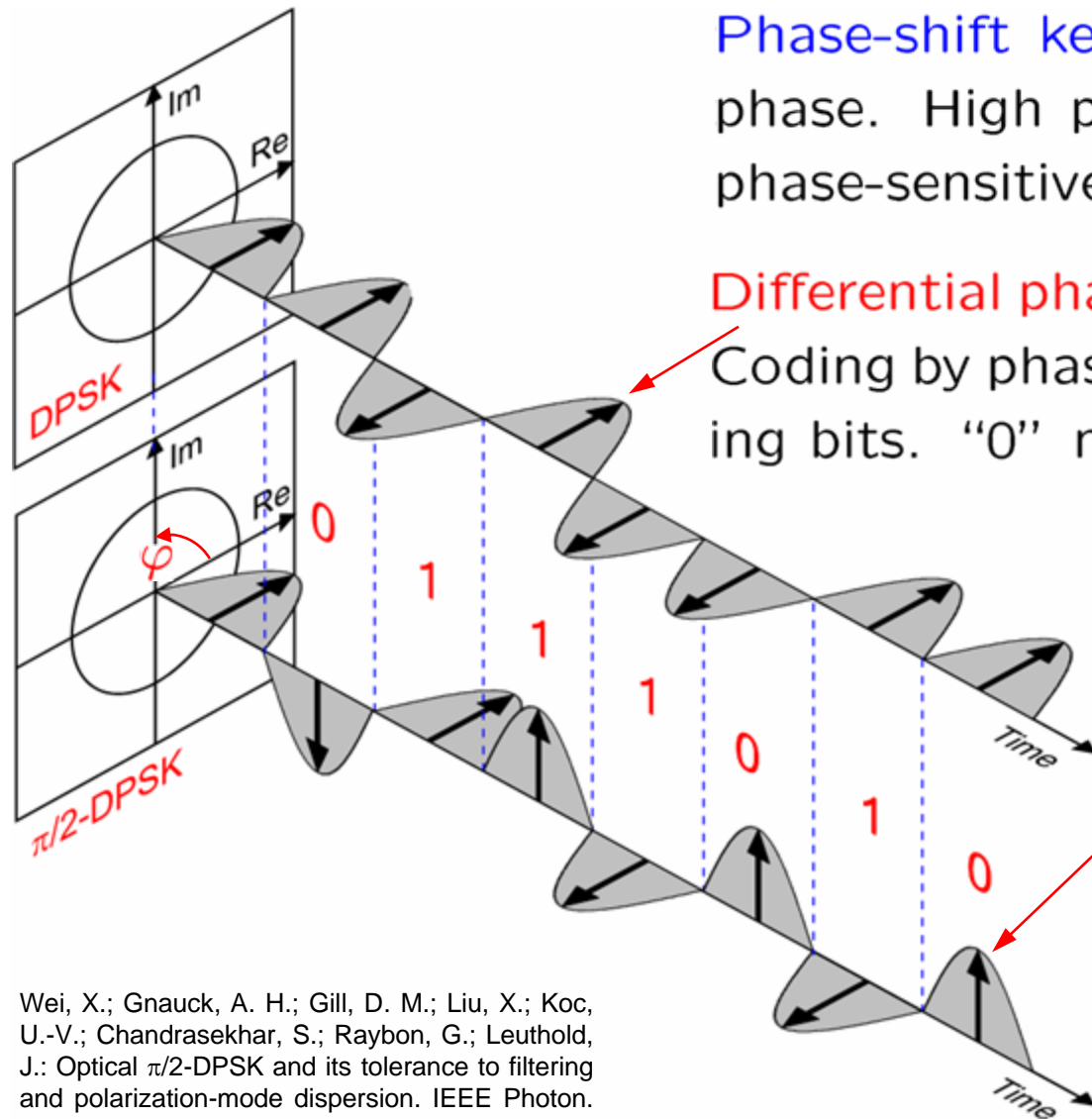
PoISK:



# (Binary) Phase-Shift Keying

**Phase-shift keying (BPSK):** Coding by phase. High phase purity Tx, coherent phase-sensitive Rx, highly stable LO.

**Differential phase-shift keying (DBPSK):** Coding by phase *difference* of neighbouring bits. "0" no, "1"  $\pi$ -change.



**$\pi/2$ -DBPSK:**

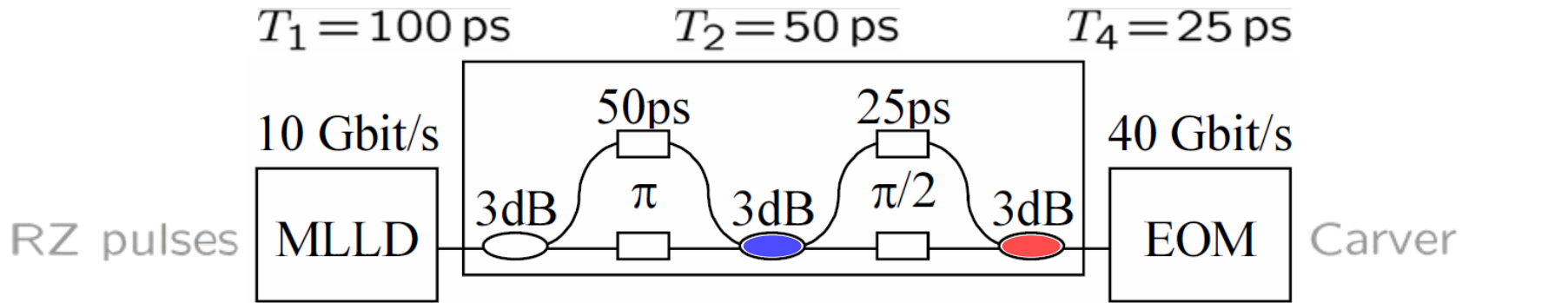
"0"  $\Delta\varphi = -\pi/2$  (cw),  
 "1"  $\Delta\varphi = +\pi/2$  (ccw).

DPSK with progressive  $\pi/2$ -phase per bit.

Wei, X.; Gnauck, A. H.; Gill, D. M.; Liu, X.; Koc, U.-V.; Chandrasekhar, S.; Raybon, G.; Leuthold, J.: Optical  $\pi/2$ -DPSK and its tolerance to filtering and polarization-mode dispersion. IEEE Photon. Technol. Lett. 15 (2003) 1639–1641



# Generation of Alternating-Phase Optical Pulses



(i) Pulses interleaved by  $T_1/2 = 50 \text{ ps}$  differ by carrier phase  $\pi$

(ii) Pulses interleaved by  $T_2/2 = 25 \text{ ps}$  differ by carrier phase  $\frac{\pi}{2}$

(i)  $\pi$ -alternating phase

(ii) Progressive  $\pi/2$  phase increase

Carrier-suppressed (CS) RZ

Nbo impulses differ by phase  $\pi$

- $\pi/2$ -AP-ASK-OOK: Reduced intersymbol interference (ISI) wrt to middle impulse, robust wrt intrachannel four-wave mixing
- $\pi/2$ -DPSK: High tolerance to polarisation mode dispersion

Schnarrenberger, M., Sotobashi, H., Chujo, W. and Freude, W.: Novel intersymbol interference reduction technique by bit synchronized  $\pi/2$  phase shift. Proc. Institute of Electronics, Information and Communication Engineers (IEICE Japan) Spring Conference, Hiroshima, 28.–31.03.2000

Douglas M. Gill, D. G.; Gnauck, A. H.; Liu, X.; Wei, X.; Su, Y.:  $\pi/2$  alternate-phase on-off keyed 42.7 Gb/s long-haul transmission over 1980 km of standard single-mode fiber. IEEE Photonics Technol. Lett. 16 (2004) 906–908

Wei, X; Leuthold, J.; Dorrer, C.; Gill, D. M.; Liu, X.: Chirp reduction of  $\pi/2$  alternate-phase pulses by optical filtering. Technical Digest Optical Fiber Communication Conference (Ofc'05), Anaheim (CA), USA, 06.–11.03.2005. Paper JWA42





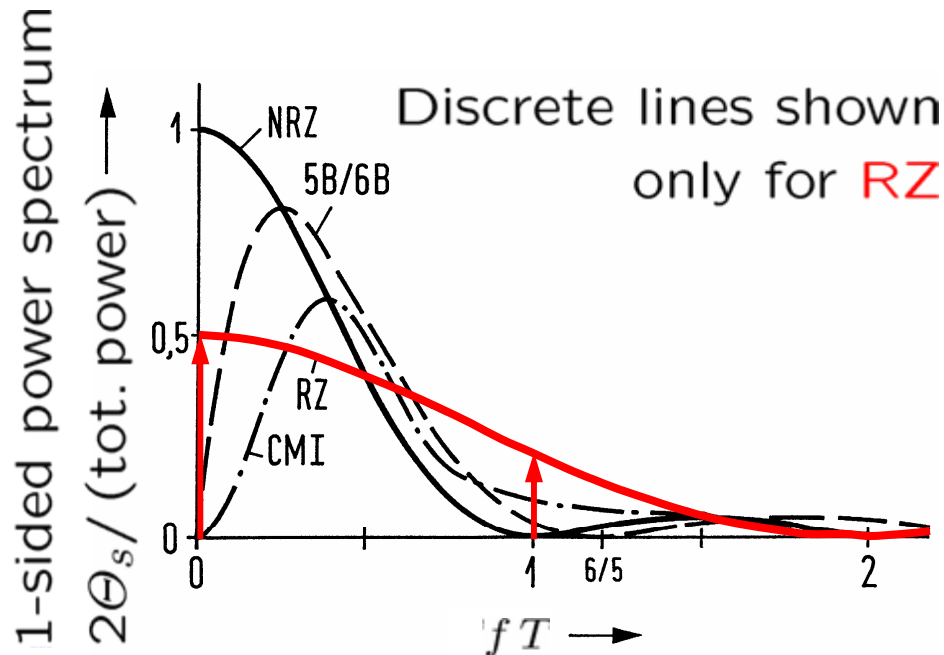
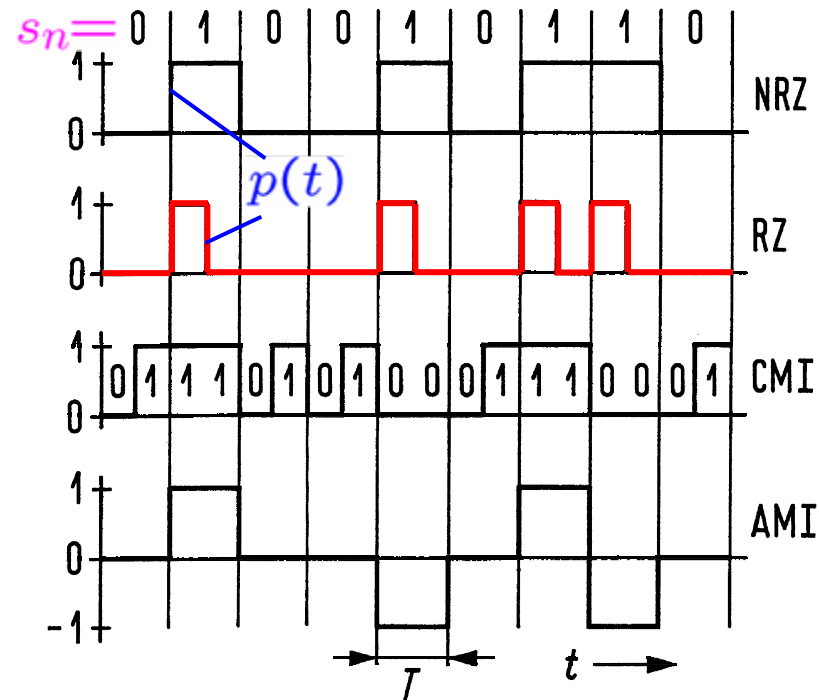
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# Coding Schemes and Calculation of One-Sided PRBS Spectra



NRZ/RZ: 
$$s(t) = \sum_{n=-\infty}^{+\infty} s_n p(t - nT) = \sum_{n=-\infty}^{+\infty} s_n p(t) * \delta(t - nT)$$

2-sided power spectr.: 
$$\Theta_s(f) = \frac{1}{4T} |\bar{p}(f)|^2 \left[ 1 + \frac{1}{T} \sum_{n=-\infty}^{+\infty} \delta(f - n/T) \right]$$

Hözlner, E.; Holzwarth, H.: Pulstechnik, Vol. I, 2. Ed. ("Pulse technology", in German). Berlin: Springer 1982 ([General PRBS spectra](#), Eq. (6.48,49))

Cattermole, K. W.; O'Reilly, J. J. (Eds.): Mathematical topics in telecommunications. Vol. 2: ...randomness... London: Pentech 1984 (Chapter 15)

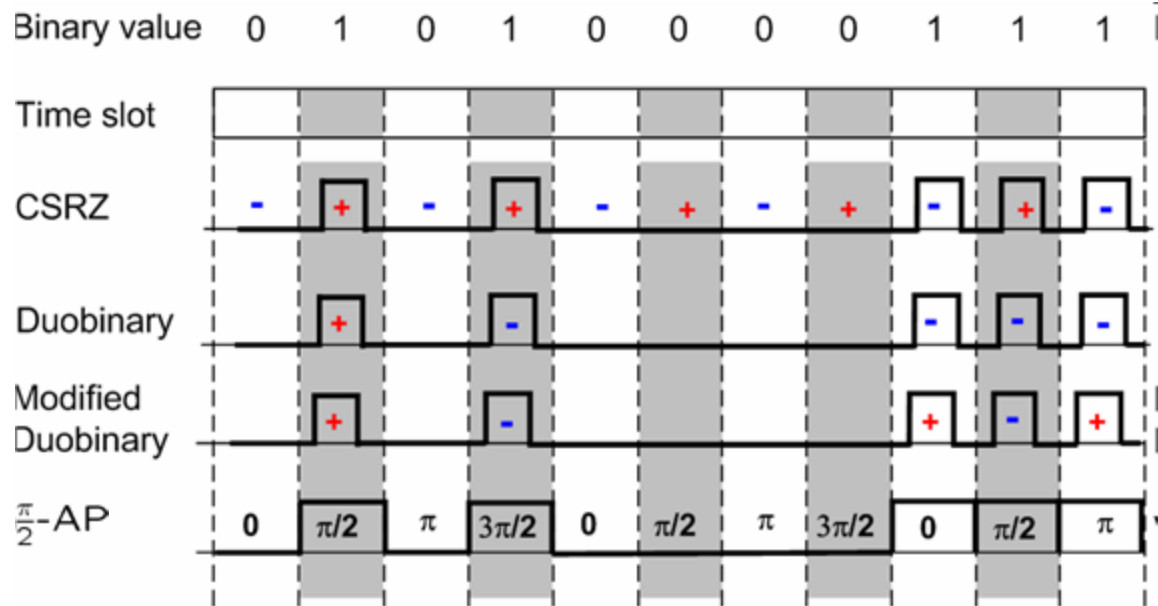
Grau, G.; Freude, W.: Optische Nachrichtentechnik, 3. Ed. Berlin: Springer 1991 (Sect. 7.2)

Agrawal, G. P.: Lightwave technology. Telecommunication systems. Hoboken (NJ): John Wiley & Sons 2005 (Sect. 2.2)

Ip, E.; Kahn, J. M.: Power spectra of return-to-zero optical signals. J. Lightw. Technol. 24 (2006) 1610-1618



# Bipolar Binary ASK Coding Schemes



**CSRZ:** Carrier-suppressed RZ,  $\pi$ -phase change from bit to bit

**DB:** Duobinary, high/zero power for mark/space,  
0/ $\pi$ -phase for even/odd number of "0" since last "1"

**AMI:** Modified duobinary,  $\pi$ -phase from "1" to subsequent "1"

**$\frac{\pi}{2}$ -AP:** Progressive  $\pi/2$ -alternating phase change for improving ISI



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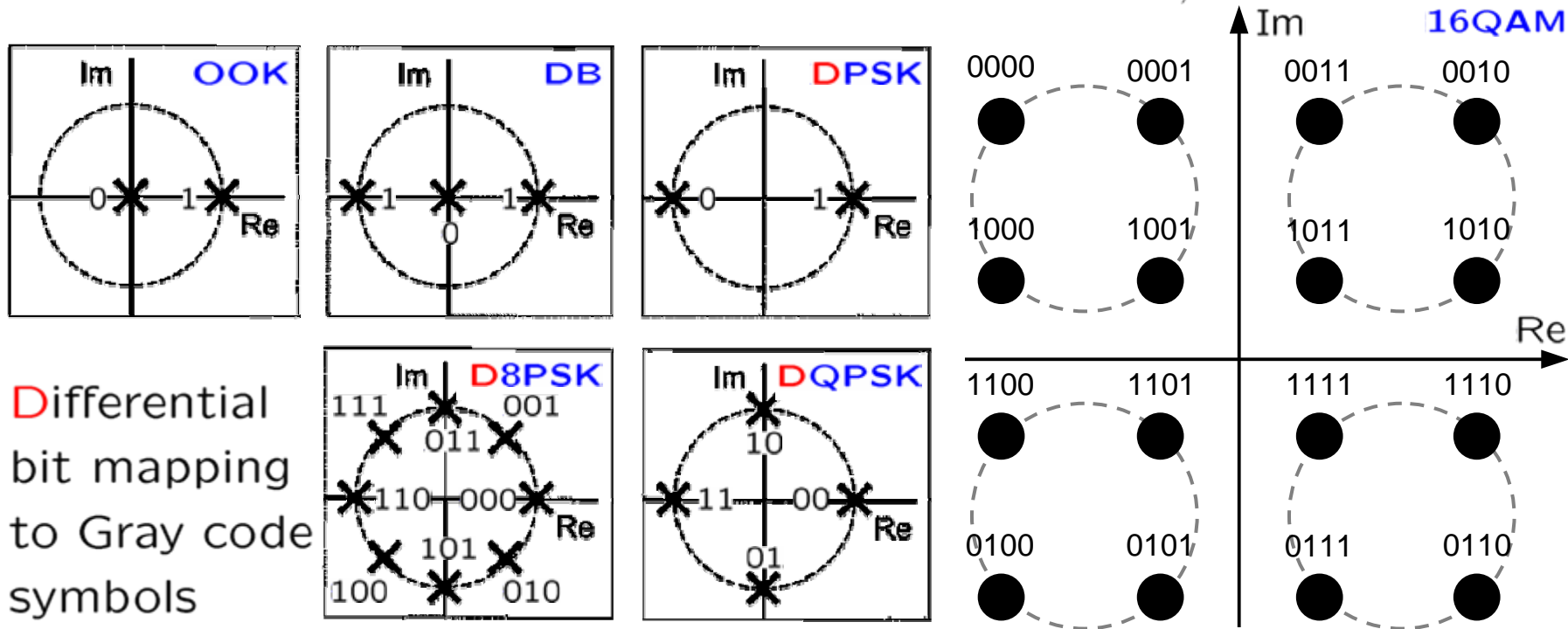


# Symbol Diagrams for Digital Modulation Formats

Analytic digital modulation signal:

$$\underline{s}(t) = \sum_{n=-\infty}^{+\infty} \underline{s}_n p(t - nT) = \sum_{n=-\infty}^{+\infty} (s_n + j s_{n,i}) p(t - nT)$$

Symbol diagram for  $\text{Re}\{\underline{s}_n\} = s_n$  and  $\text{Im}\{\underline{s}_n\} = s_{n,i}$ :



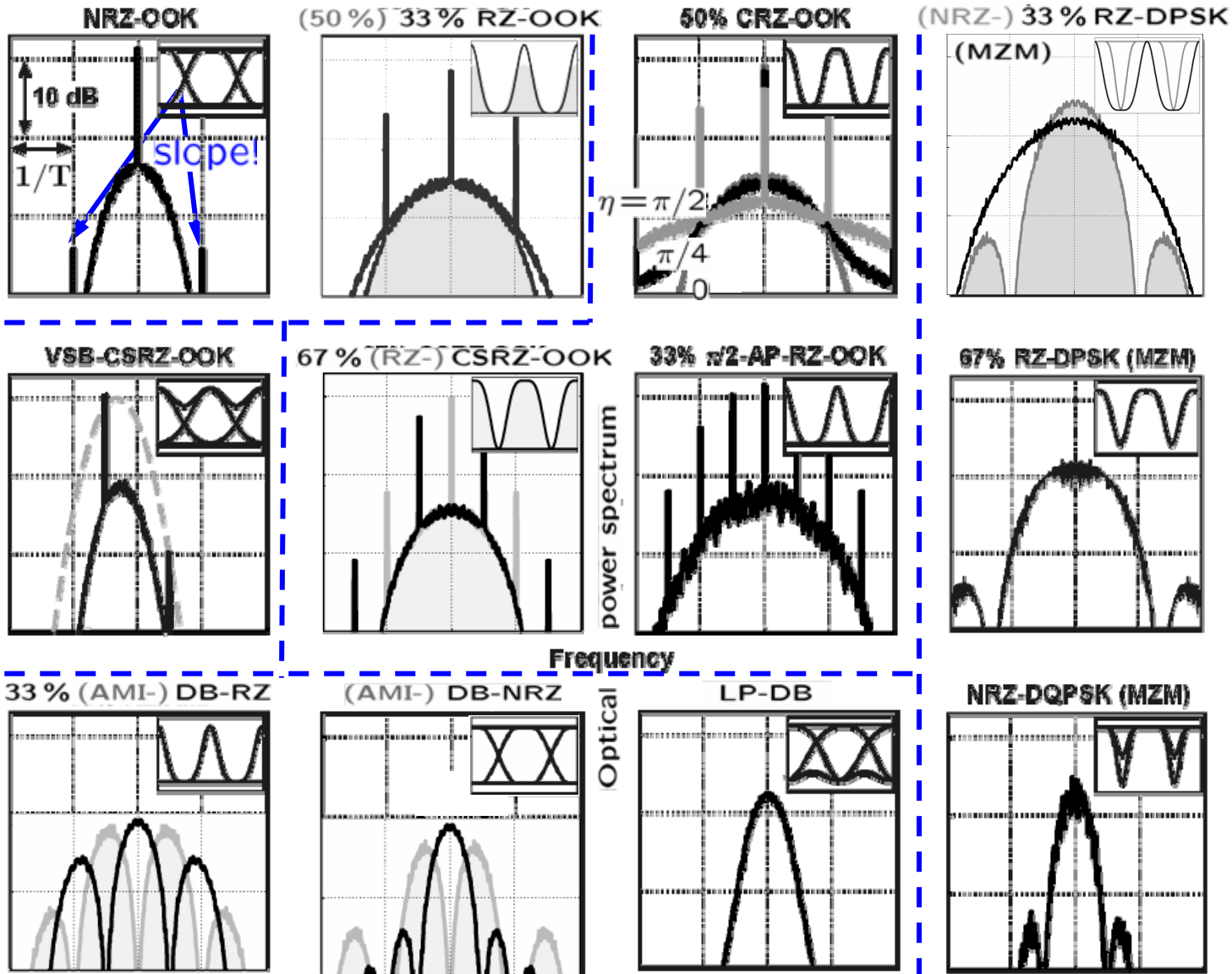
Kikuchi, N.; Sekine, K.; Sasaki, S.: Proposal of inter-symbol interference (ISI) suppression technique for optical multilevel signal generation. Ecoc'06 Tu4.2.1



# Spectra for Amplitude- and Phase-Coded Modulation Formats

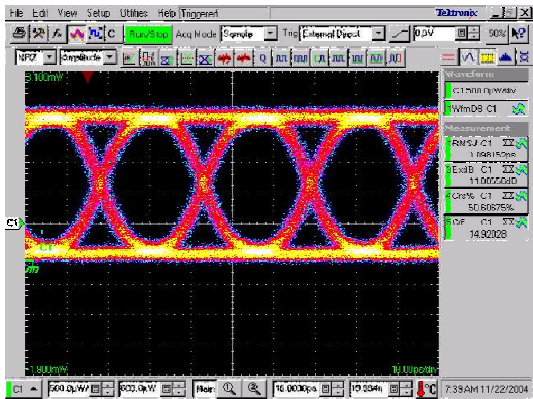
Compiled and modified materials from:

Gnauck, A. H.: Advanced amplitude- and phase coded formats for 40-Gb/s fiber transmission. Proc. 17th Annual Meeting of the IEEE Lasers and Electro-Optics Society (LEOS 2004), Puerto Rico, USA, November 7-11, 2004, Paper WR1  
 Gnauck, A. H.; Liu, X.; Wei, X.; Gill, D. M.; Burrows, E. C.: Comparison of modulation formats for 42.7-Gb/s single-channel transmission through 1980 km of SSMF. IEEE Photon. Technol. Lett. 16 (2004) 909-911  
 Winzer, P. J.: Optical transmitters, receivers, and noise. Wiley Encyclopedia of Telecommunications (2002) <http://www.mrw.interscience.wiley.com/eot/articles/eot404>

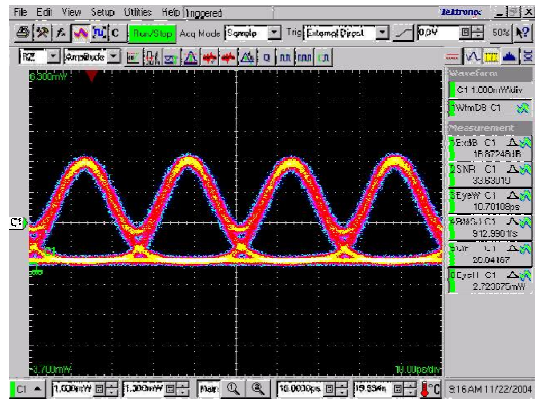


# Eye Diagrams and Spectra for NRZ, CSRZ, 33 % RZ at 40 Gbit/s

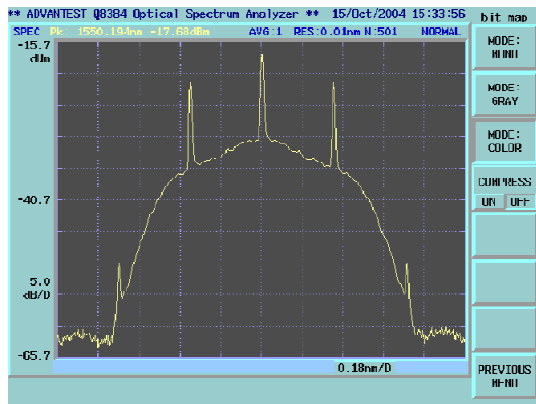
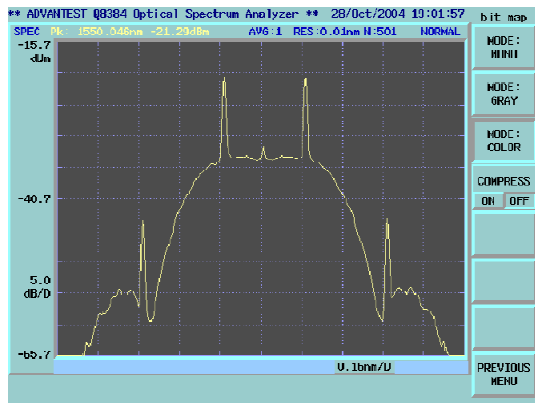
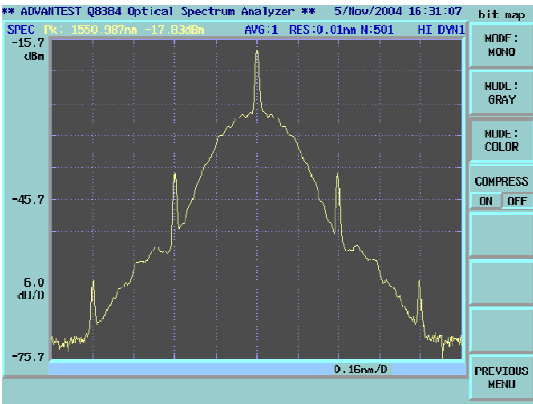
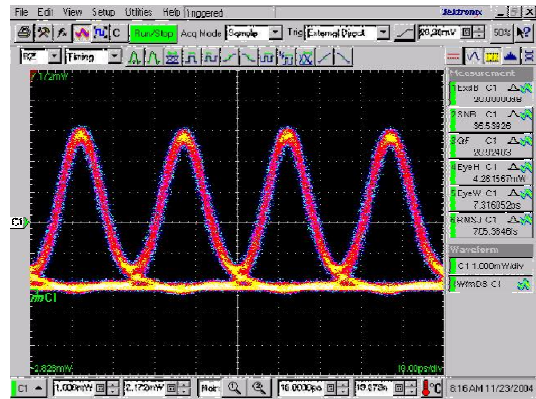
**NRZ**  
Non Return to Zero



**CS-RZ**  
Carrier-Suppressed Return to Zero



**33% RZ**  
Return to Zero



Pincemin, E. et al.: Robustness of the OOK modulation formats at 40 Gbit/s in the practical system infrastructure. Ecoc'05 We4.P.112  
 Gosselin, S.; Joindot, M.: Key drivers and technologies for future optical networks. Ecoc'06 We2.2.1 (Tutorial, Slide 43)



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# 40 Gbit/s Digital Modulation Formats

**40 GHz baud (symbol) rate (1 bit / baud):**

**On-off keying (OOK)** Non-return-to-zero (NRZ), chirped NRZ, RZ (duty cycle 33; 50%), chirped RZ (CRZ,  $x\%$ ), carrier-suppressed RZ (CSRZ, 66%), chirped CSRZ

**Duobinary (DB)** NRZ-DB, chirped NRZ-DB, RZ-DB (33; 50%), CRZ-DB, CSRZ-DB, chirped CSRZ-DB

**Vestigial sideband (VSB)** SB/carrier (partially) filtered

**Different. (binary) phase shift keying (D(B)PSK)** NRZ-DPSK, chirped NRZ-DPSK, RZ-DPSK (33; 50%), CRZ-DPSK, CSRZ-DPSK, chirped CSRZ-DPSK

**20 GHz baud rate (2 bit / baud):**

**Diff. quaternary PSK (DQPSK)** (N)RZ, CSRZ (33; 50; 66%)

**5 GHz baud rate (3 bit / baud):**

**Differential octonary PSK (D8PSK)**

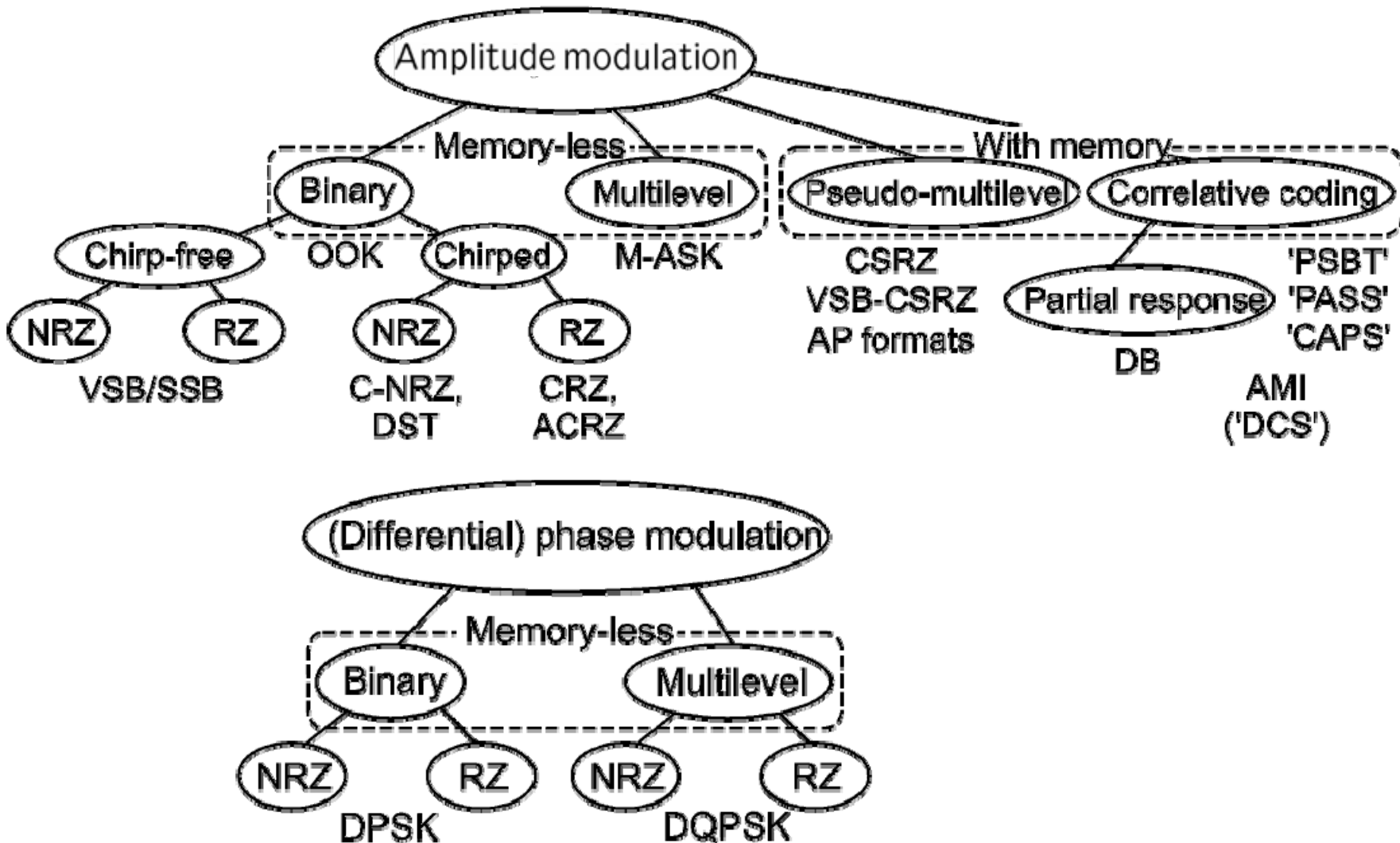
**2.5 GHz baud rate (4 bit / baud):**

**Seno-denary quadrature amplitude modulation (16QAM)**





# A Family of Amplitude and Phase Modulation Formats



# Performance Values of Various Modulation Formats at 40 Gbit/s

Required OSNR at BER =  $10^{-3}$  (7 %-FEC  $\Rightarrow < 10^{-12}$ , 42.7 Gbit/s)

Modulation format	TX complexity	RX complexity	OSNR <sub>req</sub>			CD [ps/nm] (2-dB pen.)	DGD [ps] (1-dB pen.)
			Back-to-back	10 OADMs (0.4 b/s/Hz)	5 OADMs (0.8 b/s/Hz)		
NRZ-OOK	1 MZM	1 PD	15.9 dB	18.2 dB	n/a	54	8
50% RZ-OOK	1-2 MZMs	1 PD	14.4 dB	15.8 dB	n/a	48	10
67% CSRZ-OOK	2 MZMs	1 PD	14.9 dB	14.2 dB	n/a	42	11
DB	1 MZM	1 PD	16.6 dB	14.2 dB	18.4 dB	211 (152)	6
33%RZ-AMI	1-2 MZMs, 1 DI	1 PD	13.4 dB	14.8 dB	n/a	49	10
VSBS-NRZ-OOK	1 MZM + 1 OF	1 PD	16.4 dB	15.6 dB	17.3 dB	63 (155)	6
VSBS-CSRZ	2 MZMs + 1 OF	1 PD	14.8 dB	14.7 dB	16.7 dB	51 (154)	11
NRZ-DPSK	1 MZM	1 DI + 2 PDs	11.7 dB	12.1 dB	17.6 dB	74 (161)	10
50% RZ-DPSK	1-2 MZMs	1 DI + 2 PDs	11.1 dB	11.5 dB	17.0 dB	50 (161)	10
NRZ-DQPSK	2 nested MZMs	2 DIs + 4 PDs	13.2 dB	12.6 dB	12.9 dB	168 (176)	20
50% RZ-DQPSK	2 nested MZMs + 1 PC	2 DIs + 4 PDs	12.2 dB	12.0 dB	12.0 dB	161 (186)	21

85 GHz 43 GHz  
filter BW

PD: photodiode; OF: optical filter; PC: pulse carver; 100 GHz ITU ch spacing;  
12.5 GHz noise reference bandwidth



# Best Transmission Capacities in the Laboratory

## Recent data:

- 40 Gbit/s over 10<sup>6</sup> km (33 % RZ, Lucent El. Lett. 02)
- 160 × 42.7 Gbit/s over 3 200 km (Lucent Ofc03 PD)
- 373 × 10 Gbit/s over 11 000 km (Tyco OFC03 PD)
- 185 × 42.7 Gbit/s over 8 700 km (KDD Ecoc03 PD)
- 89 × 42.7 Gbit/s over 4 000 km (CSRZ-DPSK, Lucent Ecoc03 Tu4.6)
- 64 × 10 Gbit/s over 13 100 km (RZ, Tyco Ecoc03 Tu4.6)
- 149 × 42.7 Gbit/s over 6 120 km (DPSK, Alcatel Ofc04 PDP36)
- 151 × 43 Gbit/s over 4 080 km (DQPSK, Alcatel Ecoc05 PD)
- 640 Gbit/s over 480 km (DQPSK, HHI Ecoc05 We3.2.2)

Maximum capacity transmitted over one fiber today: ~ 7 Tbit/s

Leuthold, J.; Raybon, G.; Su, Y.; Essiambre, R.; Cabot, S.; Jaques J.; Kauer, M.: 40 Gbit/s transmission and cascaded all-optical wavelength conversion over 1 000 000 km. Electron. Lett. 38 (2002) 890–892

Gosselin, S.; Joindot, M.: Key drivers and technologies for future optical networks. Ecoc'06 We2.2.1 (Tutorial, Slide 12)



# Outline

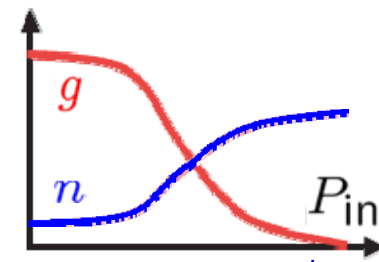
- Modulation techniques
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# Semiconductor Optical Amplifier (SOA) for Signal Processing

- \* SOA carrier lifetimes  $\tau \sim 0.1 \dots 1 \text{ ns}$  (EDFA:  $\tau \sim 10 \text{ ms}$ )
- \* Transient gain variation. For Gbit/s data rates:
  - ★ makes WDM amplifier application difficult, but is
  - ★ good for nonlinear operations, i. e., signal processing.
- \* Fast intraband processes may be exploited, namely:
  - saturation of power gain constant  $g(f) = -2k_0 n_i(f)$ , and its
  - associated change of refractive index  $n(f)$ :
    - ◇ analytic refractive index  $\underline{n}(f) = n(f) - j n_i(f)$ ,
    - ◇ power gain  $G = \exp(gz)$ , field  $\propto \sqrt{G} \exp[j(\omega t - k_0 n z)]$
- \*  $g(f, N)$  and  $n(f, N)$  depend on carrier concentration  $N$ , and are coupled via the Kramers-Kronig (Hilbert transform) relation:

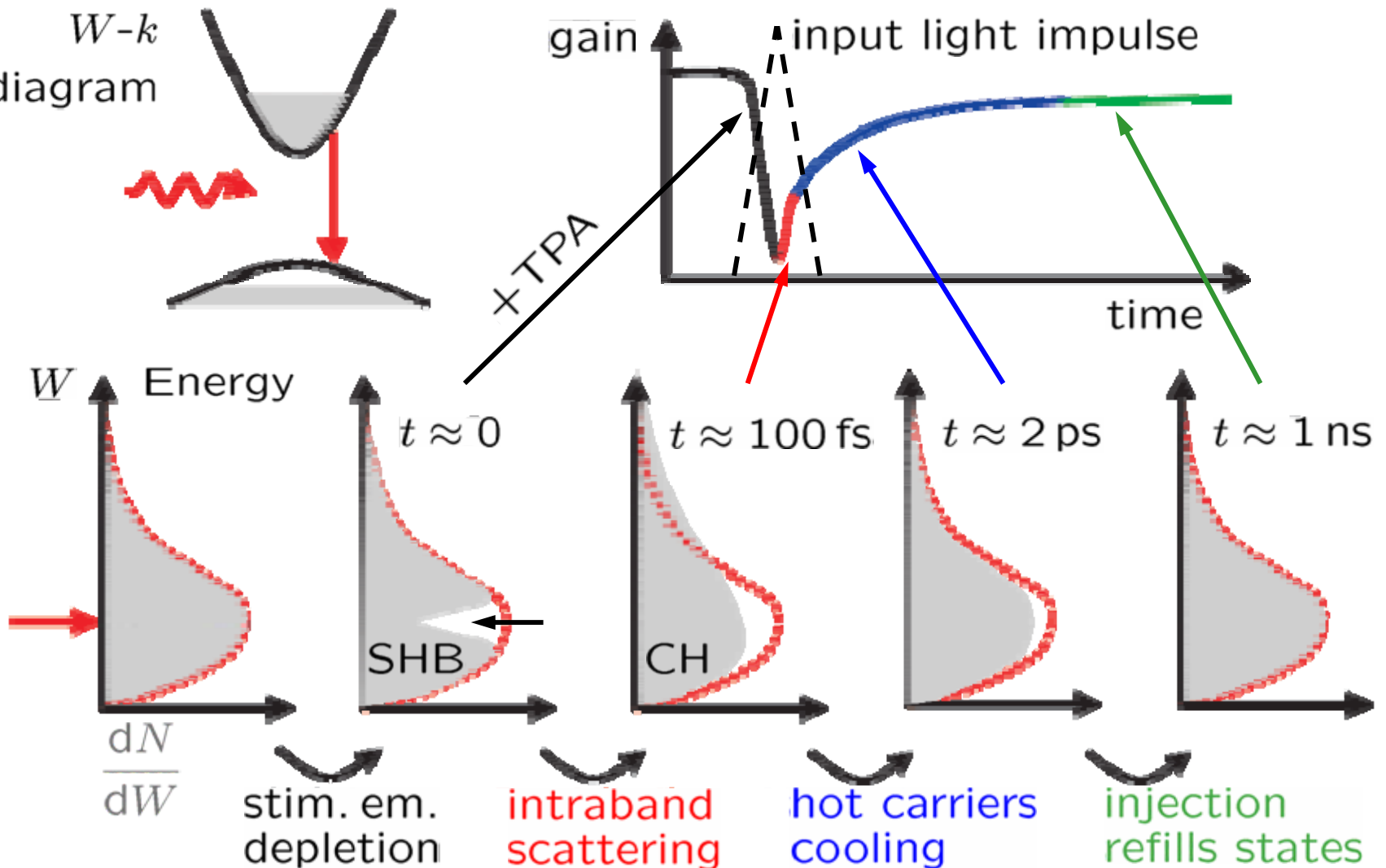
$$n(f) = 1 - \frac{2}{\pi} \int_0^{\infty} f' \frac{-n_i(f')}{f'^2 - f^2} df' = 1 - \frac{c}{2\pi^2} \int_0^{\infty} \frac{g(f')}{f'^2 - f^2} df'$$



Nussenzveig, H. M.: Causality and dispersion relations. Vol. 95 in "Mathematics in science and engineering", Ed. R. Bellmann. New York: Academic Press 1972. Sect. 1.6



# Carrier/Gain Depletion and Recovery in an SOA



Modified from: Mørk, J. et al. IEEE LEOS Newsletter 16 (2002) 21–24. Fig. 2. — Mørk, J. et al. Optics & Photonics News July (2003) 42–48



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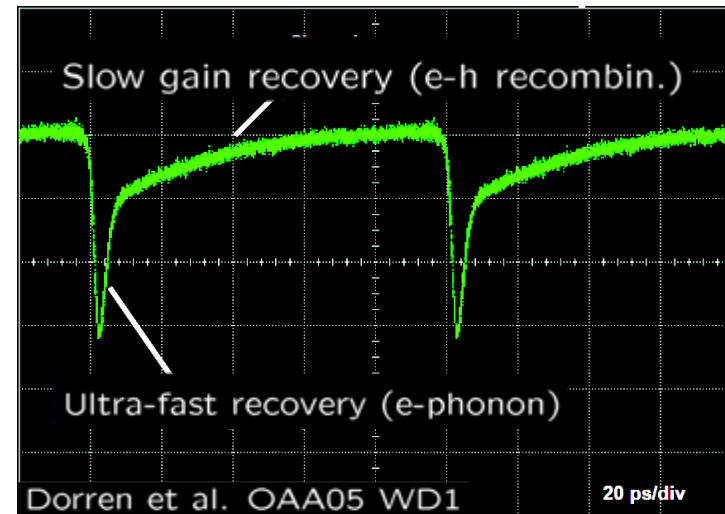
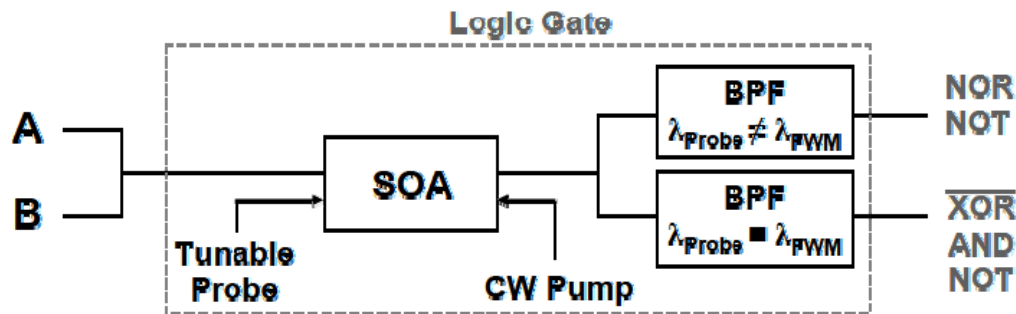




# Reconfigurable Optical Logic Gate with SOA

## Key ideas:

- Non-interferometric structure based on single SOA
- FWM and XGM exploited simultaneously
- Counter-propagating CW pump speeds up SOA gain recovery



Berrettini, G.; Malacarne, A.; Ghelfi, P.; Bogoni, A.; Potí, L.: Reconfigurable all-optical logic gate based on a single SOA with improved dynamics. Ofc 2006 Paper OFJ5

Berrettini, G.; Simi, A.; Malacarne, A.; Bogoni, A.; Potí, L.: Ultrafast integrable and reconfigurable XNOR, AND, NOR, and NOT photonic logic gate. IEEE Photon. Technol. Lett. 18 (2006) 917–919

Dorren, H. J. S.; Hill, M. T.; Liu, Y.; Tangdiongga, E.; Smit, M. K.; Khoe, G. D.: Optical signal processing and telecommunication applications. Optical Amplifiers and Their Applications (OAA), 7–10 August 2005, Budapest, Hungary. Paper WD1

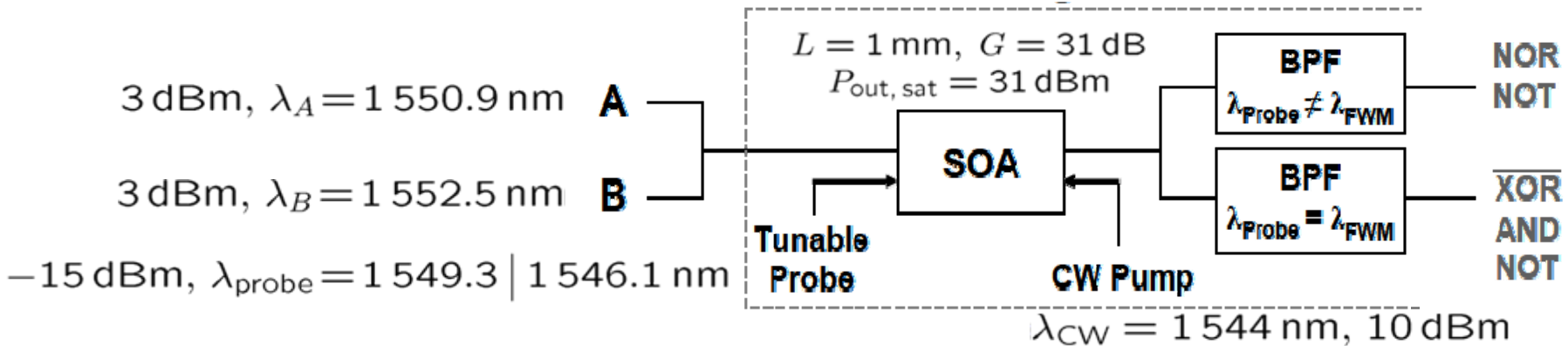
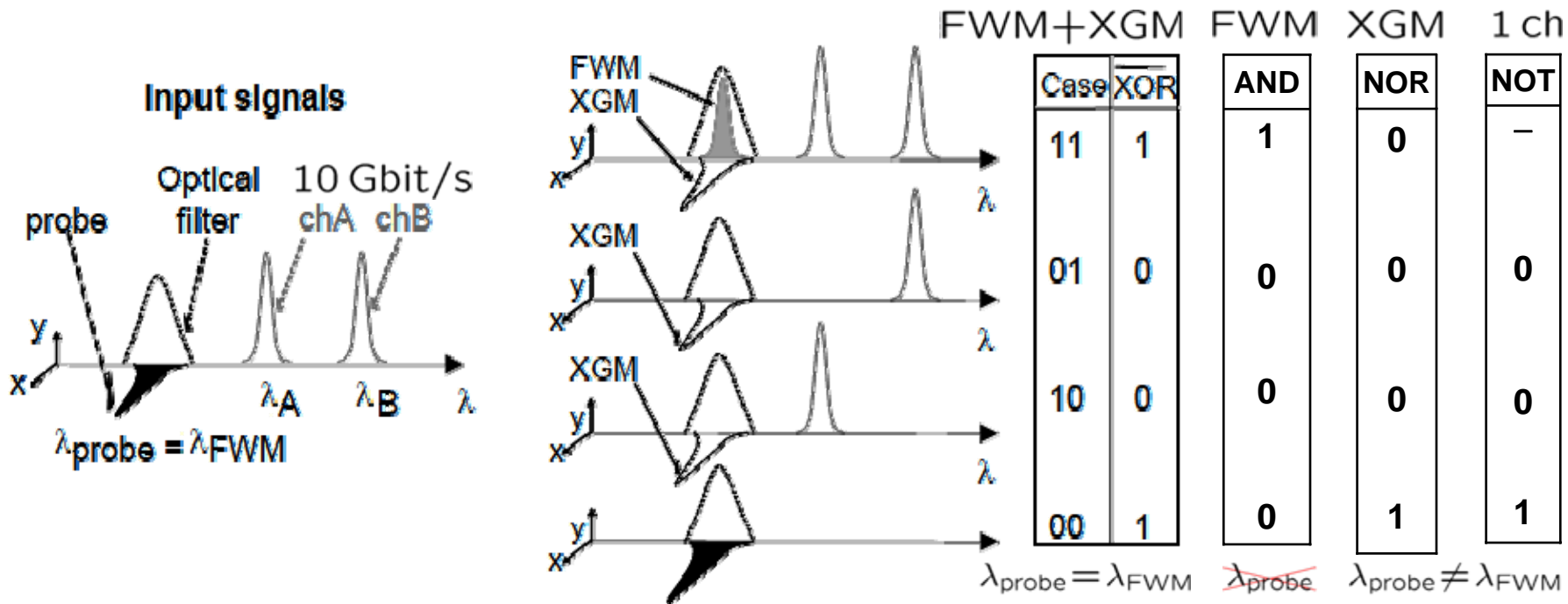
Dorren, H. J. S.; Yang, X.; Mishra, A. K.; Li, Z.; Ju, H.; de Waardt, H.; Khoe, G.-D.; Simoyama, T.; Ishikawa, H.; Kawashima, H.; Hasama, T.: All-optical logic based on ultrafast gain and index dynamics in a semiconductor optical amplifier. IEEE J. Sel. Topics Quantum Electron. 10 (2004) 1079–1092

Kang, I.; Dorrer C.; Leuthold, J.: All-optical XOR operation of 40 Gbit/s phase-shift-keyed data using four-wave mixing in semiconductor optical amplifier. Electron. Lett. 40 (2004) No. 8

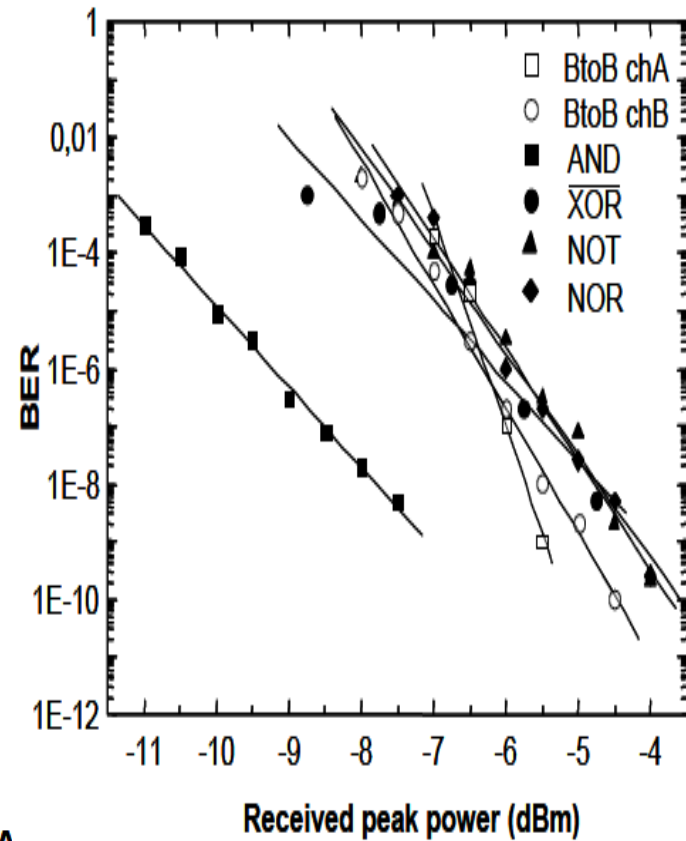
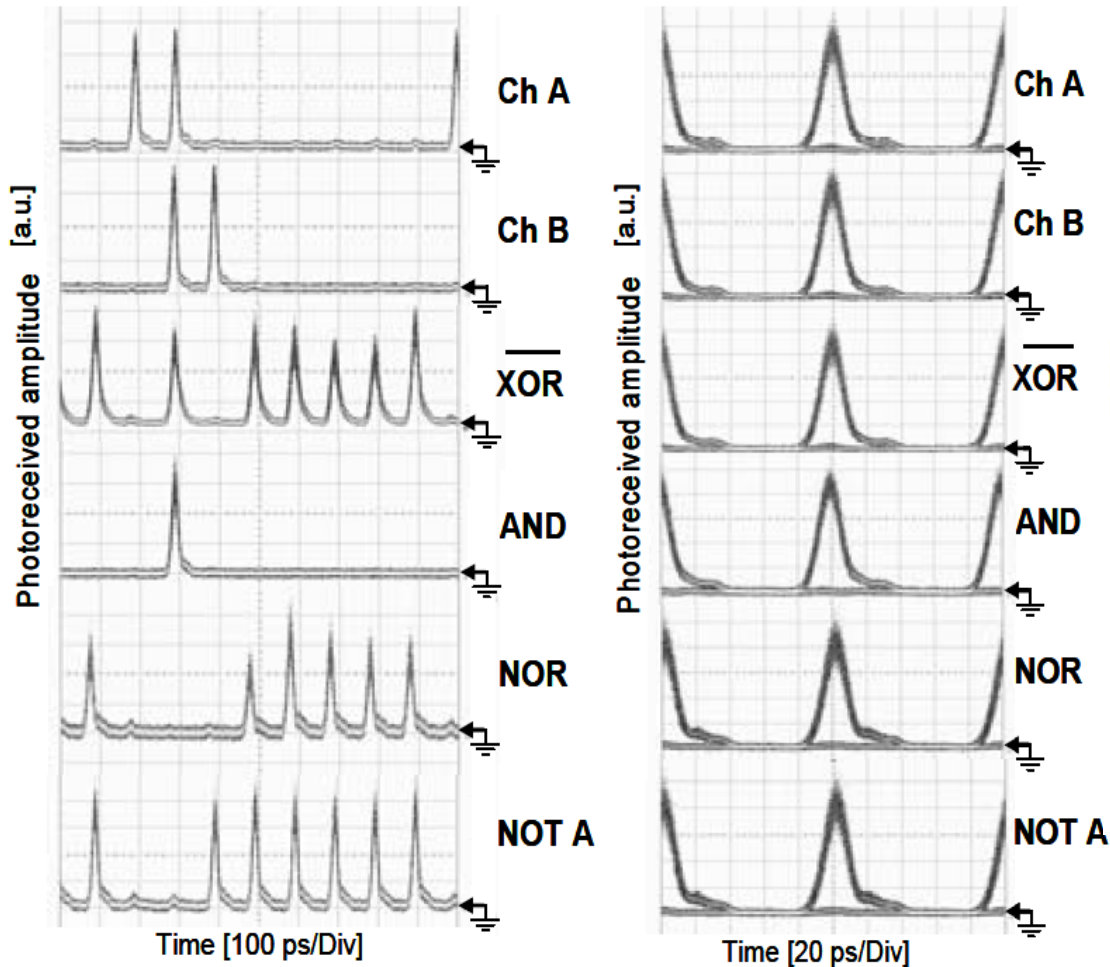




# Optical Logic XNOR Gate



# Optical Logic XNOR Gate — Performance at 10 Gbit/s



Power penalty 0.5 dB,  
AND with regeneration  
(-2 dB penalty, noise  
compression in SOA)

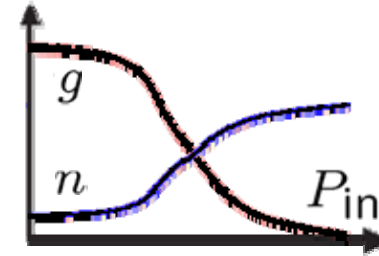
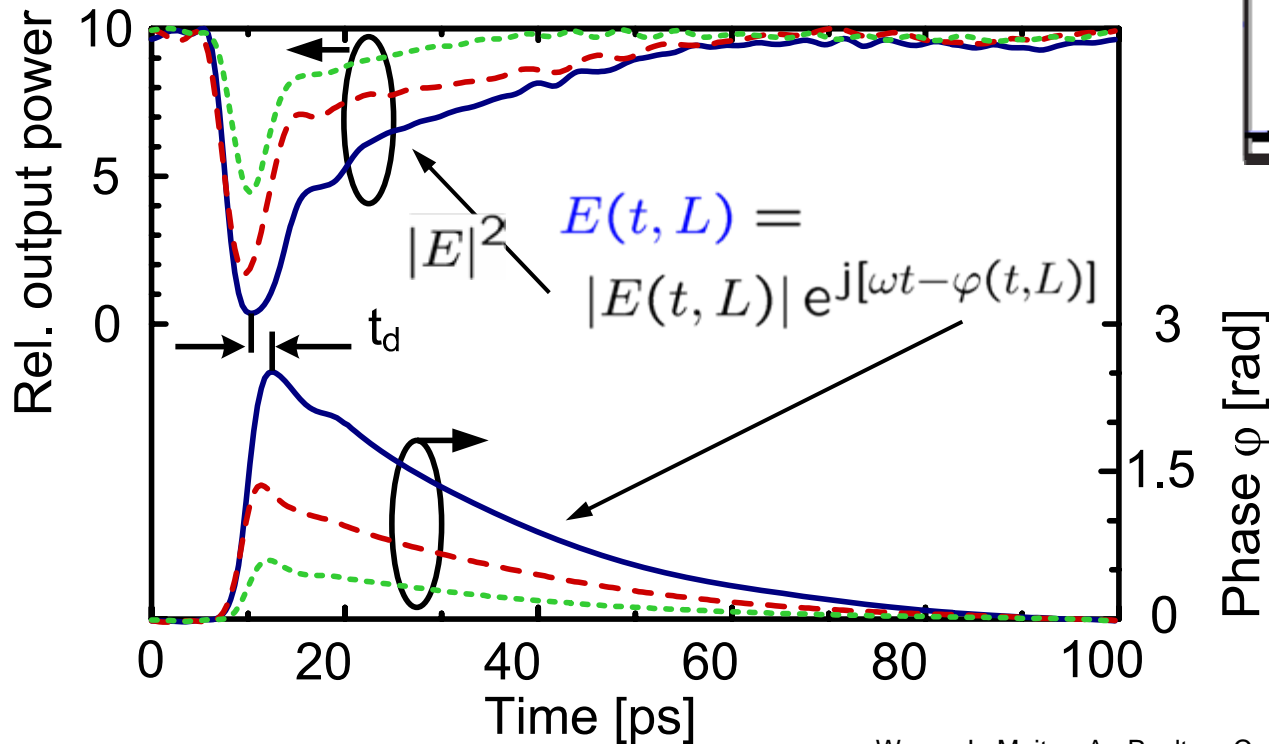
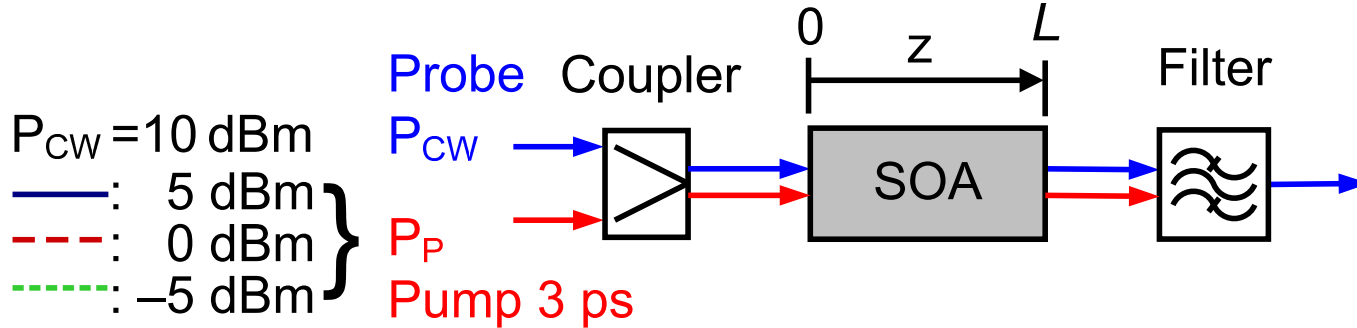


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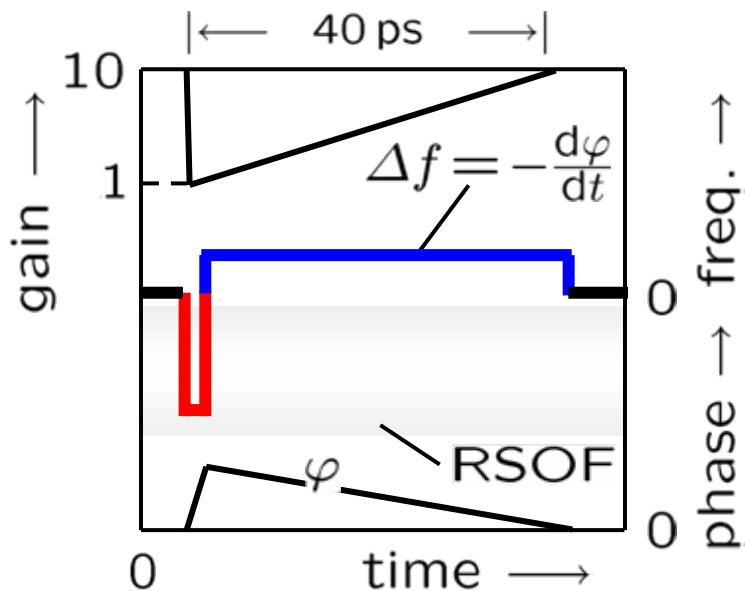
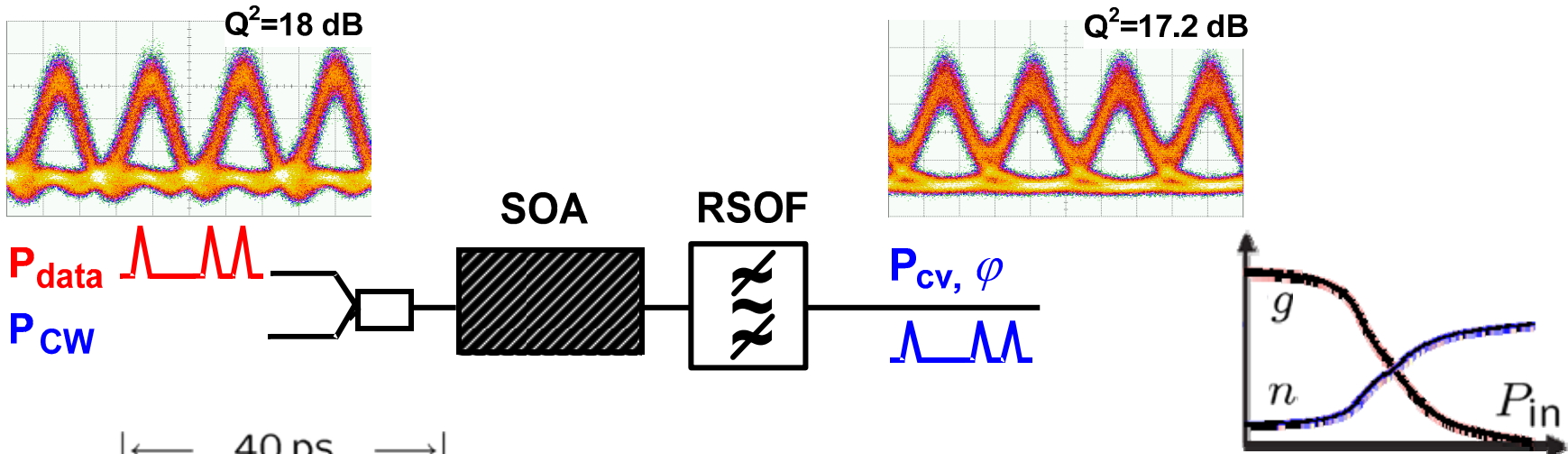
# Measured Output Power and Phase for an SOA



Wang, J.; Maitra, A.; Poulton, C. G.; Freude, W.; Leuthold, J.:  
 Temporal dynamics of the alpha factor in semiconductor  
 optical amplifiers. J. Lightw. Technol. (Aug. 2006, submitted)



# SOA $\lambda$ -Converter & Red Shift Optical Filter (RSOF)

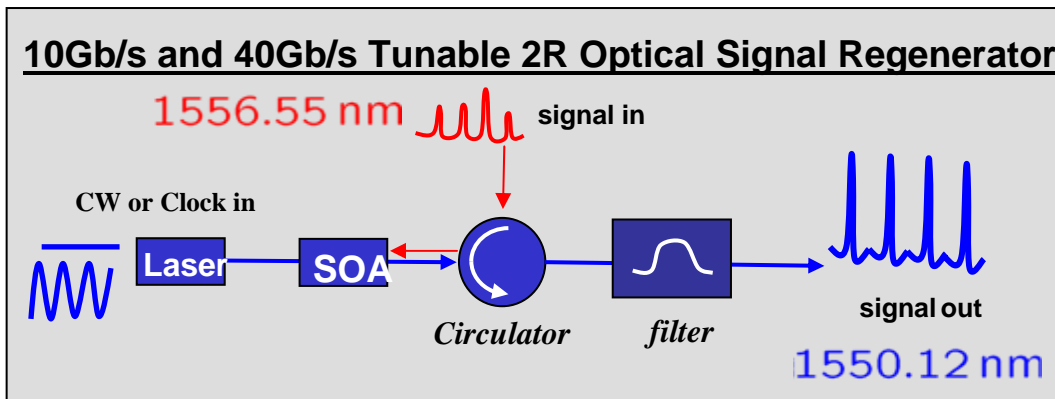


- $P_{data} = 0 \rightarrow P_{CV} = 0$  by filter
- $P_{data} \neq 0 \rightarrow$  red-chirped  $P_{CV}$  passes RSOF
- In effect, this is the action of a **low-pass filter (integrator)**.

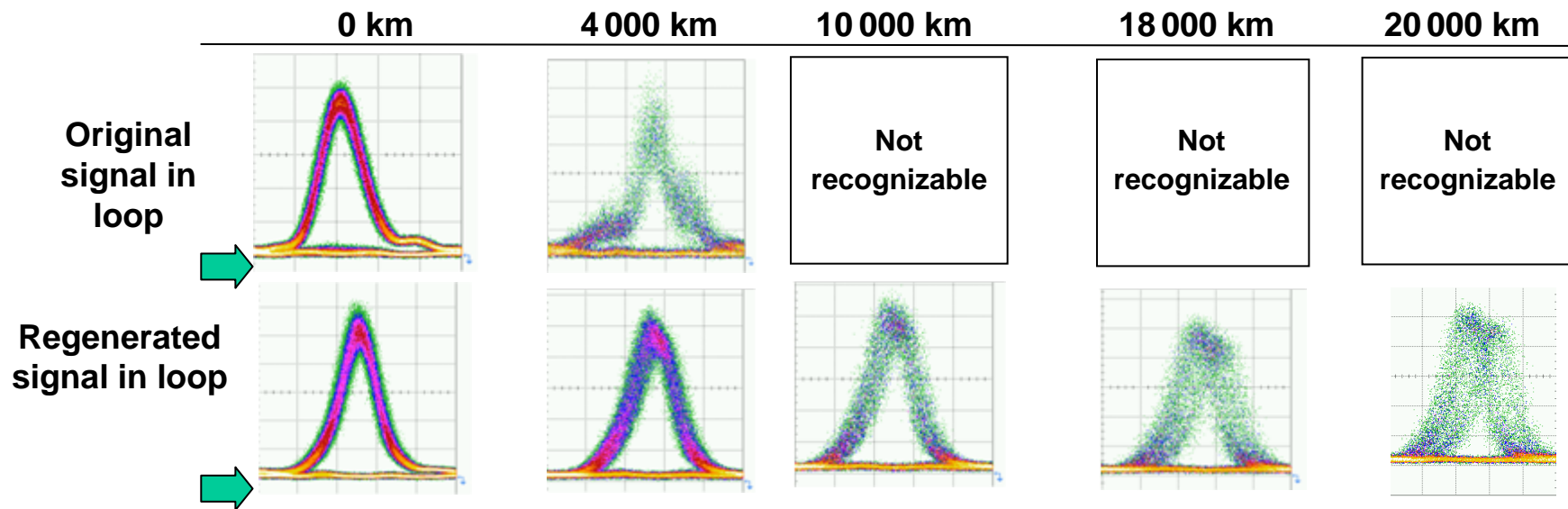
Leuthold, J.; Ryf, R.; Maywar, D. N.; Cabot, S.; Jaques, J.; Patel, S. S.: Regenerative all-optical wavelength converter in a transparent demonstration over 42 nodes and 16800 km. J. Lightwave Technol. 21 (2003) 2863–2870  
 Chayet, H.; Ben Ezra, S.; Shachar, N.; Tzadok, S.; Tsadka, S.; Leuthold, J.: Regenerative all-optical wavelength converter based on semiconductor optical amplifier and sharp frequency response filter. Ofc 2004 ThS2



# 2R or 3R Regeneration with SOA $\lambda$ -Converter and RSOF



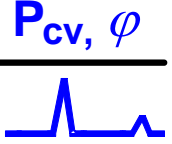
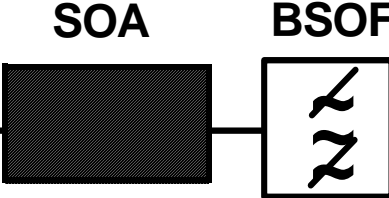
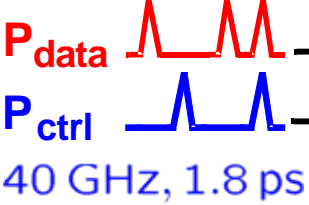
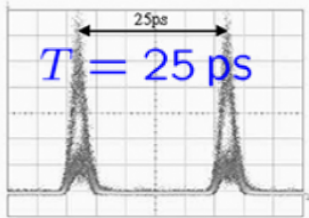
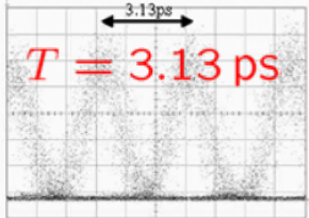
10 Gb/s, RZ,  
PRBS:  $2^{31}-1$



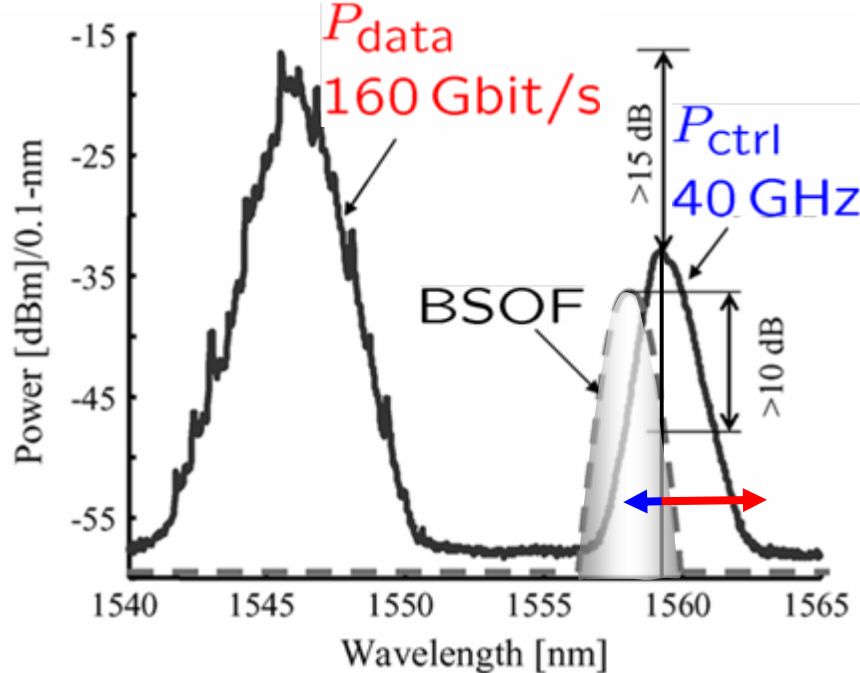
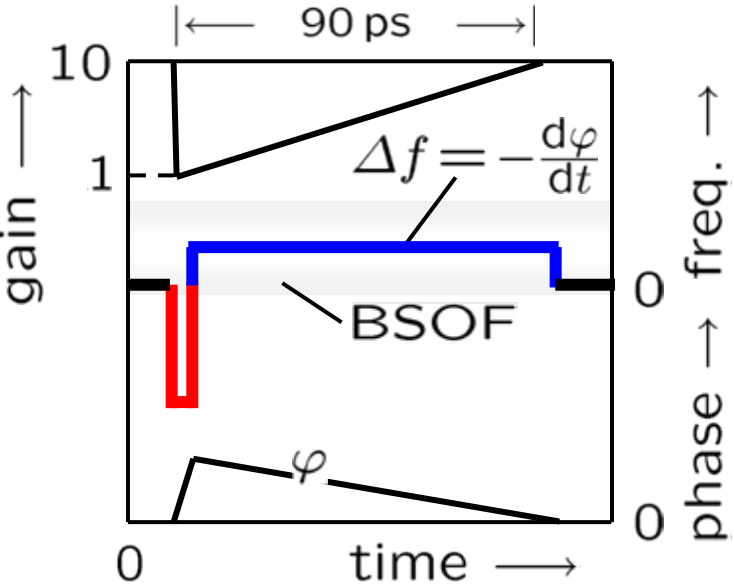
Chayet, H.; Ben Ezra, S.; Shachar, N.; Tzadok, S.; Tsadka, S.; Leuthold, J.: Regenerative all-optical wavelength converter based on semiconductor optical amplifier and sharp frequency response filter. Ofc 2004 Paper ThS2 (Kaillight Photonics)



# OOK 320-to-40 Gbits/s Demux & Blue Shift Optical Filter (BSOF)



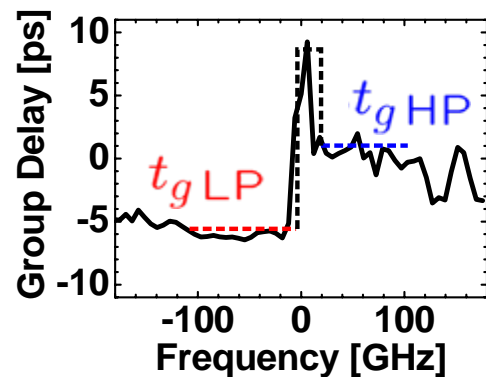
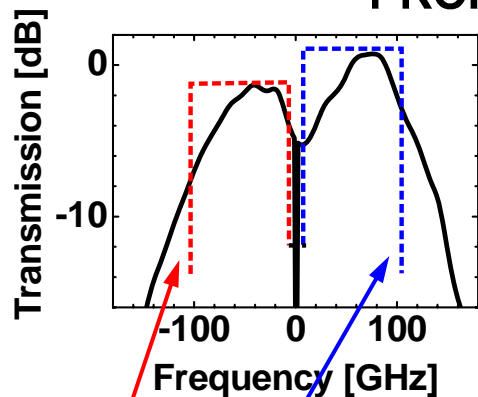
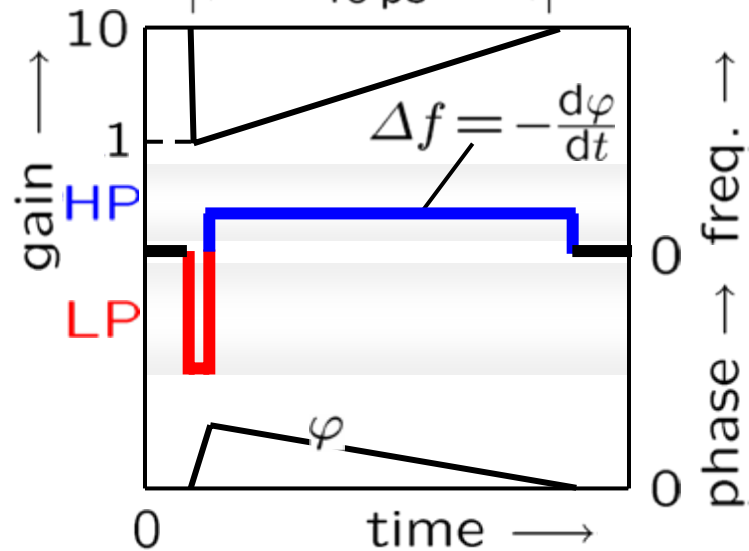
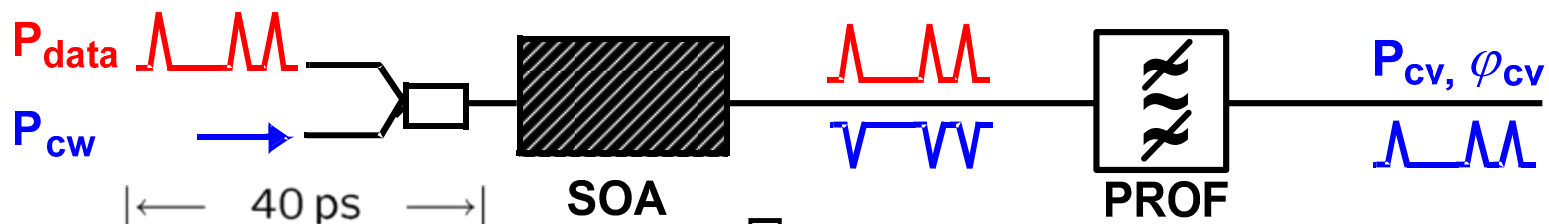
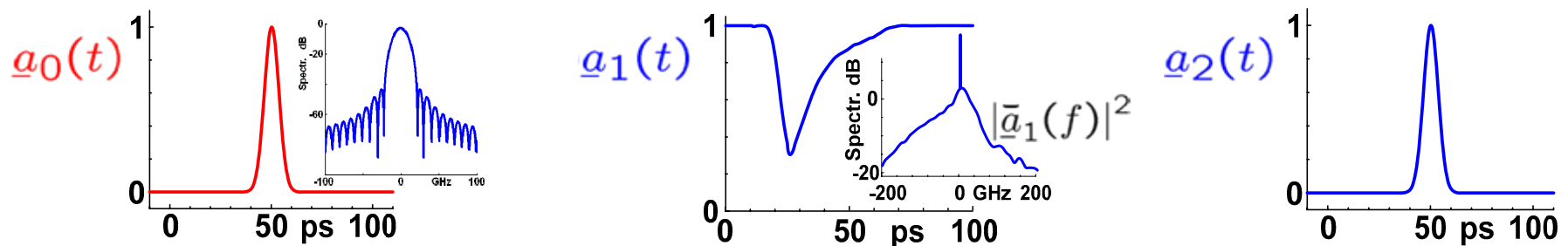
- In effect, this is the action of a high-pass filter (differentiator).



Tangdionga, E.; Liu, Y.; de Waardt, H.; Khoe, G. D.; Dorren, H. J. S.: 320-to-40-Gb/s demultiplexing using a single SOA assisted by an optical filter. IEEE Photon. Technol Lett. 18 (2006) 908–910



# SOA $\lambda$ -Converter & Pulse Reformatting Filter (PROF)



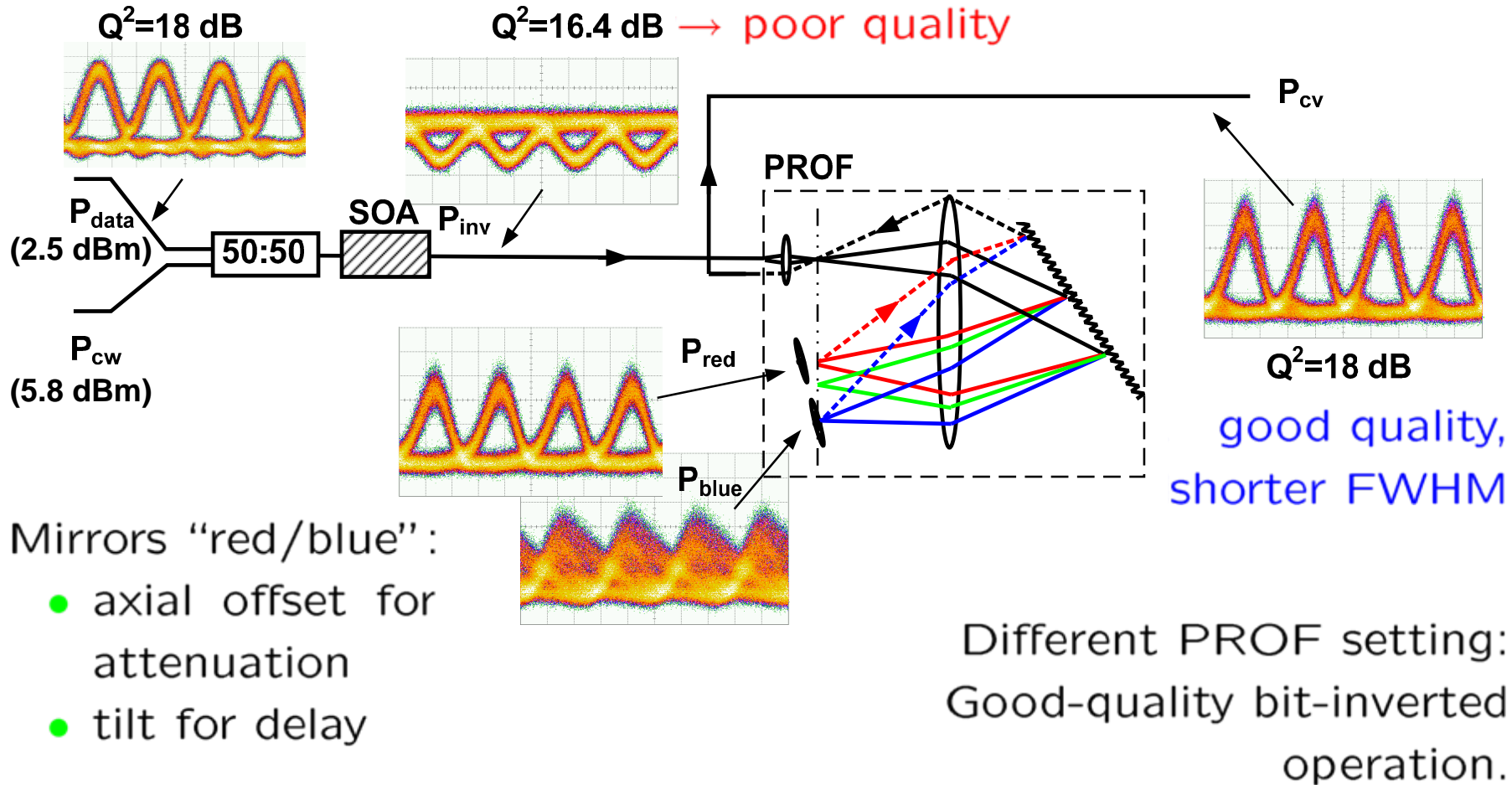
$$\bar{h}_{LP} + \bar{h}_{HP} \approx \bar{h}_{PROF}(f) = \frac{\bar{a}_2(f)}{\bar{a}_1(f)}$$

Leuthold, J.; Marom, D. M.; Cabot, S.; Jaques, J. J.; Ryf, R.; Giles, C. R.: All-optical wavelength conversion using a pulse reformatting optical filter. J. Lightwave Technol. 22 (2004) 186–192.





# Eye Diagrams for 40 Gbit/s Non-Inverted RZ→RZ $\lambda$ -Converter



Leuthold, J.; Marom, D. M.; Cabot, S.; Jaques, J. J.; Ryf, R.; Giles, C. R.: All-optical wavelength conversion using a pulse reformatting optical filter. J. Lightwave Technol. 22 (2004) 186–192.

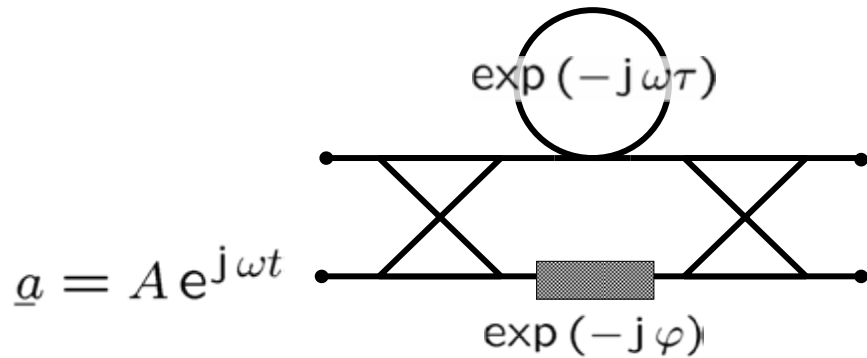


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# Mach-Zehnder Interferometer Field Transfer Function



$$a_\Sigma = j A e^{j\omega t} e^{-j\frac{\omega\tau+\varphi}{2}} \cos\left(\frac{\omega\tau-\varphi}{2}\right)$$

$$a_\Delta = j A e^{j\omega t} e^{-j\frac{\omega\tau+\varphi}{2}} \sin\left(\frac{\omega\tau-\varphi}{2}\right)$$

Field transfer functions:

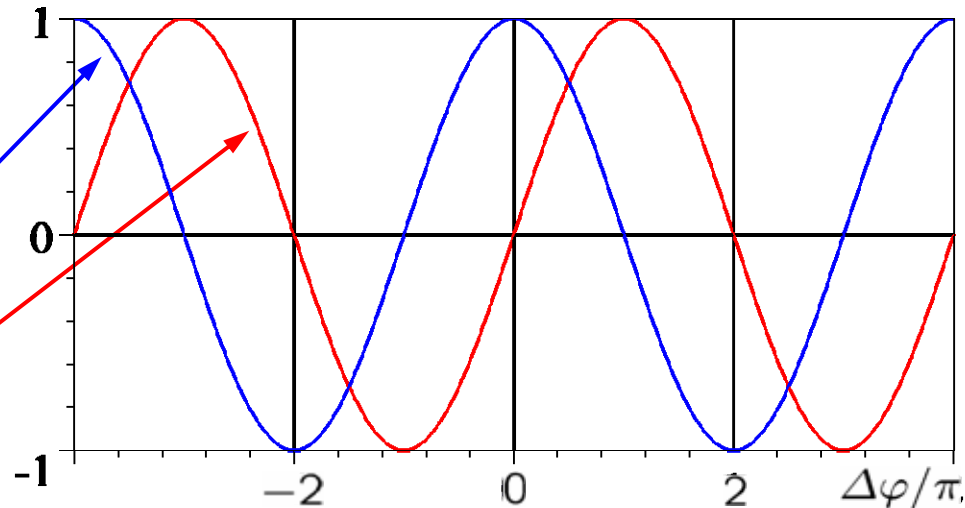
$$T_\Sigma(f) = \frac{a_\Sigma}{a} = j e^{-j\frac{\omega\tau+\varphi}{2}} \cos\left(\frac{\omega\tau-\varphi}{2}\right), \quad T_\Delta(f) = \frac{a_\Delta}{a} = j e^{-j\frac{\omega\tau+\varphi}{2}} \sin\left(\frac{\omega\tau-\varphi}{2}\right)$$

MZM:  $\omega_0\tau = \varphi_1$ ,  $\varphi = \varphi_2$ ,  
 $\varphi_1 + \varphi_2 = 2\phi$ ,  $\varphi_1 - \varphi_2 = \Delta\varphi$ :

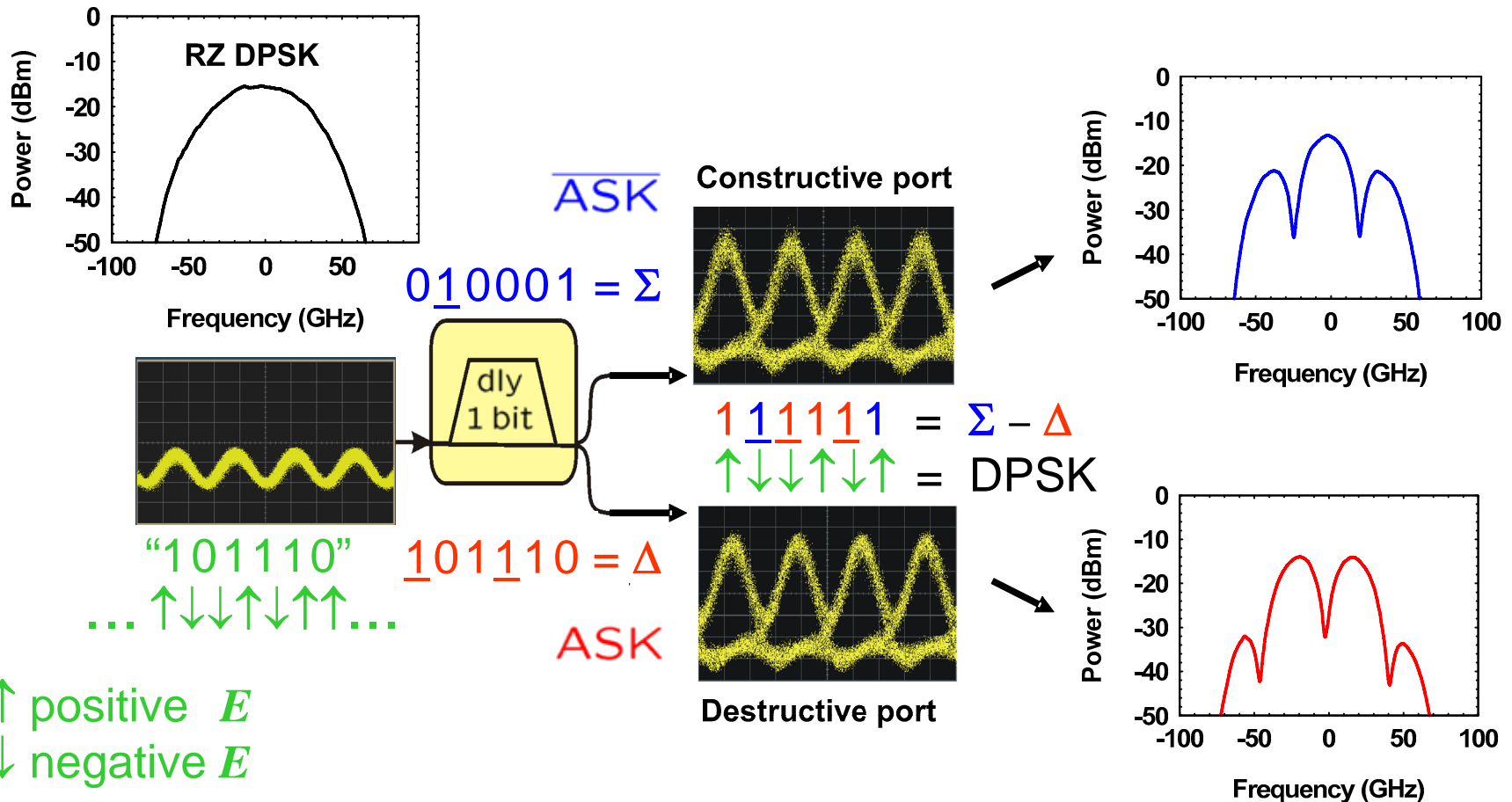
$$T_\Sigma(f_0, \tau) = j e^{-j\phi} \cos\left(\frac{\Delta\varphi}{2}\right)$$

$$T_\Delta(f_0, \tau) = j e^{-j\phi} \sin\left(\frac{\Delta\varphi}{2}\right)$$

Push-push:  $\phi(t)$  — Push-pull:  $\Delta\varphi(t)$



# DPSK Receiver — DPSK → ASK Conversion



DPSK phase noise → ASK amplitude (and phase) noise

DPSK amplitude noise → ASK amplitude noise

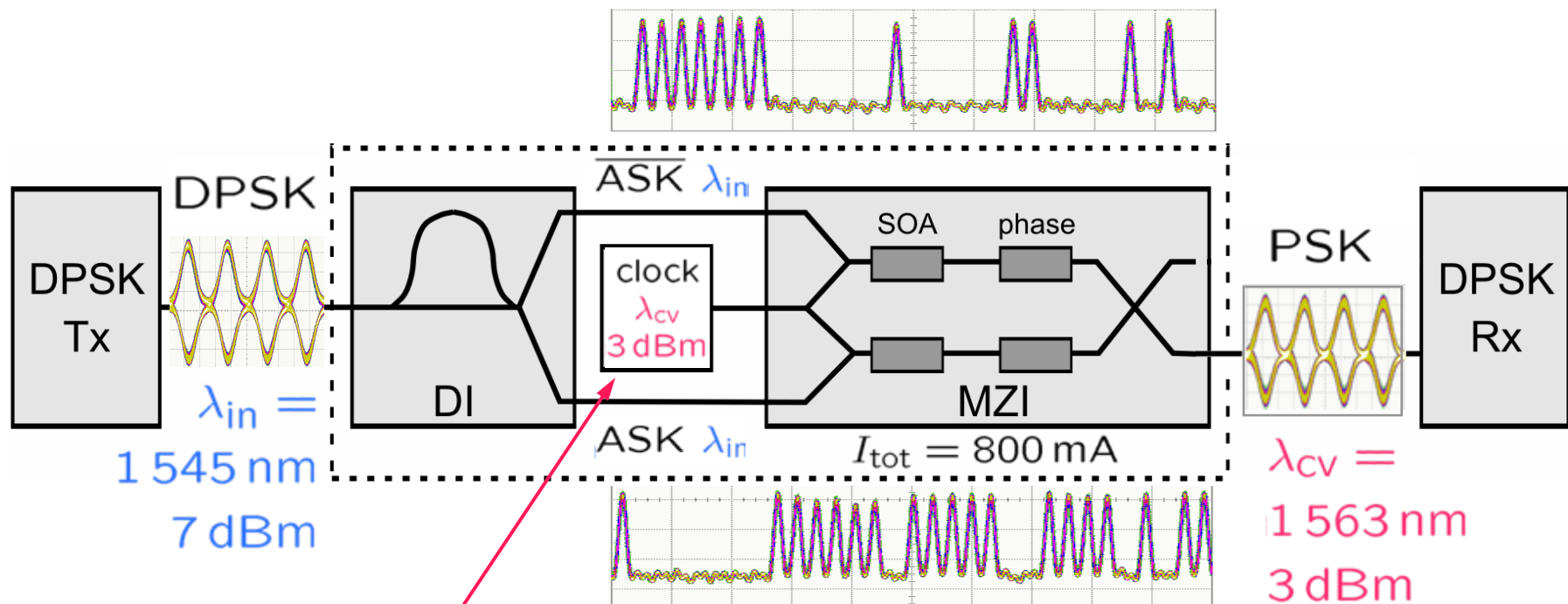


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# All-Optical DPSK $\lambda$ -Converter



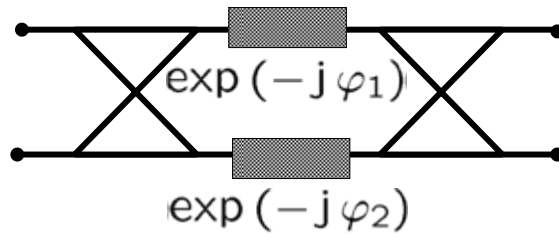
Clock  $\lambda_{cv}$  provides Retiming

Sartorius, B.; Bornholdt, C.; Slovak, J.; Schlak, M.; Schmidt, Ch.; Marculescu, A.; Vorreau, P.; Tsadka, S.; Freude, W.; Leuthold, J.: All-optical DPSK wavelength converter based on MZI with integrated SOAs and phase shifters. *Ofc 2006 OWS6*

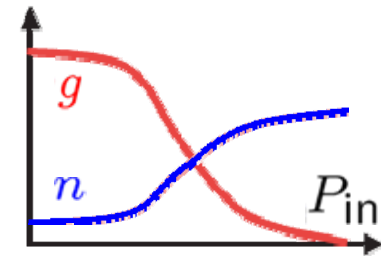
Kang, I.; Dorrer, C.; Zhang, L.; Rasras, M.; Buhl, L.; Bhardwaj, A.; Gomez, S.; Wong-Foy, A.; Chen, Y. F.; Patel, S.; Neilson, D. T.; Jaques, J.; Giles, C. R.: Regenerative all optic wavelength conversion of 40 Gb/s DPSK signals using a semiconductor optical amplifier Mach-Zehnder interferometer. *Ecoo 2005 Proc. 6 (2005) 29-31. PDP-Th4.3.3*



# DPSK $\lambda$ -Conversion Process by Cross-Phase Modulation (XPM)

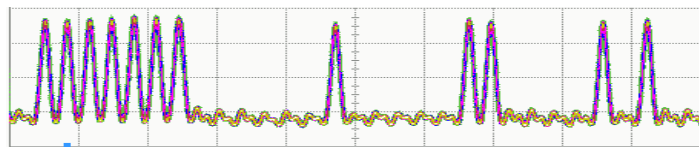


$$\varphi_{1,2} = k_0 n_{1,2} L$$



$\overline{\text{ASK}}$  data:

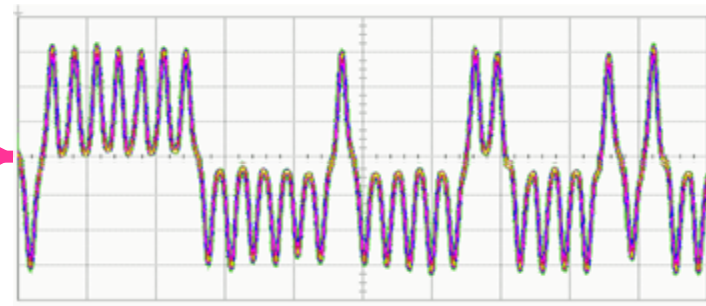
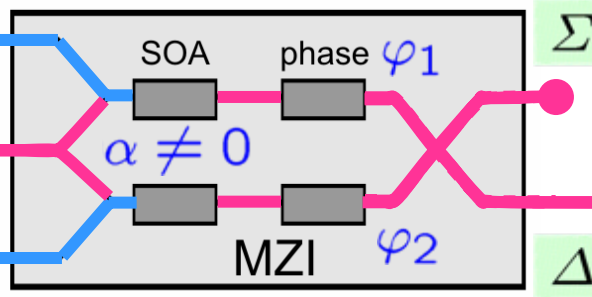
$$\varphi_1 + \pi \rightarrow \Delta = j e^{-j\frac{\pi}{2}} \sin\left(\frac{\pi}{2}\right) = +1$$



$\overline{\text{ASK}} \lambda_{in}$

clock  
 $\lambda_{cv}$

$\text{ASK} \lambda_{in}$



ASK data:

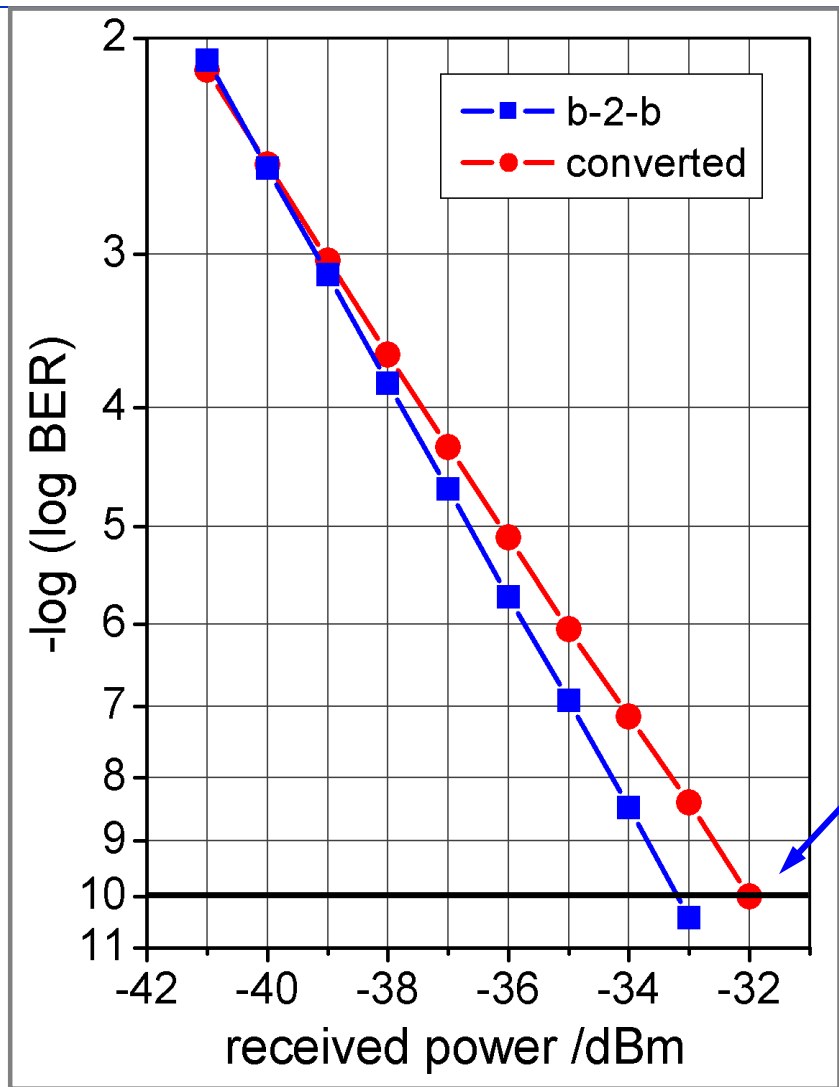
$$\varphi_2 + \pi \rightarrow \Delta = j e^{-j\frac{\pi}{2}} \sin\left(-\frac{\pi}{2}\right) = -1$$

No data,  $\varphi_1 = 0$ :

phase tuning  $\Delta\varphi = \varphi_1 - \varphi_2 = 0 \rightarrow \Delta = j e^{-j0} \sin\left(\frac{\Delta\varphi}{2}\right) = 0$



# DPSK $\lambda$ -Conversion Results for 31 Gbit/s



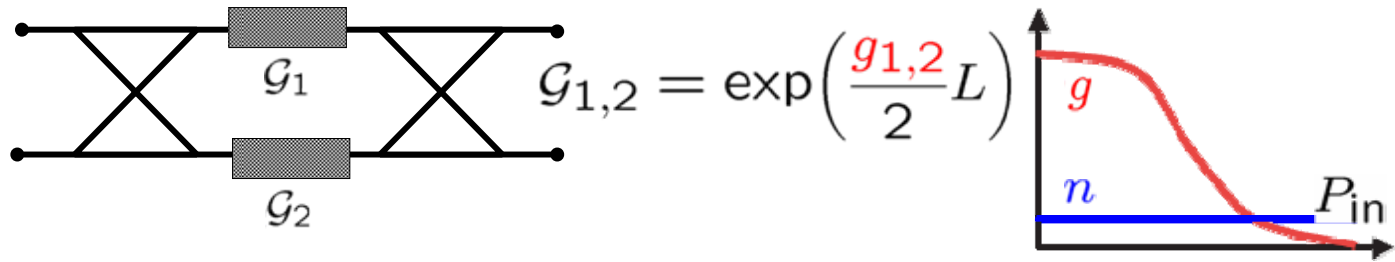
- Error-free  $\lambda$ -conversion
- For BER =  $10^{-10}$   
1 dB penalty only

Sartorius, B.; Bornholdt, C.; Slovak, J.; Schlak, M.; Schmidt, Ch.; Marculescu, A.; Vorreau, P.; Tsadka, S.; Freude, W.; Leuthold, J.: All-optical DPSK wavelength converter based on MZI with integrated SOAs and phase shifters. *Ofc 2006 OWS6*



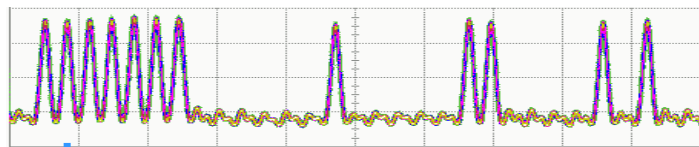


# DPSK $\lambda$ -Conversion Process by Cross-Gain Modulation (XGM)



ASK data:

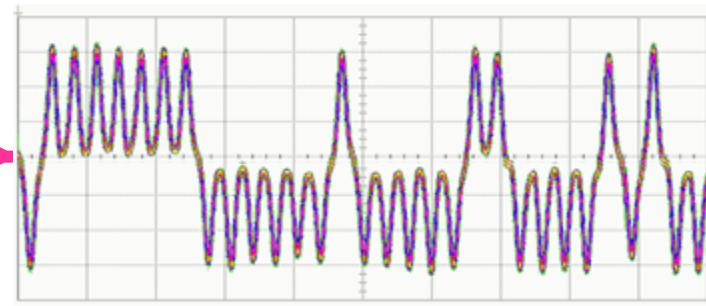
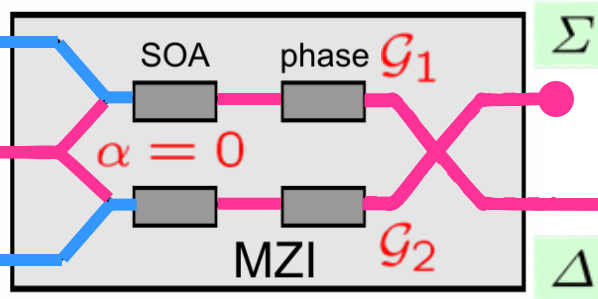
$$G_1 - \delta G \rightarrow \Delta = +\delta G$$



ASK  $\lambda_{in}$

clock  
 $\lambda_{cv}$

ASK  $\lambda_{in}$



ASK data:

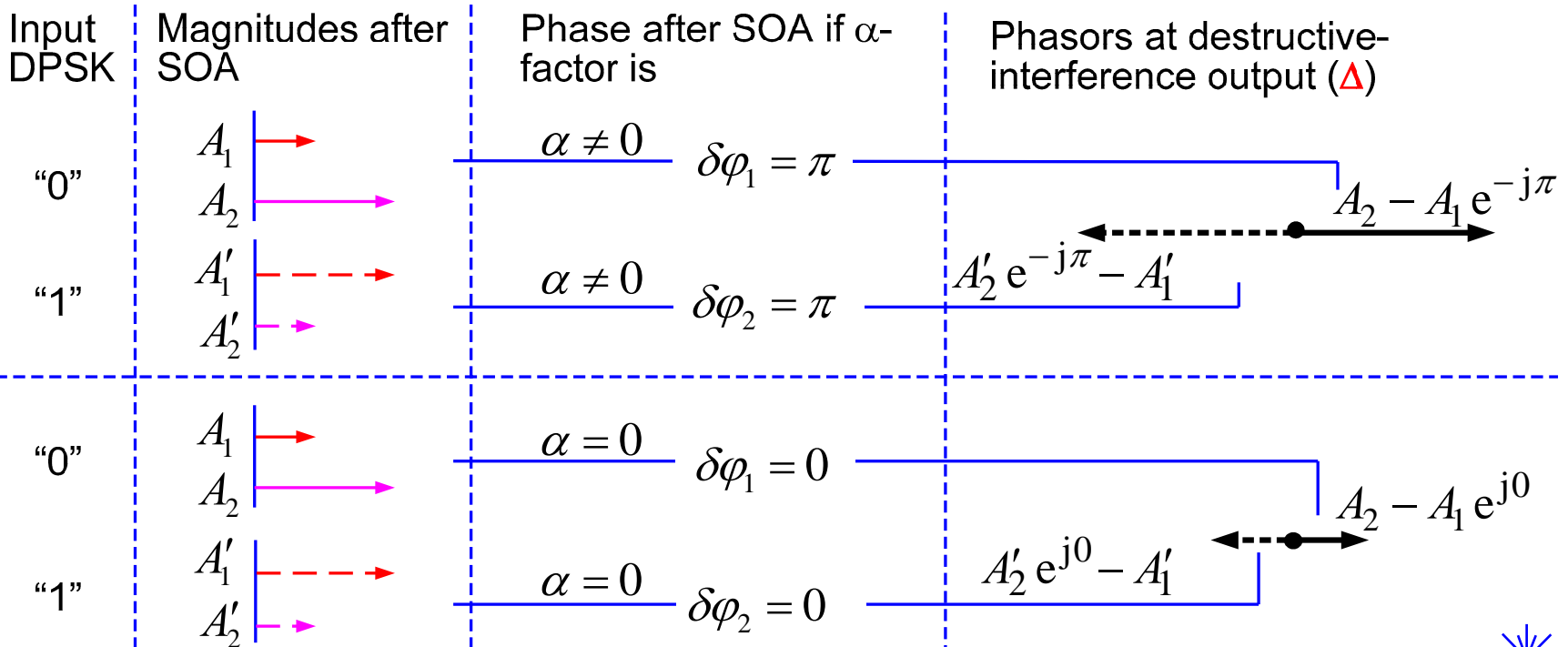
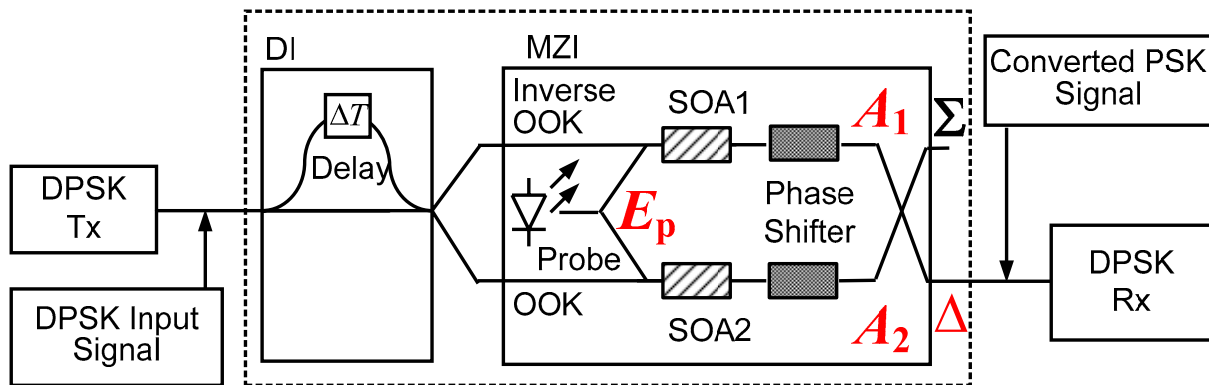
$$G_2 - \delta G \rightarrow \Delta = -\delta G$$

No data,  $\varphi_{1,2} = 0$ :

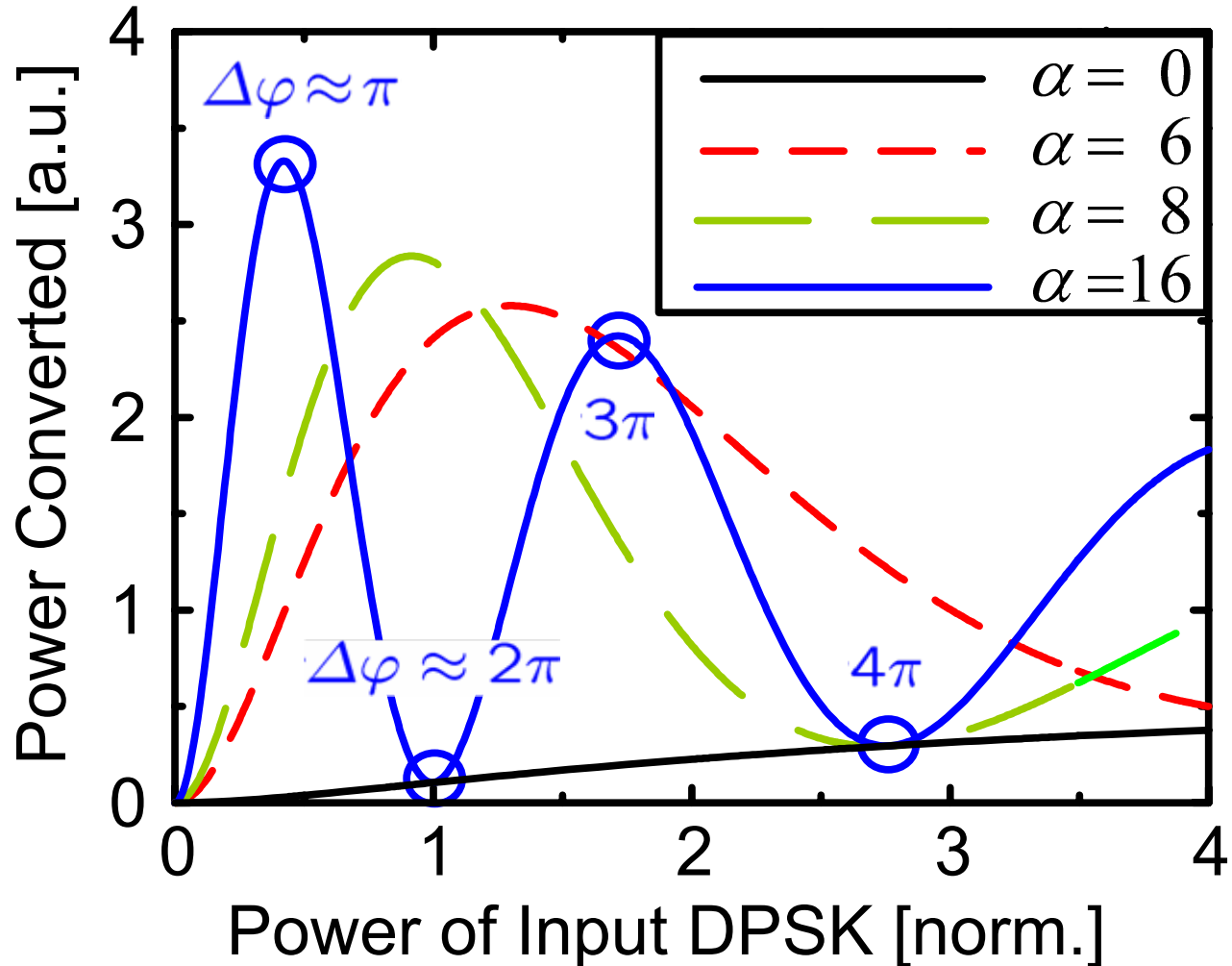
gain tuning  $\Delta G = G_2 - G_1 = 0 \rightarrow \Delta = \Delta G = 0$



# DPSK $\lambda$ -Conversion by (XPM & XGM) or XGM

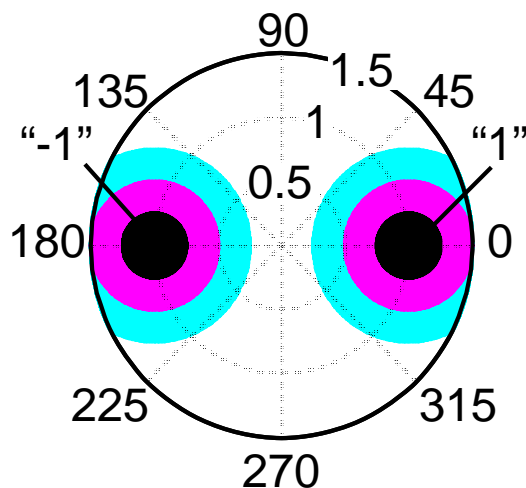


# DPSK $\lambda$ -Converter Operating Point for Various SOA $\alpha$ -Factors



# DPSK $\lambda$ -Converter — Symbol Diagrams and Regeneration

Input DPSK signal

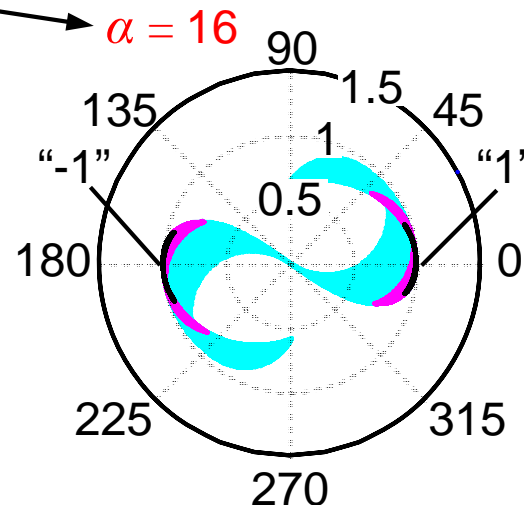
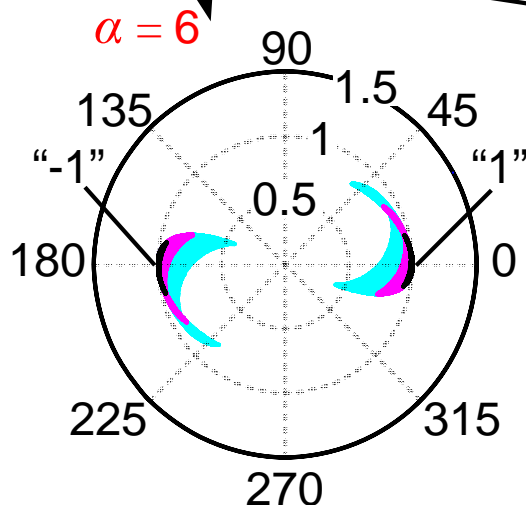
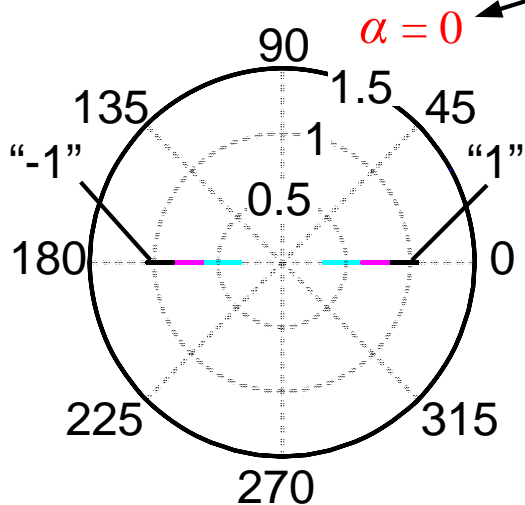


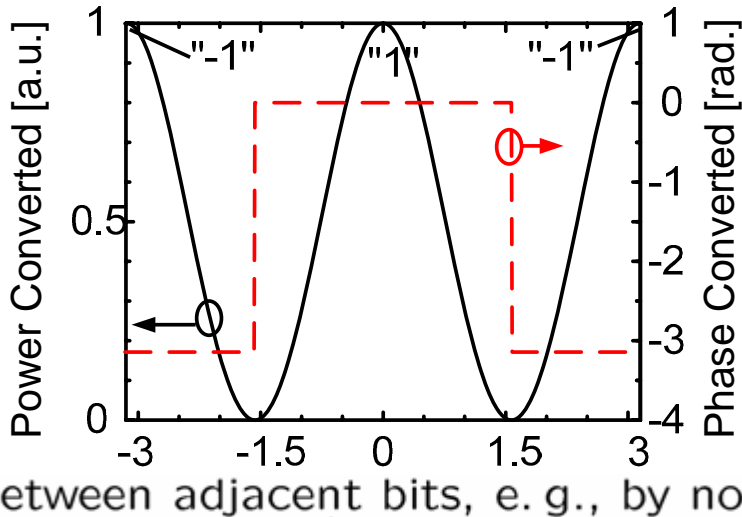
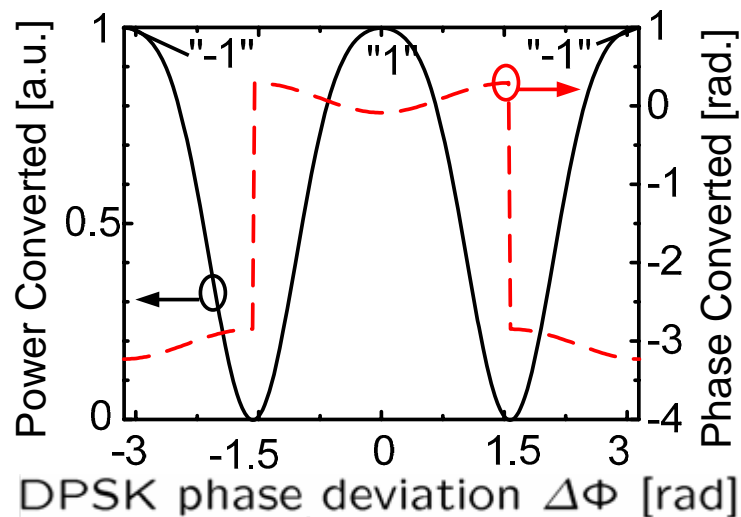
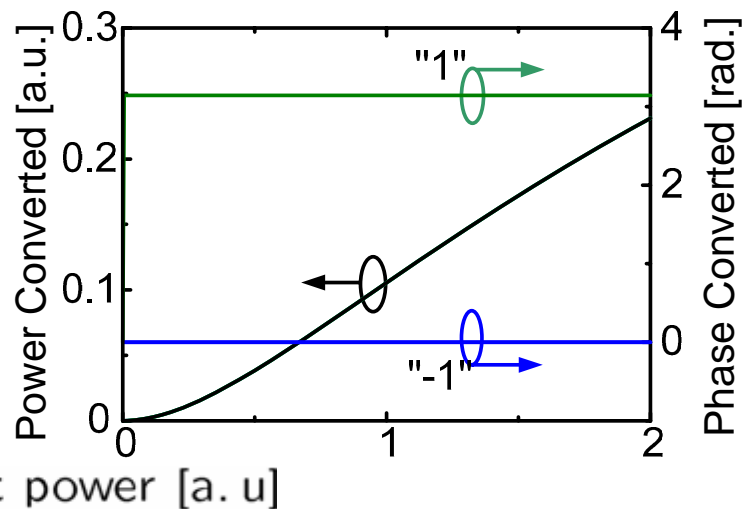
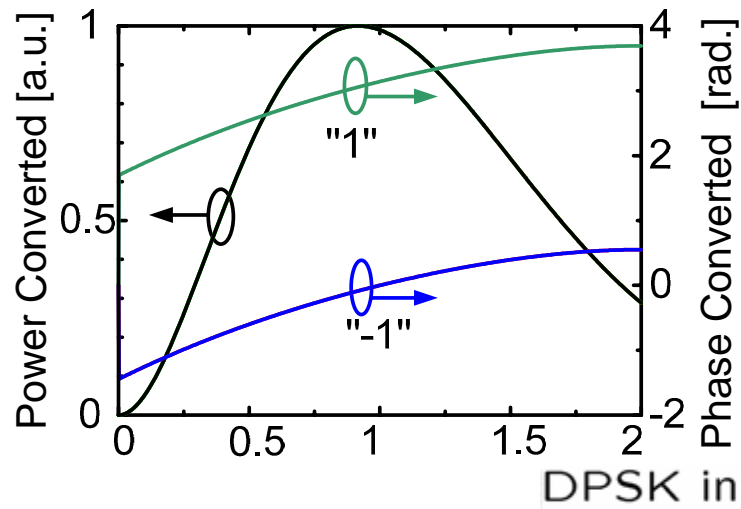
Shaded areas show phase and amplitude jitter of input DPSK signal. Jitter  $(dA/A_s) \cdot \exp(j\varphi)$  with

- $dA/A_s$ : 75 %
- $dA/A_s$ : 50 %
- $dA/A_s$ : 25 %

and random  $\varphi \in [-\pi, \pi]$ .  
 $A_s$  is the operating point amplitude.

Converted PSK signal

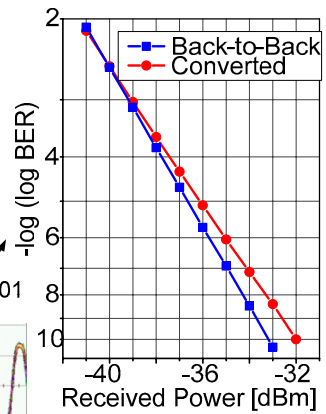
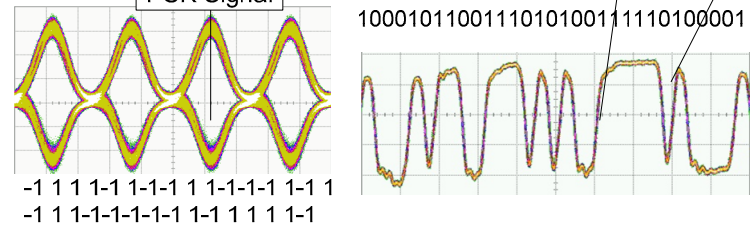
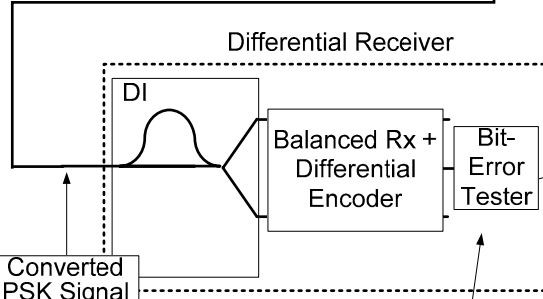
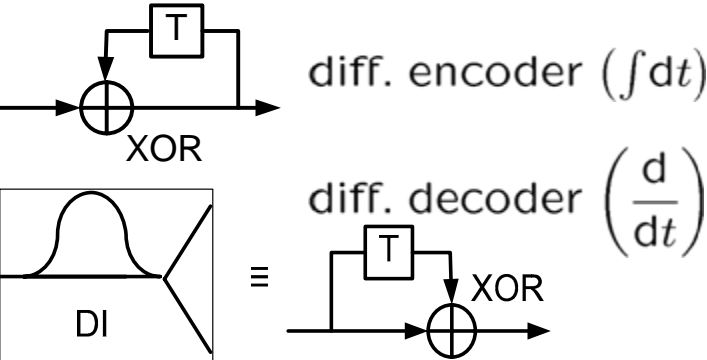
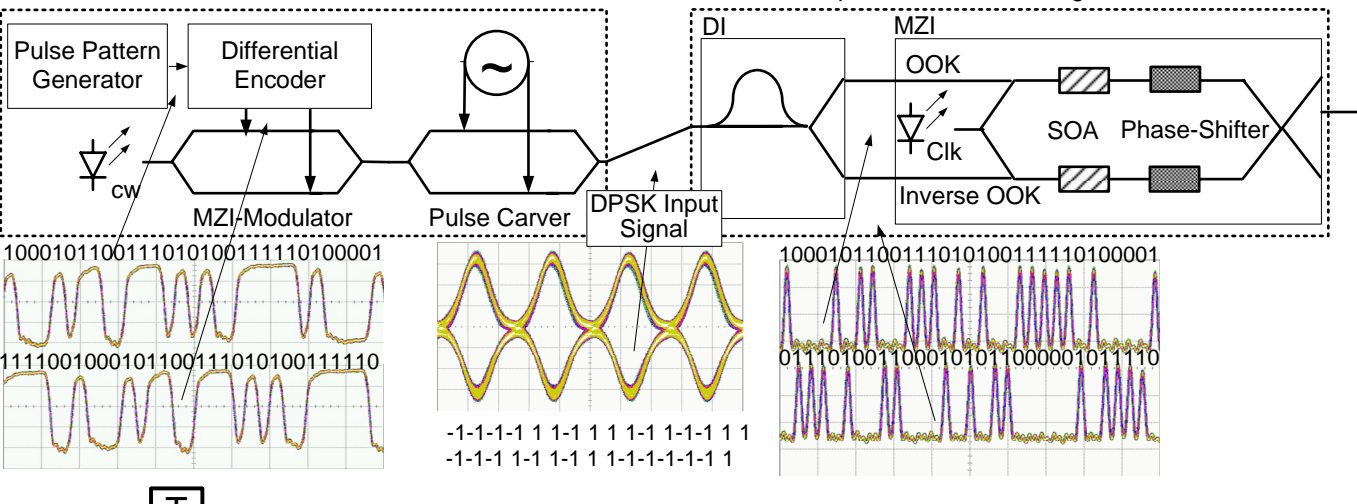


$\alpha = 8$ DPSK  $\lambda$ -Converter — Regeneration $\alpha = 0$ 

# DPSK $\lambda$ -Converter — Set-Up and Performance

DPSK Transmitter

All Optical DPSK Wavelength Converter



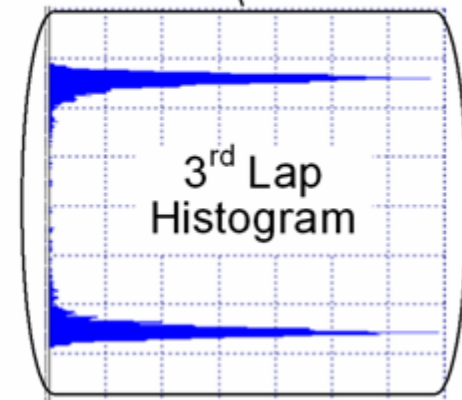
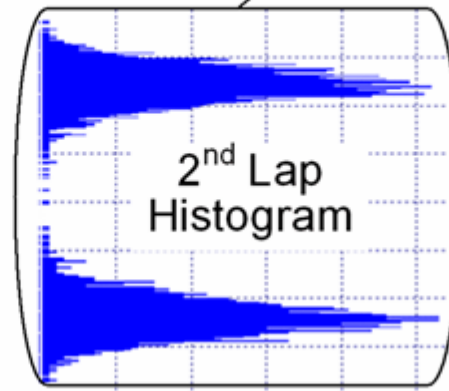
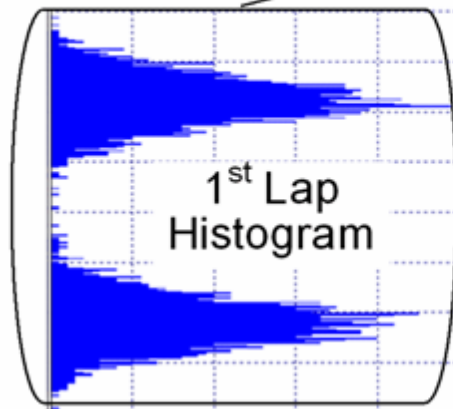
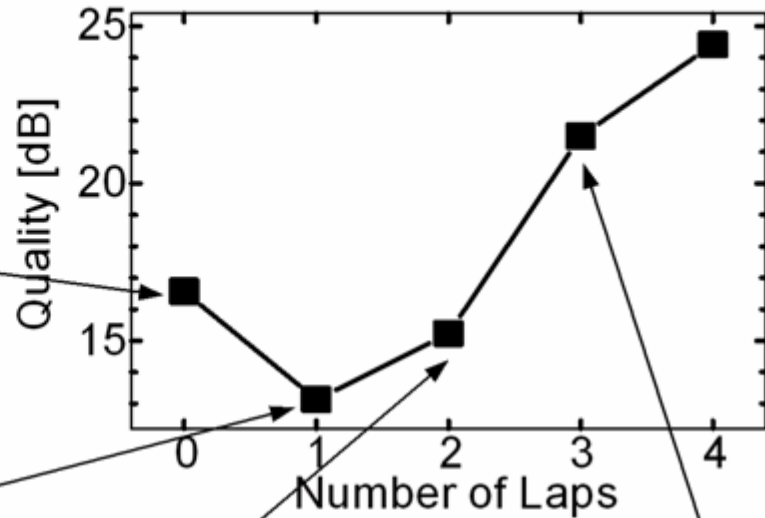
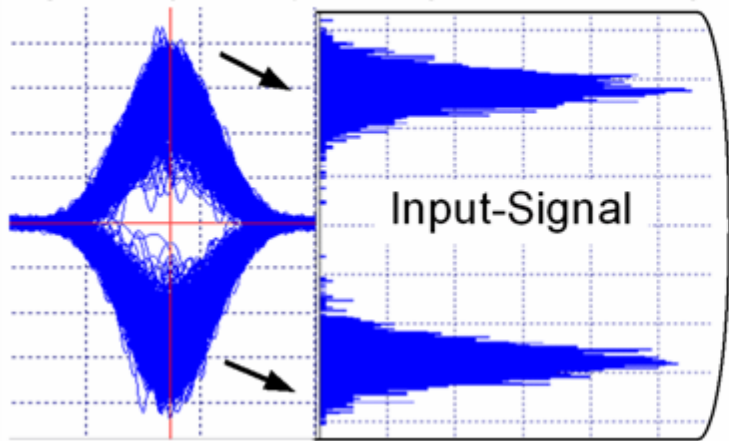
Any number of AOWC can be cascaded for an equal number of DI and differential encoders

Vorreau, P.; Marculescu, A.; Wang, J.; Böttger, G.; Sartorius, B.; Bornholdt, C.; Slovak, J.; Schlak, M.; Schmidt, Ch.; Tsadka, S.; Freude, W.; Leuthold, J.: Cascadability and regenerative properties of SOA all-optical DPSK wavelength converters. IEEE Photon. Technol. Lett. (2006) (in press)



# DPSK $\lambda$ -Converter — Cascadability

Input Signal Eye-Diagram & Histogram



Vorreau, P.; Marculescu, A.; Wang, J.; Böttger, G.; Sartorius, B.; Bornholdt, C.; Slovak, J.; Schlak, M.; Schmidt, Ch.; Tsadka, S.; Freude, W.; Leuthold, J.: Cascadability and regenerative properties of SOA all-optical DPSK wavelength converters. IEEE Photon. Technol. Lett. (2006) (in press)



# Outline

- Modulation techniques
  - Analogue, digital, coding
  - Symbol diagrams, spectra
  - Benefits, transmission capacity
- SOA gain and phase recovery
  - Gain-phase coupling
  - Physical explanation
- SOA signal processing
  - Logic gate
  - OOK wavelength converter
  - DPSK wavelength converter
- **Summary**





# Summary

Standard modulation format:

- OOK

Novel modulation formats:

- Phase shaping OOK
- DPSK

SOA gain and phase recovery:

- Ultrafast
- Logic gates

SOA wavelength converter:

- OOK with RSOF, BSOF, PROF
- DPSK

SOA DPSK wavelength converter:

- Regeneration
- Cascadability

