

# Wavelet FDTD Methods and Applications in Nano-Photonics

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# Outline

- **Wavelets**  
What are they good for?
- **Finite-differences in time-domain**  
Yee's leapfrog algorithm  
Numerical dispersion, stability and accuracy  
Higher-order finite-differences  
Method of weighted residuals: Collocation
- **Wavelet FDTD**  
Numerical dispersion, stability and accuracy  
What should be further done?  
Dispersive and nonlinear media
- **Application examples in nano-photonics**  
Waveguide roughness  
Slow light in a photonic crystal with disorder  
Four-wave mixing in a microring resonator  
Switching of a Bragg grating
- **Summary and further reading**



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- **Wavelets**
  - What are they good for? **Notation**
- Finite-differences in time-domain
  - Yee's leapfrog algorithm
  - Numerical dispersion, stability and accuracy
  - Higher-order finite-differences
  - Method of weighted residuals: Collocation
- Wavelet FDTD
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Convolution:

$$v(t) * w(t) = \int_{-\infty}^{+\infty} v(t) \underbrace{w(t' - t)}_{\downarrow} dt = \int_{-\infty}^{+\infty} \bar{v}(f) \bar{w}(f) e^{j2\pi ft'} df$$

Correlation:

$$v(t) \otimes w(t) = \int_{-\infty}^{+\infty} v(t) \overbrace{w^*(t - t')}^{\uparrow} dt = \int_{-\infty}^{+\infty} \bar{v}(f) \bar{w}^*(f) e^{j2\pi ft'} df$$

Inner product:

$$\langle w(t) | v(t) \rangle = \int_{-\infty}^{+\infty} v(t) w^*(t) dt = \int_{-\infty}^{+\infty} \bar{v}(f) \bar{w}^*(f) df$$



# Fourier Transform



$$\bar{v}(f) = \langle e^{+j2\pi ft} | v(t) \rangle = \int_{-\infty}^{+\infty} v(t) e^{-j2\pi ft} dt$$

$$v(t) = \langle e^{-j2\pi ft} | \bar{v}(f) \rangle = \int_{-\infty}^{+\infty} \bar{v}(f) e^{+j2\pi ft} df$$



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- **Wavelets**
  - What are they good for? **Fourier and uncertainty**
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# Fourier Transform and Uncertainty

A well localized time function  $v_s(t) = \frac{1}{\sqrt{s}} v(t/s)$  (small  $s$ ) has a widespread Fourier spectrum  $\bar{v}_s(f) = \sqrt{s} \bar{v}(sf)$ :

$$\bar{v}_s(f) = \int_{-\infty}^{+\infty} \underbrace{\frac{1}{\sqrt{s}} v(t/s)}_{v_s(t)} e^{-j2\pi ft} dt = \sqrt{s} \bar{v}(sf),$$

Energy conservation:  $W_t = \int_{-\infty}^{+\infty} |v_s(t)|^2 dt = \int_{-\infty}^{+\infty} |\bar{v}_s(f)|^2 df = W_f$

Expected location in time and frequency domain:

$$\bar{t} = \frac{1}{W_t} \int_{-\infty}^{+\infty} t |v_s(t)|^2 dt, \quad \bar{f} = \frac{1}{W_f} \int_{-\infty}^{+\infty} f |\bar{v}_s(f)|^2 df$$

Location variances in time and frequency domain:

$$\sigma_t^2 = \frac{1}{W_t} \int_{-\infty}^{+\infty} (t - \bar{t})^2 |v_s(t)|^2 dt, \quad \sigma_f^2 = \frac{1}{W_f} \int_{-\infty}^{+\infty} (f - \bar{f})^2 |\bar{v}_s(f)|^2 df$$

Uncertainty of time and frequency spread:  $\sigma_t^2 (2\pi\sigma_f)^2 \geq \frac{1}{4}$





# Fourier Transform, Uncertainty and Compact Support

Uncertainty of time and angular frequency spread:  $\sigma_t^2 \sigma_\omega^2 \geq \frac{1}{4}$

Gaussian wave function: zero mean,  $\omega = 2\pi f$ ,  $\int_{-\infty}^{+\infty} v_s(t) dt = 1$ :

$$v(t) = \frac{1}{\sqrt{2\pi\sigma^2}} e^{-\frac{t^2}{2\sigma^2}}, \quad \bar{v}(f) = e^{-\frac{\omega^2}{2/\sigma^2}} = e^{-\frac{\omega^2}{2\sigma'^2}}, \quad \sigma^2 \sigma'^2 = 1$$

Variances of location probabilities in time and spectrum:

$$\sigma_t^2 = \frac{1}{W_t} \int_{-\infty}^{+\infty} t^2 v^2(t) dt = \frac{\sigma^2}{2}, \quad \sigma_\omega^2 = \frac{1}{W_\omega} \int_{-\infty}^{+\infty} \omega^2 \bar{v}^2(f) d\omega = \frac{1/\sigma^2}{2}$$

Minimum location uncertainty (Gaussian wave function):  $\sigma_t^2 \sigma_\omega^2 = \frac{1}{4}$

Compact support: If  $v(t) \neq 0 \forall t \in [a, b]$  has compact support, then  $\bar{v}(f)$  cannot have compact support included in  $[-c, +c]$ :

$$v(t) \stackrel{?}{=} \int_{-c}^{+c} df \bar{v}(f) e^{j2\pi ft} = \int_a^b dt' v(t') \underbrace{\int_{-c}^{+c} df e^{j2\pi f(t'-t)}}_{\substack{= \delta(t'-t) \\ c \rightarrow \infty}} \neq v(t)$$



# Short-Time Fourier Transform

**Fourier transform** measures similarity of signal  $v(t)$  with harmonic function localized in frequency  $f'$ :

$$\bar{v}(f') = \langle e^{j2\pi f't} | v(t) \rangle = \int_{-\infty}^{+\infty} v(t) e^{-j2\pi f't} dt$$

**Gabor transform** measures similarity of signal  $v(t)$  with a window  $w(t - t')$  localized at time  $t'$  and frequency  $f'$  ( $\hat{=}$  similarity of windowed signal  $v(t) w(t - t')$  with harmonic function):

$$\begin{aligned} \bar{v}(t', f') &= \langle w(t - t') e^{j2\pi f't} | v(t) \rangle = \int_{-\infty}^{+\infty} v(t) w^*(t - t') e^{-j2\pi f't} dt \\ &= \langle e^{j2\pi f't} | v(t) w^*(t - t') \rangle = \int_{-\infty}^{+\infty} v(t) w^*(t - t') e^{-j2\pi f't} dt \end{aligned}$$

**Spread of Gabor window in time and frequency**  $\sigma_t, \sigma_f$  is fixed and independent of its position in time and frequency  $t' = \bar{t}, f' = \bar{f}$ .

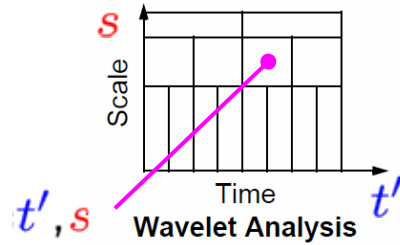
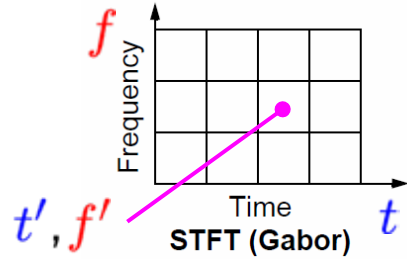
**Further functions** to which a signal may be compared?



# Short-Time Fourier Trafo (STFT) — Contin. Wavelet Trafo (CWT)

Gabor “atom” (1946) is finite-length window  $w(t)$  at  $t = t'$  multiplied with impulse response of a zero-width bandpass at  $f = f'$  (uncertainty rectangles  $\sigma_t \sigma_\omega \geq \frac{1}{2}$ ):

$$w_{t',f'}(t) = w(t - t') e^{j2\pi f' t}, \quad \bar{v}(t', f') = \int_{-\infty}^{+\infty} v(t) w^*(t - t') e^{-j2\pi f' t} dt$$



Wavelet  $\psi(t)$  is function with  $\int_{-\infty}^{+\infty} \psi(t) dt = 0$  ( $\bar{\psi}(0) = 0$ ). Dilated with scale factor  $s$ , translated by  $t' \rightarrow$  wavelet trafo  $\hat{v}(t', s)$  of  $v(t)$ :

$$\psi_{t',s}^s(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t - t'}{s}\right), \quad \hat{v}(t', s) = \int_{-\infty}^{+\infty} v(t) \frac{1}{\sqrt{s}} \psi^*\left(\frac{t - t'}{s}\right) dt$$

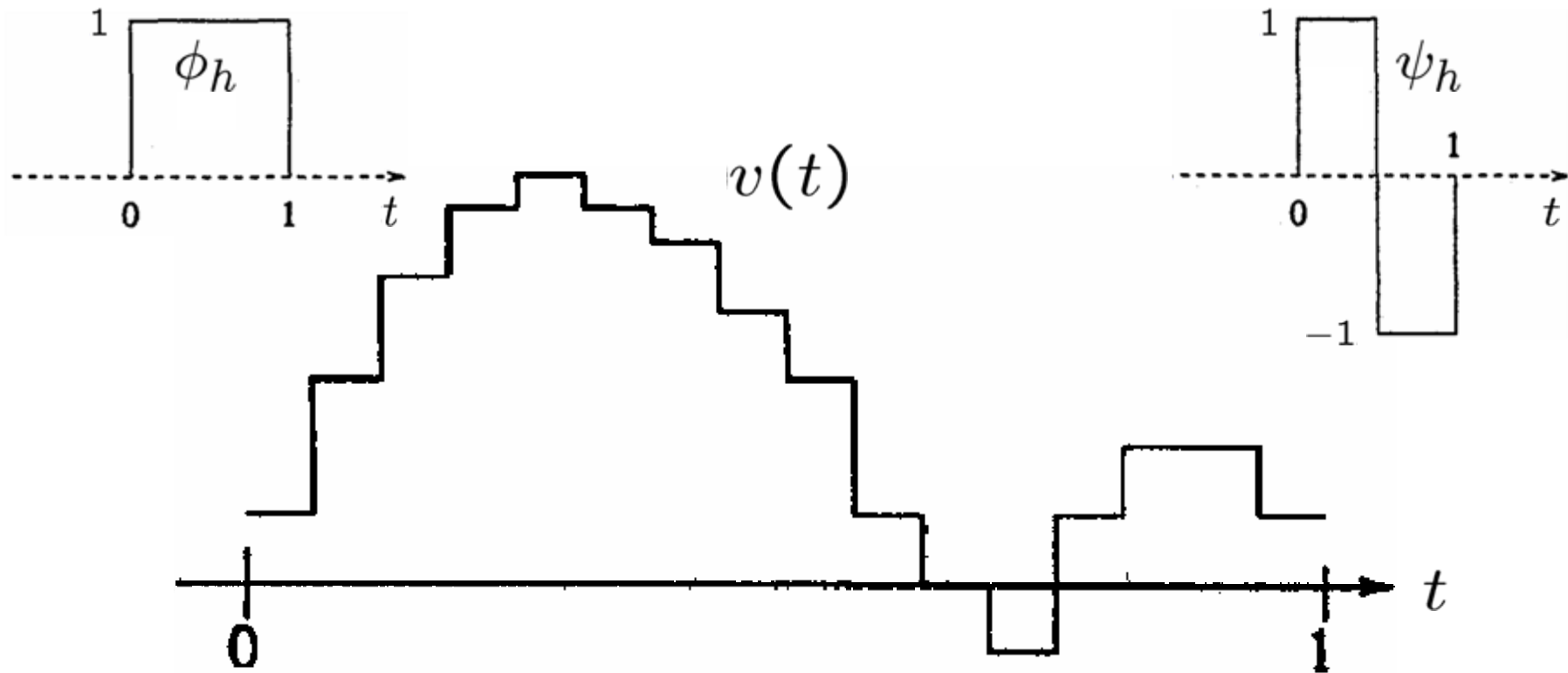


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# The Shape of Wavelets — Haar Wavelet and Scaling Function



Wavelet  $\psi(t)$  **dilated** and **translated**:  $\psi_{t'}^s(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t-t'}{s}\right)$

Not sufficient, DC part is missing:

Scaling function  $\phi(t)$  **dilated** and **translated**:  $\phi_{t'}^s(t) = \frac{1}{\sqrt{s}} \phi\left(\frac{t-t'}{s}\right)$



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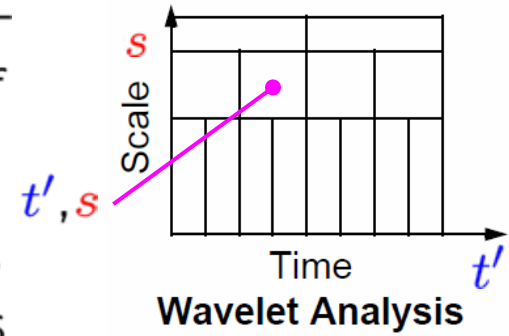
# Continuous Wavelet Transform — “Mother” of Wavelets



**Wavelet**  $\psi(t)$  (“ondelette”, small wave) is waveform of limited duration with average value of zero ( $\hat{=}$  impulse response of “bandpass”).

**Wavelet**  $\psi(t)$  dilated with scale factor  $s \neq 0$ , translated by  $t' \rightarrow$  wavelet transform  $\hat{v}(t', s)$  is correlation of  $v(t)$  with BP impulse response:

$$\psi_{t'}^s(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t - t'}{s}\right), \quad \hat{v}(t', s) = \int_{-\infty}^{+\infty} v(t) \frac{1}{\sqrt{s}} \psi^*\left(\frac{t - t'}{s}\right) dt$$



**Wavelet transform** is complete and energy conserving if  $\psi(t)$  with:

- **Admissibility condition**  $0 < C_\psi := \int_0^\infty \left| \frac{1}{\sqrt{f}} \bar{\psi}(f) \right|^2 df < \infty$
- **Zero mean** (convergence!)  $\int_{-\infty}^{+\infty} \psi(t) dt = 0, \quad \bar{\psi}(0) = 0$
- **Normalization**  $\|\psi(t)\|^2 := \int_{-\infty}^{+\infty} |\psi(t)|^2 dt = 1$



# Continuous Wavelet Transform (CWT) for “Mother” Wavelet $\psi(t)$

**Wavelet transform**  $\hat{v}(t', s)$  of  $v(t)$  is cross-correlation ( $\otimes$ ) with dilated (scaled) impulse response  $\psi^s(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t}{s}\right)$  of a band-pass:

$$\hat{v}(t', s) = \int_{-\infty}^{+\infty} dt v(t) \frac{1}{\sqrt{s}} \psi^*\left(\frac{t-t'}{s}\right) = v(t) \otimes \frac{1}{\sqrt{s}} \psi\left(\frac{t}{s}\right)$$

**Bandpass**, because **Fourier transform**  $\bar{\psi}^s(f)$  of  $\psi^s(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t}{s}\right)$  is

$$\bar{\psi}^s(f) = \int_{-\infty}^{+\infty} \frac{1}{\sqrt{s}} \psi\left(\frac{t}{s}\right) e^{-j2\pi ft} dt = \sqrt{s} \bar{\psi}(sf),$$

and because of  $\int_{-\infty}^{+\infty} \psi(t) dt = 0$  or  $\bar{\psi}(0) = 0$ : The DC component  $f = 0$  of the transfer function disappears  $\rightarrow$  oscillation around time axis  $\rightarrow$  naming “wavelet”,

$$\bar{\psi}^s(0) = \int_{-\infty}^{+\infty} \frac{1}{\sqrt{s}} \psi\left(\frac{t}{s}\right) dt = 0.$$

**CWT**: Correlation with dilated (scaled) impulse responses  $\frac{1}{\sqrt{s}} \psi\left(\frac{t}{s}\right)$  of hypothetical *band-pass* filters  $\sqrt{s} \bar{\psi}(sf)$ .





# Inverse Continuous Wavelet Transform (ICWT) for Mother $\psi(t)$

**Wavelet transform**  $\hat{v}(t', s)$  of  $v(t)$  is cross-correlation ( $\otimes$ ) with dilated (scaled) impulse response  $\psi^s(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t}{s}\right)$  of a band-pass:

$$\hat{v}(t', s) = \int_{-\infty}^{+\infty} dt_1 v(t_1) \frac{1}{\sqrt{s}} \psi^*\left(\frac{t_1 - t'}{s}\right) = v(t_1) \otimes \frac{1}{\sqrt{s}} \psi\left(\frac{t_1}{s}\right)$$

**Inverse wavelet transform**  $v(t)$  of  $\hat{v}(t', s)$  is double integral:

$$v(t) = \frac{1}{C_\psi} \int_0^\infty \frac{1}{s^2} ds \int_{-\infty}^{+\infty} dt' \hat{v}(t', s) \frac{1}{\sqrt{s}} \psi\left(\frac{t - t'}{s}\right),$$

$$0 < C_\psi := \int_0^\infty \left| \frac{1}{\sqrt{f}} \bar{\psi}(f) \right|^2 df = \int_0^\infty \left| \frac{1}{\sqrt{s}} \bar{\psi}(sf) \right|^2 ds < \infty$$

**Energy conservation** (“Parseval”) between time and WT domain:

$$1 = \int_{-\infty}^{+\infty} dt |v(t)|^2 = \frac{1}{C_\psi} \int_0^\infty \frac{ds}{s^2} \int_{-\infty}^{+\infty} dt' |\hat{v}(t', s)|^2 \quad (s \neq 0)$$

**ICWT proof** by substituting  $\hat{v}(t', s)$ ,  $\delta(x-x_0) = \int_{-\infty}^{+\infty} e^{\pm j 2\pi(x-x_0)y} dy$ ,

$$\frac{1}{\sqrt{s}} \psi^*\left(\frac{t_1 - t'}{s}\right) = \int_{-\infty}^{+\infty} df_1 \sqrt{s} \bar{\psi}^*(sf_1) e^{j 2\pi f_1 t'} e^{-j 2\pi f_1 t_1}.$$



# Scaling Function is Aggregation (“Father”) of Wavelets ◀

When CWT  $\hat{g}(t', s)$  is known only for  $s < s_0$ , to recover  $g(t)$  we need complimentary information corresponding to  $\hat{g}(t', s)$  for  $s > s_0$ .

Inverse wavelet transform  $v(t)$  of  $\hat{v}(t', s)$  then splits in **two parts**:

$$v(t) = \underbrace{\frac{1}{C_\psi} \int_0^{s_0} \frac{1}{s^2} ds \int_{-\infty}^{+\infty} dt' \hat{v}(t', s) \frac{1}{\sqrt{s}} \psi\left(\frac{t-t'}{s}\right)}_{v_{BP}(t)} + \underbrace{\frac{1}{C_\psi} \int_{s_0}^{\infty} \frac{1}{s^2} ds (\dots)}_{v_{LP}(t)}$$

Meaning of  $v_{LP}(t)$ :

$$v(t) = v_{BP}(t) + v_{LP}(t) = \frac{1}{C_\psi} \int_{-\infty}^{+\infty} dt_1 v(t_1) \int_{-\infty}^{+\infty} df_1 e^{-j2\pi f_1(t_1-t)} \times \left[ \int_0^{s_0 f_1} \frac{df}{f} |\bar{\psi}(f)|^2 + \underbrace{\int_{s_0 f_1}^{\infty} \frac{df}{f} |\bar{\psi}(f)|^2}_{=: |\sqrt{s_0} \bar{\phi}(s_0 f_1)|^2} \right]$$

Scaling function  $\phi^s(t) = \frac{1}{\sqrt{s}} \phi\left(\frac{t}{s}\right)$  defined as aggregation of wavelets  $\psi^s(t)$  at scales  $s > s_0$ . Its FT has arbitrary phase.



# Meaning of Scaling Function

Computation of  $v_{LP}(t)$ :

$$|\sqrt{s_0} \bar{\phi}(s_0 f_1)|^2 := \int_{s_0 f_1}^{\infty} \left| \frac{1}{\sqrt{f}} \bar{\psi}(f) \right|^2 df = \int_{s_0}^{\infty} \left| \frac{1}{\sqrt{s}} \bar{\psi}(s f_1) \right|^2 ds,$$

$$v_{LP}(t) = \frac{1}{C_\psi} \int_{-\infty}^{+\infty} dt_1 v(t_1) \int_{-\infty}^{+\infty} df_1 e^{-j2\pi f_1(t_1-t)} \int_{s_0 f_1}^{\infty} \frac{df}{f} |\bar{\psi}(f)|^2$$

Substituting scaling function  $\frac{1}{\sqrt{s_0}} \phi\left(\frac{t}{s_0}\right)$  for last integral:

$$v_{LP}(t) = \frac{1}{C_\psi} \int_{-\infty}^{+\infty} dt_1 v(t_1) \int_{-\infty}^{+\infty} df_1 e^{-j2\pi f_1(t_1-t)} |\sqrt{s_0} \bar{\phi}(s_0 f_1)|^2,$$

$$\sqrt{s_0} \bar{\phi}(s_0 f_1) = \int_{-\infty}^{+\infty} dt_i \frac{1}{\sqrt{s_0}} \phi\left(\frac{t_i}{s_0}\right) e^{-j2\pi f_1 t_i}, \quad i = 2, 3$$

Substituting  $\int_{-\infty}^{+\infty} df_1 e^{-j2\pi f_1(t_1-t-t_2+t_3)} = \delta(t_1 - t - t_2 + t_3)$ , integrating over  $t_3$  and substituting  $t' = t_1 - t_2$ :

$$v_{LP}(t) = \frac{1}{C_\psi} \int_{-\infty}^{+\infty} dt' \underbrace{\int_{-\infty}^{+\infty} dt_1 v(t_1) \frac{1}{\sqrt{s_0}} \phi^*\left(\frac{t_1 - t'}{s_0}\right)}_{=: \bar{v}(t', s_0)} \frac{1}{\sqrt{s_0}} \phi\left(\frac{t - t'}{s_0}\right)$$



# Scaling Function as Low-Pass Impulse Response

From admissibility condition  $0 < C_\psi := \int_0^\infty \left| \frac{1}{\sqrt{f}} \bar{\psi}(f) \right|^2 df < \infty$ ,

$$\lim_{f_1 \rightarrow 0} \left| \sqrt{s_0} \bar{\phi}(s_0 f_1) \right|^2 = \lim_{f_1 \rightarrow 0} \int_{s_0 f_1}^\infty \left| \frac{1}{\sqrt{f}} \bar{\psi}(f) \right|^2 df = C_\psi \quad \rightsquigarrow$$

Scaling function is impulse response of *low-pass* filter,  $|\bar{\phi}(0)|^2 \neq 0$ .

Normalization of scaling function:

$$\int_{-\infty}^{+\infty} \phi(t) dt = 1 \quad \left( \text{wavelet: } \int_{-\infty}^{+\infty} \psi(t) dt = 0 \right)$$

Scaling function transform  $\bar{v}(t', s)$  is cross-correlation with dilated (scaled) impulse response  $\frac{1}{\sqrt{s_0}} \phi\left(\frac{t}{s_0}\right)$  of a hypothetical *low-pass* filter  $\sqrt{s_0} \bar{\phi}(s_0 f)$ :

$$\bar{v}(t', s_0) = \int_{-\infty}^{+\infty} v(t_1) \frac{1}{\sqrt{s_0}} \phi^*\left(\frac{t_1 - t'}{s_0}\right) dt_1 = v(t_1) \otimes \frac{1}{\sqrt{s_0}} \phi\left(\frac{t_1}{s_0}\right)$$



# Wavelet and Scaling Function Transform



$v(t) = v_{\text{BP}}(t) + v_{\text{LP}}(t)$ , BP and LP parts take form of a convolution:

$$v_{\text{BP}}(t) = \frac{1}{C_\psi} \int_0^{s_0} \frac{ds}{s^2} \underbrace{\int_{-\infty}^{+\infty} dt' \hat{v}(t', s)}_{\hat{v}(t', s) * \frac{1}{\sqrt{s}} \psi\left(\frac{t'}{s}\right)} \frac{1}{\sqrt{s}} \psi\left(\frac{t-t'}{s}\right)$$

$$v_{\text{LP}}(t) = \frac{1}{C_\psi} \underbrace{\int_{-\infty}^{+\infty} dt' \bar{v}(t', s_0)}_{\bar{v}(t', s_0) * \frac{1}{\sqrt{s_0}} \phi\left(\frac{t'}{s_0}\right)} \frac{1}{\sqrt{s_0}} \phi\left(\frac{t-t'}{s_0}\right)$$

Wavelet and scaling function transform are cross-correlations ( $\otimes$ ) with dilated (scaled) impulse responses of BP and LP filters:

$$\hat{v}(t', s) = \int_{-\infty}^{+\infty} v(t_1) \frac{1}{\sqrt{s}} \psi^*\left(\frac{t_1-t'}{s}\right) dt_1 = v(t_1) \otimes \frac{1}{\sqrt{s}} \psi\left(\frac{t_1}{s}\right)$$

$$\bar{v}(t', s) = \int_{-\infty}^{+\infty} v(t_1) \frac{1}{\sqrt{s}} \phi^*\left(\frac{t_1-t'}{s}\right) dt_1 = v(t_1) \otimes \frac{1}{\sqrt{s}} \phi\left(\frac{t_1}{s}\right)$$



# Moments of a Wavelet

2D CWT  $\hat{v}(t', s)$  of 1D function  $v(t)$ , 4D transform for 2D signal.

Desirable: Fast decay of  $\hat{v}(t', s)$  wrt to small scales  $s$  (fine details)

Moments of a wavelet (from admissibility condition  $\rightarrow M_0 = 0$ ):

$$M_p = \int_{-\infty}^{+\infty} t^p \psi(t) dt$$

Taylor expansion of  $v(t)$  for  $t' = 0$ ,  $p$ th derivative  $v^{(p)}(t)$ :

$$\hat{v}(t', s) = \int_{-\infty}^{+\infty} v(t_1) \frac{1}{\sqrt{s}} \psi^*\left(\frac{t_1 - t'}{s}\right) dt_1 = \int_{-\infty}^{+\infty} v(t_2 + t') \frac{1}{\sqrt{s}} \psi^*\left(\frac{t_2}{s}\right) dt_2,$$

$$\hat{v}(0, s) = \sum_{p'=0}^p \frac{1}{\sqrt{s}} \frac{v^{(p')}(0)}{p'!} \int_{-\infty}^{+\infty} (s t)^{p'} \psi^*(t) s dt$$

$$= \frac{1}{\sqrt{s}} \left[ \sum_{p'=0}^p \frac{v^{(p')}(0)}{p'!} M_{p'}^* s^{p'+1} + \mathcal{O}(s^{p+2}) \right]$$

Smallness of CWT:  $M_{p' \leq p} = 0$  (or  $\lll$ )  $\rightsquigarrow \hat{v}(t', s) \propto s^{p+2}$ , approx. or-

der  $p \rightsquigarrow$  polyn.  $v(t) = \sum_{p'=0}^p v_{p'} t^{p'}$  has  $\hat{v}(t', s) = 0$ ,  $\psi$ -support  $\geq 2p - 1$ .

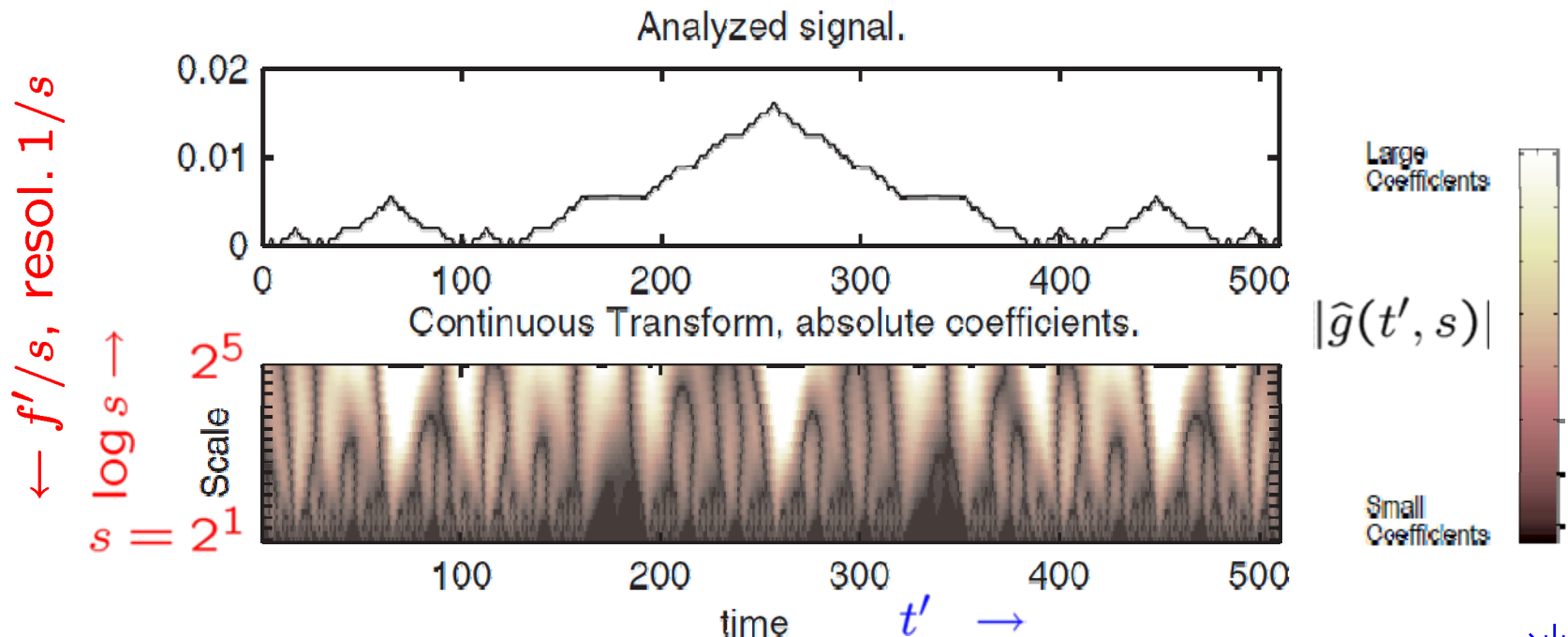


# CWT Example

Wavelet  $\frac{1}{\sqrt{s}} \psi\left(\frac{t-t'}{s}\right)$  is localized near  $t = t'$ . The FT  $\bar{\psi}(f)$  of  $\psi(t)$  be localized near  $f = f'$ . Because of

$$\sqrt{s} \bar{\psi}(sf) = \int_{-\infty}^{+\infty} \frac{1}{\sqrt{s}} \psi\left(\frac{t}{s}\right) e^{-j2\pi ft} dt,$$

spectrum of wavelet  $\sqrt{s} \bar{\psi}(sf)$  is localized near  $f = f'/s$ .



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# Discrete Wavelet Transform (DWT)

Wavelet representation of  $v(t)$  is highly redundant (2D integral):

$$v(t) = \frac{1}{C_\psi} \int_0^\infty \frac{ds}{s^2} \int_{-\infty}^{+\infty} dt' \hat{v}(t', s) \frac{1}{\sqrt{s}} \psi\left(\frac{t-t'}{s}\right)$$

Admissibility:  $0 < C_\psi := \int_0^\infty \left| \frac{1}{\sqrt{f}} \tilde{\psi}(f) \right|^2 df < \infty$

Energy conservation (“Parseval”) between time and WT domain:

$$1 = \int_{-\infty}^{+\infty} dt |v(t)|^2 = \frac{1}{C_\psi} \int_0^\infty \frac{ds}{s^2} \int_{-\infty}^{+\infty} dt' |\hat{v}(t', s)|^2 \quad (s \neq 0)$$

Conjecture: Double integral can be replaced by double sum, usually “dyadic” wavelets (scale  $s = 2^j$ , resolution  $2^{-j}$ ):

$$\psi_k^j(t) := 2^{-j/2} \psi\left(2^{-j}t - k\Delta t\right) \quad \text{for } s = 2^j, \quad t' = 2^j k \Delta t, \quad (j, k) \in \mathbb{Z}$$

Discrete wavelet representation of  $v(t)$  with orthonormal wavelets  $\int_{-\infty}^{+\infty} dt \psi_k^j(t) \psi_{k'}^{j'*}(t) = \delta_{j,j'} \delta_{k,k'}$  (no redundancy):

$$v(t) = \sum_{j,k} \hat{v}_k^j \psi_k^j(t), \quad \hat{v}_k^j = \int_{-\infty}^{+\infty} dt v(t) \psi_k^{j'*}(t)$$



# Discrete Wavelet Transform (DWT) Removes Redundancy

“Dyadic” orthonormal wavelets ( $s = 2^j$ ,  $t' = 2^j k \Delta t$ ):

$$\psi_k^j(t) := 2^{-j/2} \psi(2^{-j} t - k \Delta t), \quad \int_{-\infty}^{+\infty} dt \psi_k^j(t) \psi_{k'}^{j'*}(t) = \delta_{j,j'} \delta_{k,k'}$$

Wavelet coefficients  $\hat{v}_k^j$  in expansion  $v(t) = \sum_{j,k} \hat{v}_k^j \psi_k^j(t) \quad \forall (j, k) \in \mathbb{Z}$ :

$$\int_{-\infty}^{+\infty} dt v(t) \psi_{k'}^{j'*}(t) = \sum_{j,k} \hat{v}_k^j \underbrace{\int_{-\infty}^{+\infty} dt \psi_k^j(t) \psi_{k'}^{j'*}(t)}_{= \delta_{j,j'} \delta_{k,k'}} = \hat{v}_{k'}^{j'} \quad \blacksquare$$

Wavelet coefficient  $\hat{v}_k^j$  gives correlation of signal  $v(t)$  with wavelet  $\psi_k^j(t)$  shifted to  $t' = 2^{-j/2} k \Delta t$ , i. e., with impulse response of constant- $Q$  band-pass filters (width  $2^j \Delta f_j = \Delta f_c$  near  $2^j f_j = f_c$ ):

$$2^{j/2} \bar{\psi}(2^j f) \quad \bullet \text{---} \circ \quad 2^{-j/2} \psi(2^{-j} t), \quad Q_j = \frac{f_j}{\Delta f_j} = \frac{2^j f_c}{2^j \Delta f_c} = \frac{f_c}{\Delta f_c}$$

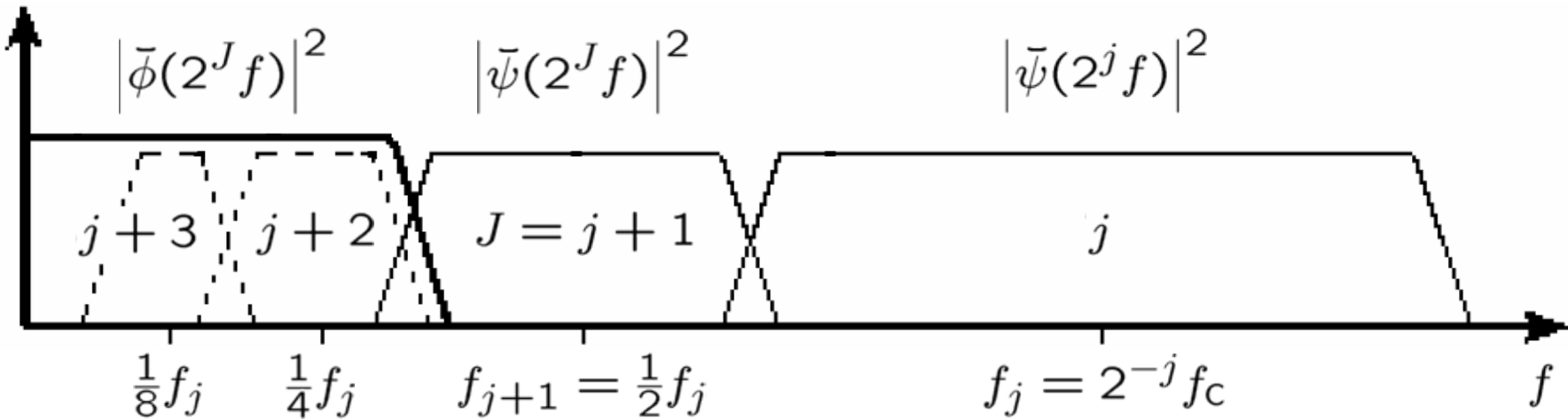
If the wavelet is a matched filter to  $\bar{v}(f)$ :

$$\bar{\psi}_k^{j*}(f) = \bar{v}(f) \rightarrow \psi_k^{j*}(t) = v(t) \rightarrow \hat{v}_k^j = \int_{-\infty}^{+\infty} dt |v(t)|^2 = 1$$



# Replacing an Infinite Set of Wavelets by one Scaling Function

$$\phi_k^j(t) := 2^{-j/2} \phi(2^{-j} t - k\Delta t), \quad \psi_k^j(t) := 2^{-j/2} \psi(2^{-j} t - k\Delta t)$$



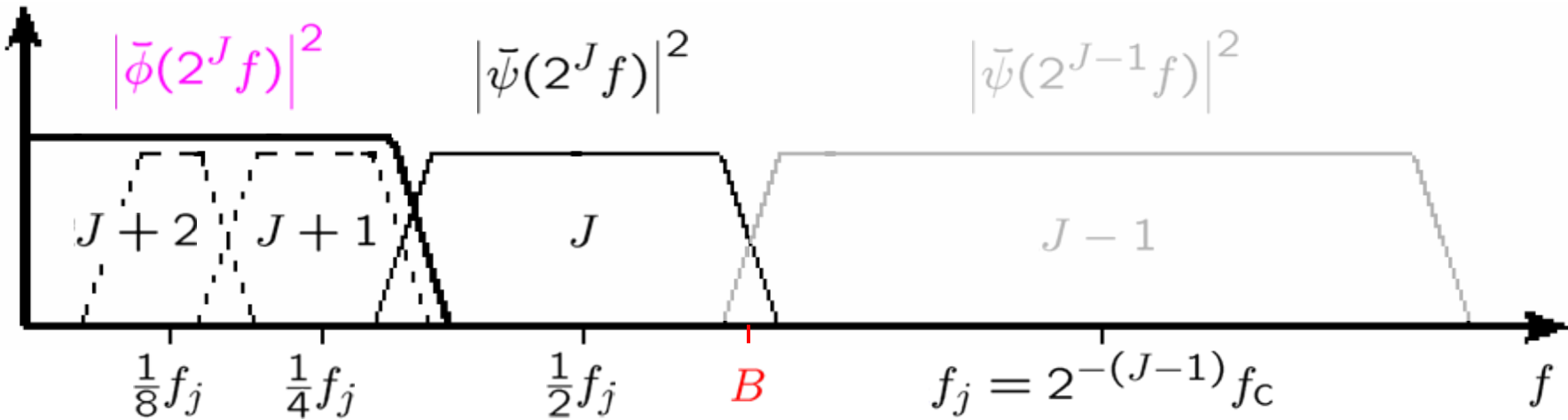
Expansion of non-bandlimited signal  $\hat{v}(t)$  ( $v_k^j \neq 0 \forall j < J$ ):

$$v(t) = \underbrace{\sum_{j=-\infty}^J \underbrace{\sum_{k=-\infty}^{+\infty} \hat{v}_k^j \psi_k^j(t)}_{\text{BP detail } D^j(t)}}_{\text{all BP details up to } D^J(t)} + \underbrace{\sum_{k=-\infty}^{+\infty} \underbrace{\sum_{j=J+1}^{+\infty} \hat{v}_k^j \psi_k^j(t)}_{\bar{v}_k^J \phi_k^J(t)}}_{\text{LP approximation } A^J(t)}$$



Bandlimited Signal  $v(t)$ ,  $\bar{v}(f > B) = 0$

$$\phi_k^j(t) := 2^{-j/2} \phi(2^{-j} t - k\Delta t), \quad \psi_k^j(t) := 2^{-j/2} \psi(2^{-j} t - k\Delta t)$$



Expansion of bandlimited signal  $v(t)$  ( $B \hat{=} J$ ):

$$\begin{aligned} v(t) &= \sum_{k=-\infty}^{+\infty} \hat{v}_k^J \psi_k^J(t) + \underbrace{\sum_{k=-\infty}^{+\infty} \bar{v}_k^J \phi_k^J(t)}_{\text{skinny } \phi^J} = \sum_{k=-\infty}^{+\infty} \bar{v}_k^{J-1} \phi_k^{J-1}(t) \\ &= \sum_{k=-\infty}^{+\infty} \hat{v}_k^J \psi_k^J(t) + \sum_{k=-\infty}^{+\infty} \hat{v}_k^{J+1} \psi_k^{J+1}(t) + \sum_{k=-\infty}^{+\infty} \bar{v}_k^{J+1} \phi_k^{J+1}(t) \end{aligned}$$

skinny  $\phi^J = \text{fat } \psi^{J+1} + \text{fat } \phi^{J+1}$ , skinny  $\phi^{J-1} = \text{fat } \psi^J + \text{fat } \phi^J$



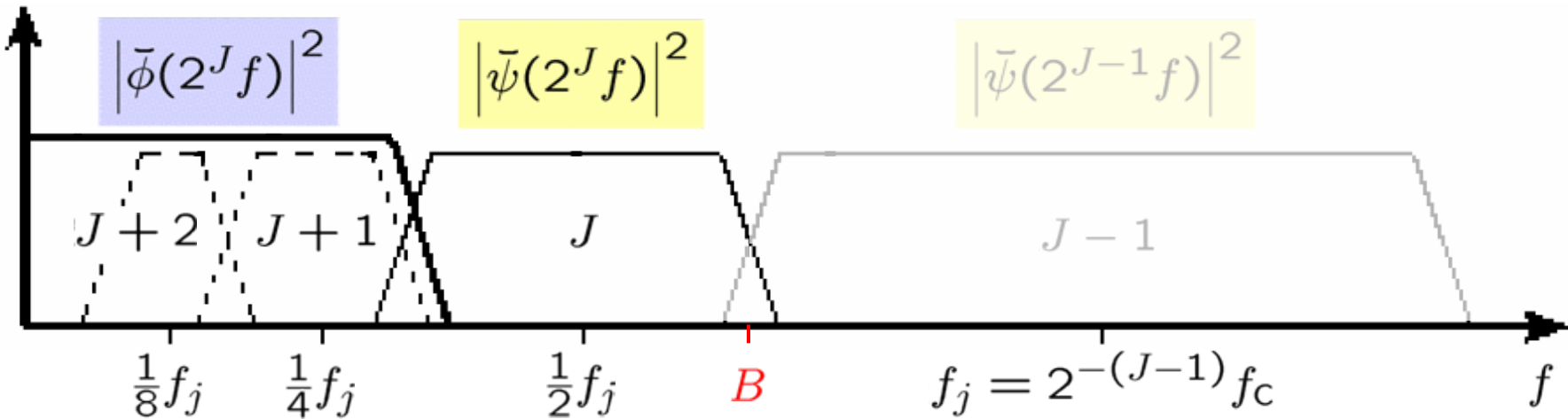
# Outline

- **Wavelets**
  - What are they good for? **Multiresolution analysis**
- Finite-differences in time-domain
  - Yee's leapfrog algorithm
  - Numerical dispersion, stability and accuracy
  - Higher-order finite-differences
  - Method of weighted residuals: Collocation
- Wavelet FDTD
  - Numerical dispersion, stability and accuracy
  - What should be further done?
  - Dispersive and nonlinear media
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- Summary and further reading



# Multiresolution Analysis (MRA)

$$\phi_k^j(t) := 2^{-j/2} \phi(2^{-j} t - k\Delta t), \quad \psi_k^j(t) := 2^{-j/2} \psi(2^{-j} t - k\Delta t)$$



Signal space  $V_{J-1}$  has orthogonal subspaces,  $V_{J-1} = V_J \oplus W_J$ , etc.

“Average” (LP)  $V_J$  spanned by orthonormal scaling fct base  $\phi_k^J(t)$

“Detail” (BP)  $W_J$  spanned by orthonormal wavelet base  $\psi_k^J(t)$

Refinement (scaling) relations given by discrete “filters”  $h_k, g_k$ :

$$\phi_k^j(t) = \sum_{k'=-\infty}^{+\infty} h_{k'} \phi_{2k+k'}^{j-1}(t), \quad \psi_k^j(t) = \sum_{k'=-\infty}^{+\infty} g_{k'} \phi_{2k+k'}^{j-1}(t)$$



# Multiresolution Analysis (MRA)

Expansion of bandlimited signal  $v(t)$  ( $B \hat{=} J$ ):

$$v(t) = A^J(t) + D^J(t) = \sum_{k=-\infty}^{+\infty} a_k^J \phi_k^J(t) + \sum_{k=-\infty}^{+\infty} d_k^J \psi_k^J(t)$$

Usual notation of averages (approximations)  $a_k^j$  and details  $d_k^j$ :

$$a_k^j := \bar{v}_k^j = \int_{-\infty}^{+\infty} dt \phi_k^{j*}(t) v(t), \quad d_k^j := \hat{v}_k^j = \int_{-\infty}^{+\infty} dt \psi_k^{j*}(t) v(t)$$

Recursion for averages and details using refinement relations:

$$\phi_k^j(x) = \sum_{k'} h_{k'} \phi_{2k+k'}^{j-1}(t) \quad \psi_k^j(x) = \sum_{k'} g_{k'} \phi_{2k+k'}^{j-1}(t)$$

$$a_k^j = \sum_{k'} h_{k'}^* a_{2k+k'}^{j-1}$$

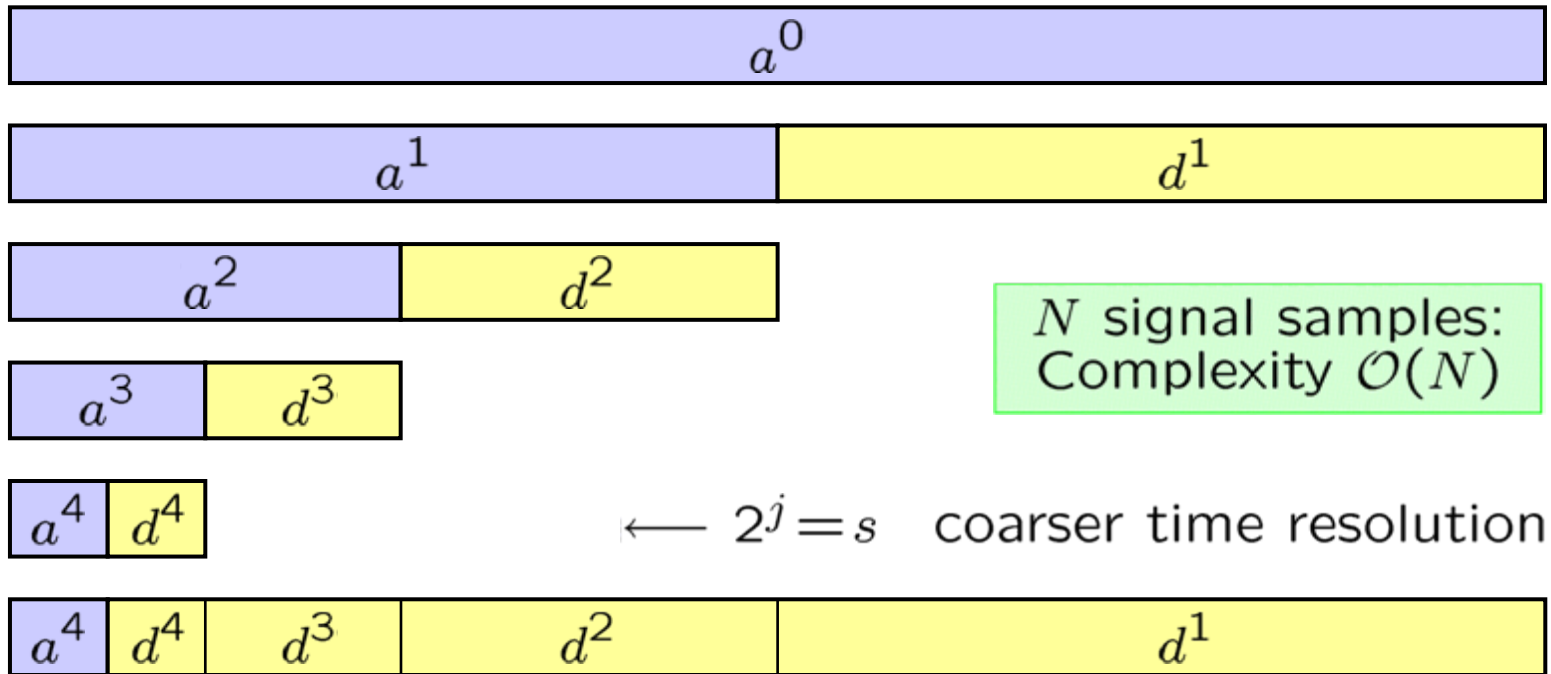
$$d_k^j = \sum_{k'} g_{k'}^* a_{2k+k'}^{j-1}$$

Filtering  $A^J, D^J$  with complex conjugate filters  $h_k^*, g_k^*$  leads to  $A^{J+1}, D^{J+1}$  at larger scale  $J+1$ . Finite-length signals  $k \in [1, K]$ :

“Filter bank iteration” with scaling / wavelet filters leaves number of coefficients constant (decimation factor 2 in subscript  $2k+k'$ ).



# FWT and Inverse FWT (IFWT, Reconstruction)



Reconstruction of bandlimited signal  $v(t)$  ( $B \cong J=1$ ):

$$\begin{aligned}
 v(t) &= \sum_{k=-\infty}^{+\infty} a_k^{J-1} \phi_k^{J-1}(t) = \sum_{k=-\infty}^{+\infty} a_k^J \phi_k^J(t) + \sum_{k=-\infty}^{+\infty} d_k^J \psi_k^J(t) \\
 &= \sum_{k=-\infty}^{+\infty} a_k^{J+3} \phi_k^{J+3}(t) + \sum_{j=J}^{J+3} \sum_{k=-\infty}^{+\infty} d_k^j \psi_k^j(t)
 \end{aligned}$$



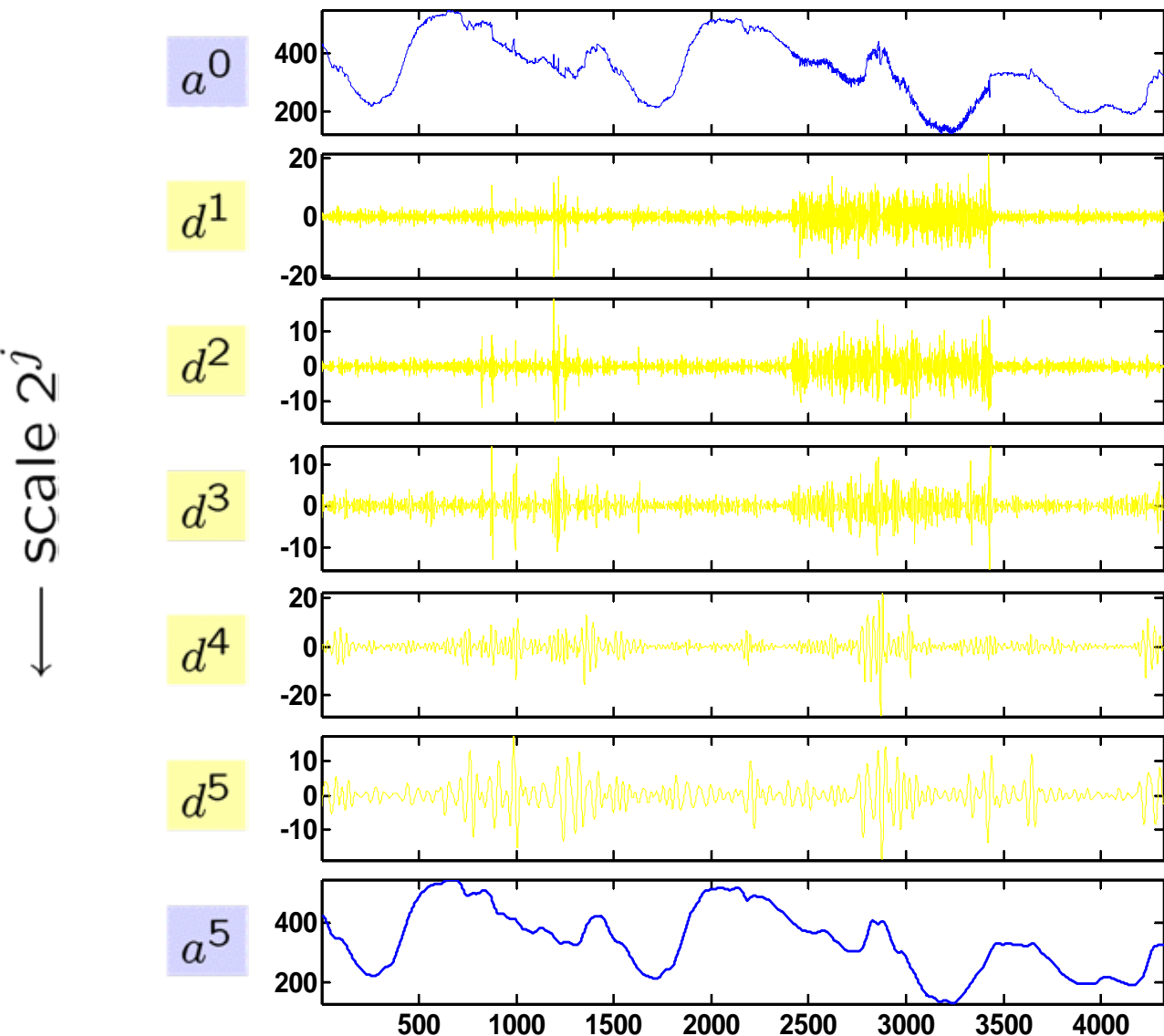


# Outline

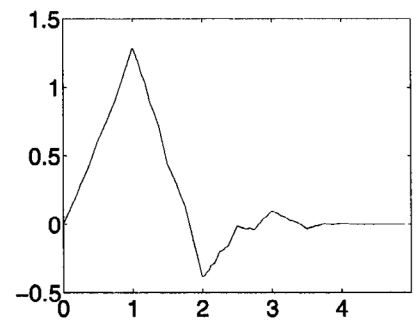
- **Wavelets**
  - What are they good for? **Daubechies wavelets**
- Finite-differences in time-domain
  - Yee's leapfrog algorithm
  - Numerical dispersion, stability and accuracy
  - Higher-order finite-differences
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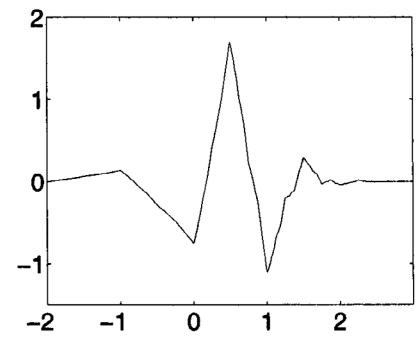
# MRA with Daubechies Wavelet D3 ( $D_p$ or $db_p$ , $p$ moments vanish)



support  $[0, 2p - 1]$   
 $\phi(t)$



$\psi(t)$



support  $[-(p - 1), p]$

D3 (db3)



# Daubechies Wavelets $D_p$ — Cascade Algorithm

Recursion for  $\phi^{Dp}$  with known filter  $h_k^{Dp}$ . Start with Haar scaling function  $\phi_h \hat{=} \phi^{D1} = \phi^{(0)}$ , iterate refinement equation:

$$\phi^{(i+1)}(t) = 2^{1/2} \sum_{k'=0}^{p-1} h_{k'}^{Dp} \phi^{(i)}(2t - k' \Delta t), \quad i = 0, 1, 2, \dots,$$

$$\psi^{Dp}(t) = 2^{1/2} \sum_{k'=-1}^1 g_{k'}^{Dp} \phi^{Dp}(2t - k' \Delta t), \quad g_k^{Dp} = (-1)^{1-k} h_{1-k}^{Dp}$$

$p$  wavelet moments vanish, good convergence for details if  $j \rightarrow -\infty$ :

$$M_p = \int_{-\infty}^{+\infty} t^p \psi^{Dp}(t) dt \quad \rightsquigarrow \quad d_k^j := \hat{v}_k^j \propto 2^{(p+2)j}$$

- Wavelet suppresses polynomial parts in  $v(t)$  up to degree  $p$ .
- Support for  $\psi^{Dp}, \psi^{Dp}$  is  $2p - 1$ , i. e.,
- filter coefficients are  $h_0^{Dp}, h_1^{Dp}, \dots, h_{2p-1}^{Dp}$ .
- Wavelets and scaling functions orthogonal in all combinations

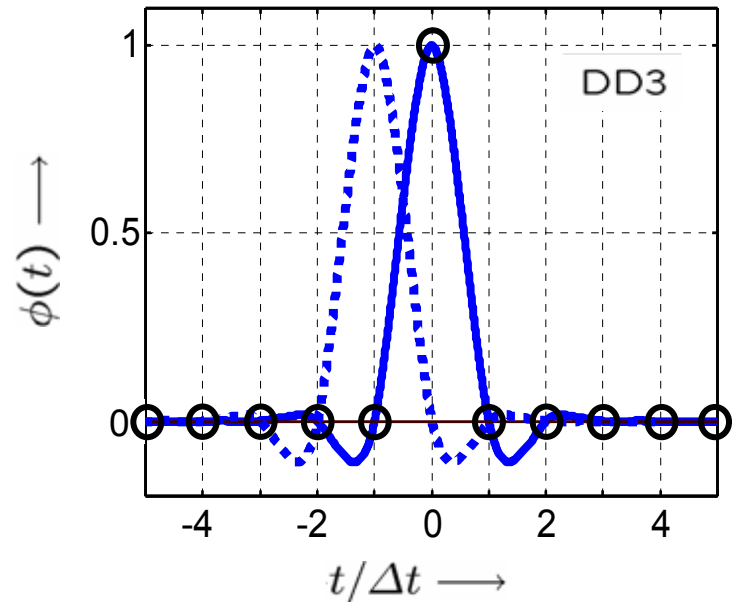
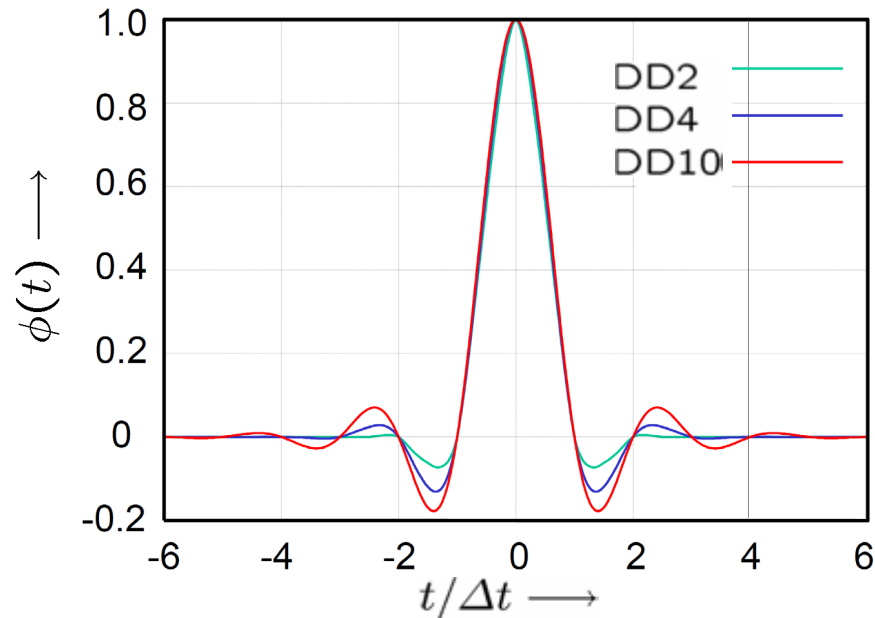


# Outline

- **Wavelets**
  - What are they good for? [Deslauriers-Dubuc](#)
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# Deslauriers-Dubuc Interpolating Scaling Function



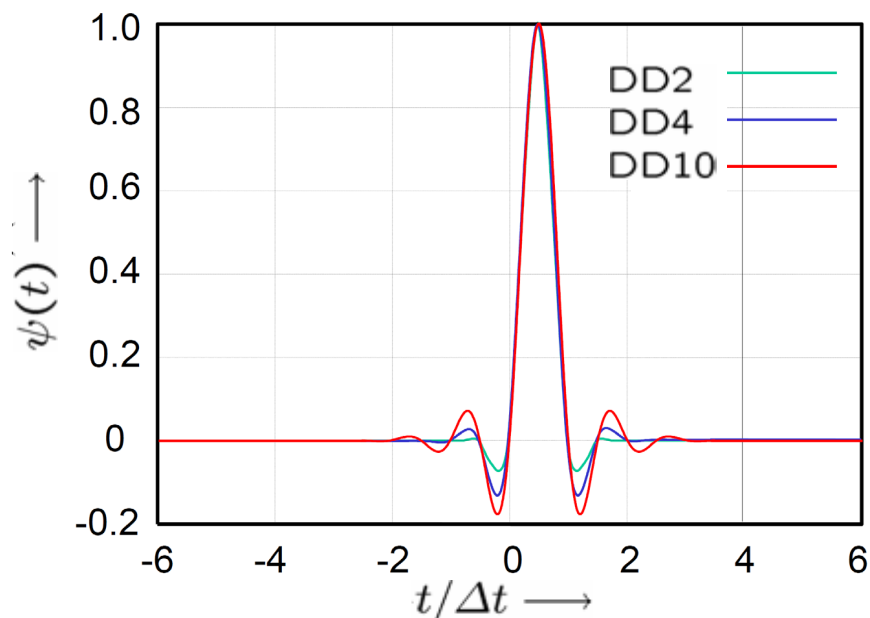
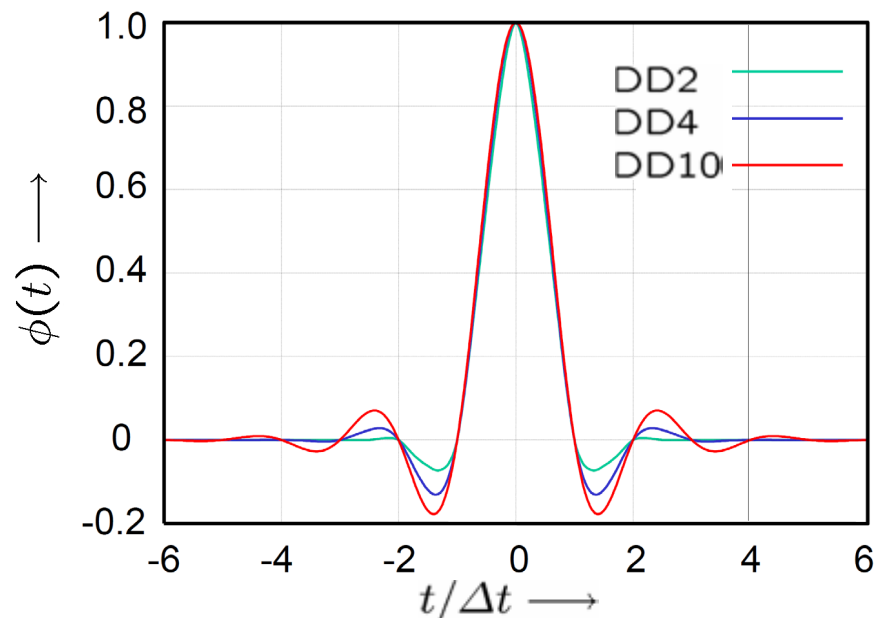
Interpolating function  $\phi(t)$  recovers  $v(t)$  from sampling  $\{v(k\Delta t)\}_{k \in \mathbb{Z}}$ .

Deslauriers-Dubuc  $\phi^{\text{DD}p}(t)$ : has degree  $L = 2p - 1$ , compact support  $[-L, L]$ , decomposes polynomial of degree  $L$ , is derived from (real) Daubechies scaling functions  $\phi^{\text{D}p}(t)$  by correlation:

$$\phi^{\text{DD}p}(t) = \int_{-\infty}^{+\infty} dt' \phi^{\text{D}p}(t') \phi^{\text{D}p}(t' - t), \quad v(t) = \sum_{k=-\infty}^{+\infty} v(k\Delta t) \phi^{\text{DD}p}(t - k\Delta t)$$



# Deslauriers-Dubuc Interpolating Wavelets



Refinement (scaling) relations given by discrete “filters”  $h_k^{\text{DD}p}$ :

$$\phi(t) = \sum_{k=-\infty}^{+\infty} h_k^{\text{DD}p} \phi(2t - k\Delta t), \quad \psi(t) = \phi(2t - \Delta t)$$

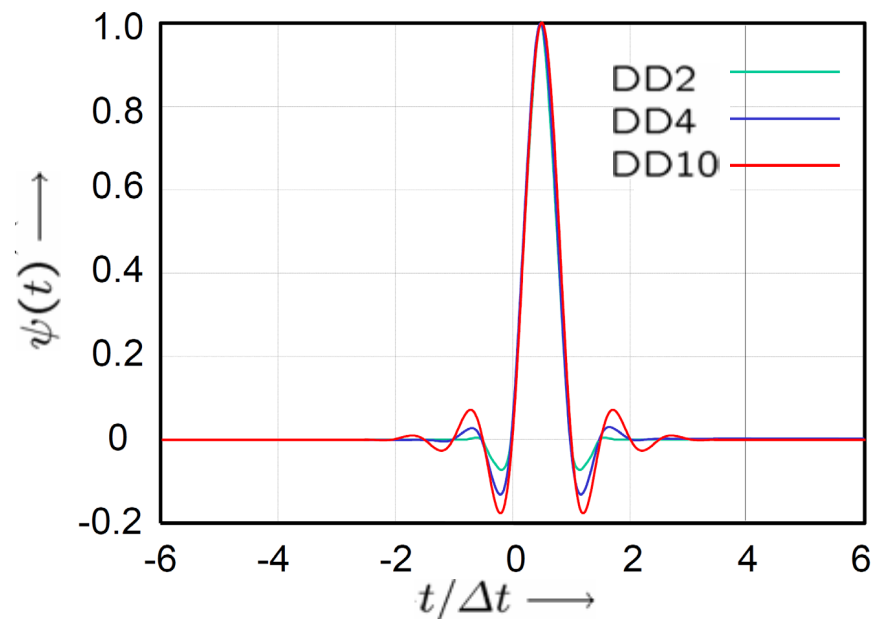
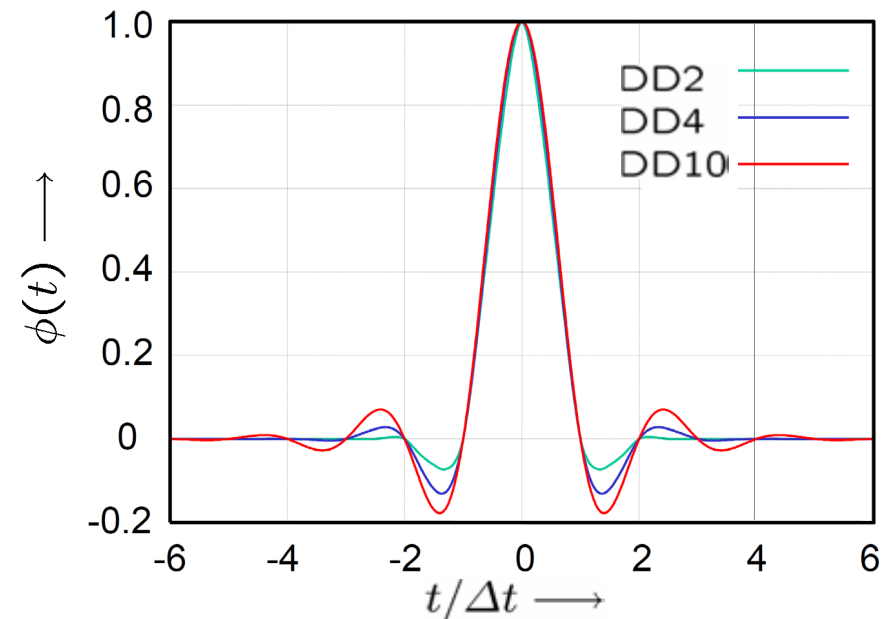
no true wavelet,  
no vanishing  
moments

Deslauriers-Dubuc filter by correlating Daubechies wavelet  $h_k^{\text{D}p}$ :

$$h_k^{\text{DD}p} = \sum_{k'=-\infty}^{+\infty} h_{k'}^{\text{D}p} h_{k'-k}^{\text{D}p}, \quad \text{symmetry: } h_{-k}^{\text{DD}p} = h_k^{\text{DD}p}$$



# Deslauriers-Dubuc Wavelets and Duals



Orthogonality:

$$\int_{-\infty}^{+\infty} dt \phi_k^j(t) \tilde{\phi}_{k'}^{j'}(t) = \delta_{j,j'} \delta_{k,k'}, \quad \int_{-\infty}^{+\infty} dt \psi_k^j(t) \tilde{\psi}_{k'}^{j'}(t) = \delta_{j,j'} \delta_{k,k'}$$

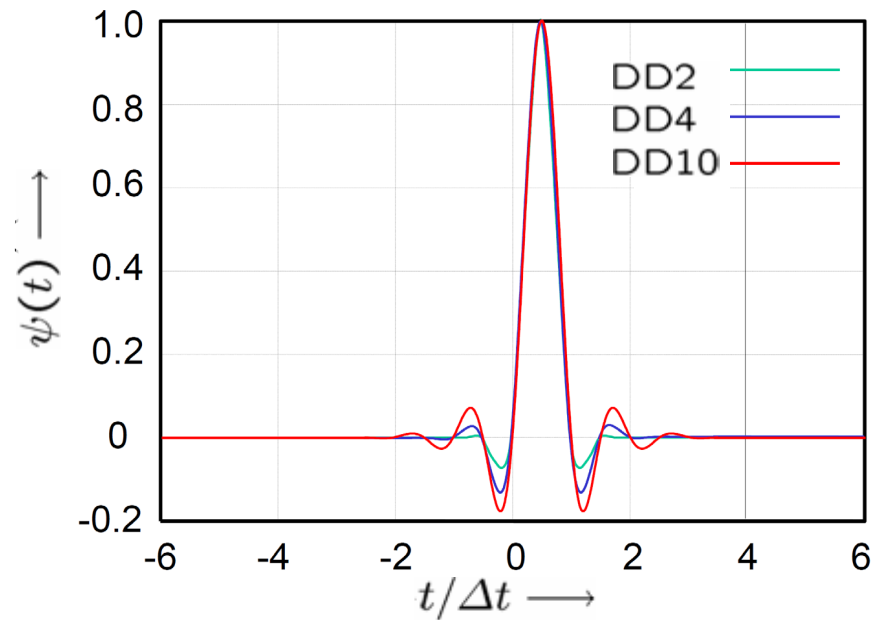
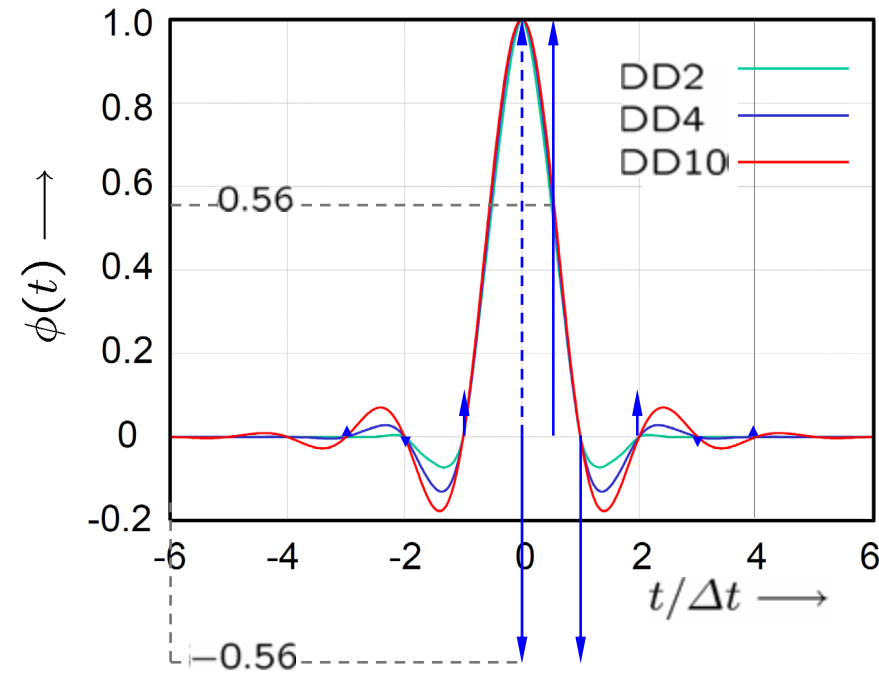
**So far:** Analyzing fct is  $(\cdot)^*$  of synthesizing fct,  $\tilde{\phi}_k^j = \phi_k^{j*}$ ,  $\tilde{\psi}_k^j = \psi_k^{j*}$

DD scaling function and “wavelet” are biorthogonal:

$$\tilde{\phi}(t) = \Delta t \delta(t), \quad \tilde{\psi}(t) = \Delta t \sum_{k'=-2p+2}^{2p} (-1)^{k'-1} h_{-k'+1}^{\text{DD}p} \delta(t - \frac{1}{2}k' \Delta t)$$



# Deslauriers-Dubuc Wavelets — Biorthogonality



DD scaling function and wavelet are **bi**orthogonal:

$$\underline{\tilde{\phi}(t) = \Delta t \delta(t)}, \quad \underline{\tilde{\psi}(t) = \Delta t \sum_{k'=-2p+2}^{2p} (-1)^{k'-1} h_{-k'+1}^{\text{DD}p} \delta(t - \frac{1}{2}k' \Delta t)}$$

$$\int_{-\infty}^{+\infty} dt \phi_k^j(t) \tilde{\phi}_{k'}^{j'}(t) = \delta_{j,j'} \delta_{k,k'}, \quad \int_{-\infty}^{+\infty} dt \psi_k^j(t) \tilde{\psi}_{k'}^{j'}(t) = \delta_{j,j'} \delta_{k,k'},$$

$$\int_{-\infty}^{+\infty} dt \phi_k^j(t) \tilde{\psi}_{k'}^{j'}(t) = 0, \quad \int_{-\infty}^{+\infty} dt \psi_k^j(t) \tilde{\phi}_{k'}^{j'}(t) = 0$$





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# Maxwell's Fundamental Equations

Ampere's law:  $\text{curl } \vec{H} = \frac{\partial \vec{D}}{\partial t}$ , Faraday's law:  $\text{curl } \vec{E} = -\frac{\partial \vec{B}}{\partial t}$

Gauss' law:  $\text{div } \vec{D} = 0$ ,  $\text{div } \vec{B} = 0$

Constitutive:  $\vec{D} = \epsilon_0 \vec{E} + \vec{P}$ ,  $\vec{B} = \mu_0 \vec{H} + \vec{M}$

Electromagnetic waves are described by:

- magnetic and electric field vectors  $\vec{H}$  and  $\vec{E}$ ,
- electric displacement  $\vec{D}$ , electric material polarization  $\vec{P}$ ,
- magnetic induction  $\vec{B}$ , magnetic material polarization  $\vec{M}$

At operating frequencies  $f$ , the medium has

- no space charges/currents, is isotropic, linear,
- has a dielectric constant  $\epsilon = \epsilon_0 \epsilon_r$ , polarisation  $\vec{P} = \epsilon_0 (\epsilon_r - 1) \vec{E}$ ,
- and a magnetic permeability  $\mu = \mu_0 \mu_r$ , pol.  $\vec{M} = \mu_0 (\mu_r - 1) \vec{H}$ .
- The vacuum speed of light is  $c = 1 / \sqrt{\epsilon_0 \mu_0}$ , and the
- medium phase velocity is  $v = c/n$  (refractive index  $n = \pm \sqrt{\mu_r \epsilon_r}$ ).



# Sampling, Nyquist Frequency and Interpolation

Sampling function  $a(x)$  is self-reciprocal in Fourier space  $\tilde{a}(\xi)$ :

$$a(x) = X_a \sum_{n=-\infty}^{+\infty} \delta(x - nX_a) = \sum_{n=-\infty}^{+\infty} \exp\left(-j2\pi n \frac{x}{X_a}\right),$$
$$\tilde{a}(\xi) = \sum_{n=-\infty}^{+\infty} \delta(\xi - n\Xi_a), \quad \Xi_a = 1/X_a$$

Function  $\Psi(x)$  assumed to have bandlimited spatial spectrum:

$$\tilde{\Psi}(\xi) = \int_{-\infty}^{+\infty} \Psi(x) \exp(j2\pi \xi x) dx, \quad \tilde{\Psi}(|\xi| > B_x) = 0$$

$\Psi(x)$  sampled with Nyquist frequency  $B_x$  at intervals  $X_a = 1/B_x$ , discrete complex  $\Psi_a(x) = \Psi(x) a(x)$  at positions  $x_n = n/B_x$  result.

Reconstruction of  $\Psi(x)$  from complex sampled data  $\Psi(n/B_x)$  by filtering with ideal lowpass (transm. 1 in band  $|\xi| < B_x/2$ , 0 outside), i. e., with interpolation recipe (**no linear interpolation**):

$$\Psi(x) = \sum_{n=-\infty}^{+\infty} \Psi(nX_a) \frac{\sin((x - nX_a)\pi B_x)}{(x - nX_a)\pi B_x}, \quad X_a = \frac{1}{B_x}, \quad X_{a1} = \frac{1}{2B_x}$$

**Oversampling needed for linear interpolation!**



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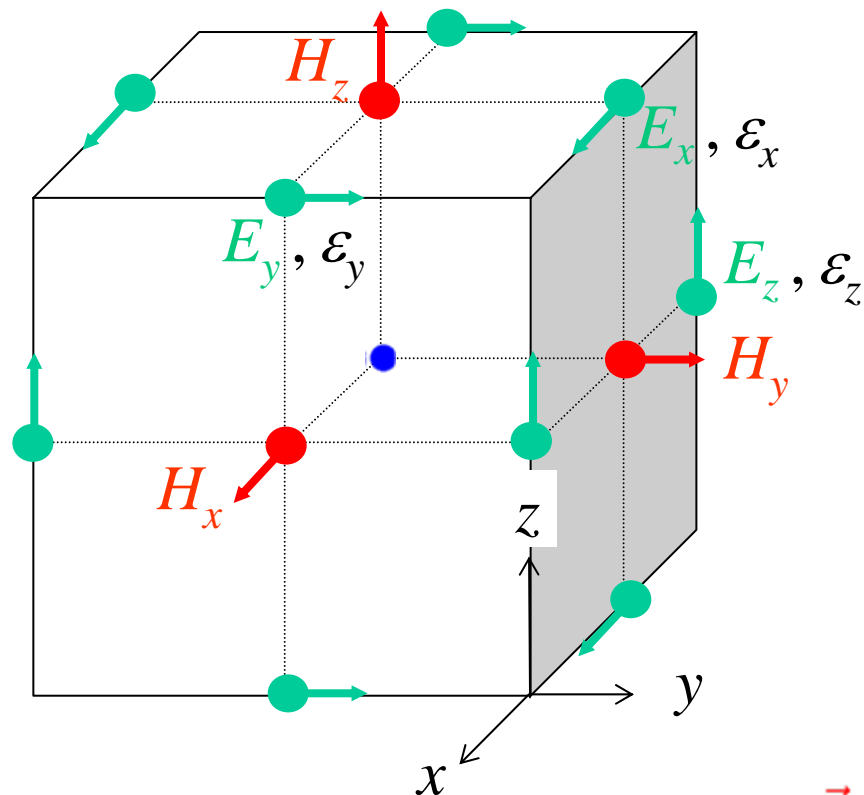
# Yee's Lattice with Cubic Unit Cells of Size $\Delta x \times \Delta y \times \Delta z$

Advantages of Yee algorithm:

- 1<sup>st</sup>-order DE for  $\vec{E}$  and  $\vec{H}$ ,
- more robust than 2<sup>nd</sup>-order wave equation for  $\vec{E}$  or  $\vec{H}$ .
- $\vec{E}$  and  $\vec{H}$  singularities naturally implemented.

cell centre:

$$(i, j, k) \equiv (i\Delta x, j\Delta y, k\Delta z)$$



$$\text{Faraday: } \text{curl } \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$$

$$\text{Ampere: } \text{curl } \vec{H} = \epsilon \frac{\partial \vec{E}}{\partial t}$$

- Each  $\vec{E}$  component surrounded by 4 circulating  $\vec{H}$  components
- each  $\vec{H}$  component surrounded by 4 circulating  $\vec{E}$  components



# Finite Differences in Cartesian Coordinates

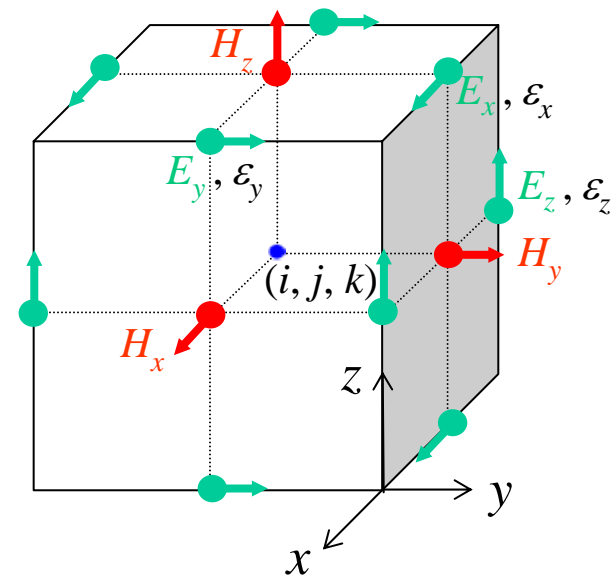
Centered finite-differences over  $\pm\frac{1}{2}\Delta x$ ,  $\pm\frac{1}{2}\Delta t$

at  $t = n\Delta t$ ,  $\vec{r} = i\Delta x \vec{e}_x + j\Delta y \vec{e}_y + k\Delta z \vec{e}_z$

for  $u(n\Delta t, i\Delta x, j\Delta y, k\Delta z) = u_{i,j,k}^n$  :

$$\frac{\partial u_{i,j,k}^n}{\partial x} = \frac{u_{i+1/2,j,k}^n - u_{i-1/2,j,k}^n}{\Delta x} + \mathcal{O}(\Delta x^2)$$

$$\frac{\partial u_{i,j,k}^n}{\partial t} = \frac{u_{i,j,k}^{n+1/2} - u_{i,j,k}^{n-1/2}}{\Delta t} + \mathcal{O}(\Delta t^2)$$



- Second-order accurate central differencing to the right and left of observation point  $(i, j, k)$  by  $\frac{1}{2}\Delta x$ .
- Interleaved  $\vec{E}$  and  $\vec{H}$  components at intervals  $\frac{1}{2}\Delta x$ .
- Second-order accurate central differencing to the past and future of observation point  $n$  by  $\frac{1}{2}\Delta t$ .
- Interleaved  $\vec{E}$  and  $\vec{H}$  components at intervals  $\frac{1}{2}\Delta t$ .
- This leads to a so-called “leapfrog” algorithm.

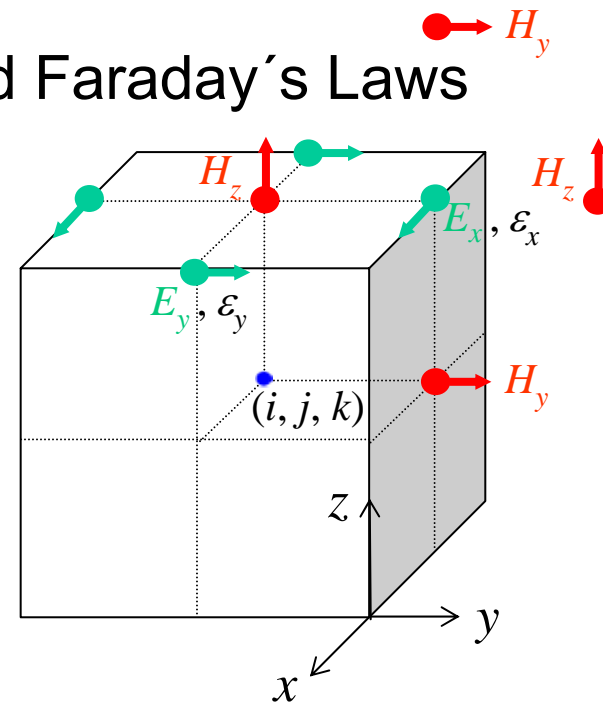




# Six Update Equations for Ampere's and Faraday's Laws

$$E_x|_{i,j+\frac{1}{2},k+\frac{1}{2}}^{n+\frac{1}{2}} = E_x|_{i,j+\frac{1}{2},k+\frac{1}{2}}^{n-\frac{1}{2}}$$

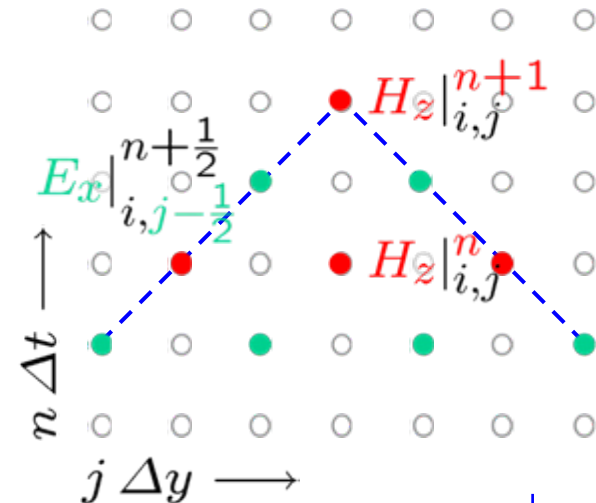
$$+ \frac{\Delta t}{\epsilon_{i,j+\frac{1}{2},k+\frac{1}{2}}} \left( \frac{H_z|_{i,j,k+\frac{1}{2}}^n - H_z|_{i,j+1,k+\frac{1}{2}}^n}{\Delta y} - \frac{H_y|_{i,j+\frac{1}{2},k+1}^n - H_y|_{i,j+\frac{1}{2},k}^n}{\Delta z} \right)$$



$$\frac{\partial H_z}{\partial t} = -\frac{1}{\mu_z} \left( \frac{\partial E_y}{\partial x} - \frac{\partial E_x}{\partial y} \right)$$

$$H_z|_{i,j,k+\frac{1}{2}}^{n+1} = H_z|_{i,j,k+\frac{1}{2}}^n$$

$$- \frac{\Delta t}{\mu_{i,j,k+\frac{1}{2}}} \left( \frac{E_y|_{i+\frac{1}{2},j,k+\frac{1}{2}}^{n+\frac{1}{2}} - E_y|_{i-\frac{1}{2},j,k+\frac{1}{2}}^{n+\frac{1}{2}}}{\Delta x} - \frac{E_x|_{i,j+\frac{1}{2},k+\frac{1}{2}}^{n+\frac{1}{2}} - E_x|_{i,j-\frac{1}{2},k+\frac{1}{2}}^{n+\frac{1}{2}}}{\Delta y} \right)$$





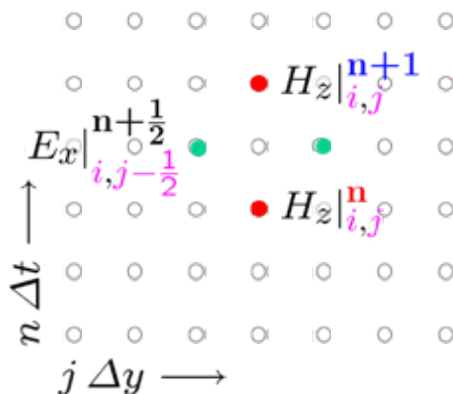
# Why Leapfrog?

**Leapfrog** A game in which players in turn vault with parted legs over others who are bending down.

(New Oxford Dictionary of English. Oxford Univ. Press, New York 1998)

$$H_z|_{i,j,k+1/2}^{n+1} = H_z|_{i,j,k+1/2}^n$$

$$- \frac{\Delta t}{\mu_{i,j,k+1/2}} \left( \frac{E_y|_{i+1/2,j,k+1/2}^{n+1/2} - E_y|_{i-1/2,j,k+1/2}^{n+1/2}}{\Delta x} - \frac{E_x|_{i,j+1/2,k+1/2}^{n+1/2} - E_x|_{i,j-1/2,k+1/2}^{n+1/2}}{\Delta y} \right)$$



Start at time  $n$

Arrive at time  $n + 1$



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- Wavelets
  - What are they good for?
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  - Higher-order finite-differences
  - Method of weighted residuals: Collocation
- Wavelet FDTD
  - Numerical dispersion, stability and accuracy
  - Summary, and what should be done further
  - Dispersive and nonlinear media
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# Numerical Dispersion

Maxwell's equations, homogeneous lossless space, parameters  
 $\epsilon = \epsilon_0 \epsilon_r$ ,  $\mu = \mu_0 \mu_r$ ,  $n^2 = \epsilon_r \mu_r$ ,  $c^2 = 1/(\epsilon_0 \mu_0)$ ,  $Z = \sqrt{\mu/\epsilon}$ :

$$j \operatorname{curl} \left( \sqrt{\frac{\mu}{\epsilon}} \vec{H} \right) = j \sqrt{\epsilon \mu} \frac{\partial}{\partial t} (\vec{E}), \quad -\operatorname{curl} (\vec{E}) = \sqrt{\epsilon \mu} \frac{\partial}{\partial t} \left( \sqrt{\frac{\mu}{\epsilon}} \vec{H} \right)$$

In compact form, and with  $\vec{V} = Z \vec{H} + j \vec{E}$ :

$$j \operatorname{curl} (Z \vec{H} + j \vec{E}) = \frac{n}{c} \frac{\partial}{\partial t} (Z \vec{H} + j \vec{E}), \quad j \operatorname{curl} \vec{V} = \frac{n}{c} \frac{\partial \vec{V}}{\partial t}$$

Plane wave  $\vec{V}(t, \vec{r}) = \vec{V}_0 \exp [j (\omega_0 t - \vec{k} \cdot \vec{r})]$  with  $|\vec{k}|^2 = (n\omega_0/c)^2$ ,  
 sampled in time  $t = n\Delta t$  and space  $\vec{r} = i \Delta x \vec{e}_x + j \Delta y \vec{e}_y + k \Delta z \vec{e}_z$ .  
 Differential operator scheme:

$$\frac{\partial \vec{V}_{i,j,k}^n}{\partial t} = \frac{\vec{V}_{i,j,k}^{n+\frac{1}{2}} - \vec{V}_{i,j,k}^{n-\frac{1}{2}}}{\Delta t} = \frac{e^{j\omega_0 \frac{\Delta t}{2}} - e^{-j\omega_0 \frac{\Delta t}{2}}}{\Delta t} \vec{V}_{i,j,k}^n = j\omega_0 \frac{\sin(\omega_0 \frac{\Delta t}{2})}{\omega_0 \frac{\Delta t}{2}} \vec{V}_{i,j,k}^n$$



# Numerical Dispersion

Plane wave  $\vec{V}(t, \vec{r}) = \vec{V}_0 \exp [j (\omega_0 t - \vec{k} \cdot \vec{r})]$  with  $|\vec{k}|^2 = (n\omega_0/c)^2$ .  
Numerical differential operators ( $q = x, y, z$ ):

$$\partial_t \vec{V}_{i,j,k}^n = j\omega_0 \frac{\sin\left(\omega_0 \frac{\Delta t}{2}\right)}{\omega_0 \frac{\Delta t}{2}} \vec{V}_{i,j,k}^n, \quad \partial_q \vec{V}_{i,j,k}^n = -j k_q \frac{\sin\left(k_q \frac{\Delta q}{2}\right)}{k_q \frac{\Delta q}{2}} \vec{V}_{i,j,k}^n$$

For a plane wave to fulfill the wave equation  $\nabla^2 \vec{V} = \frac{n^2}{c^2} \partial_t^2 \vec{V}$ :

$$\left(-k_x^2 - k_y^2 - k_z^2\right) \vec{V} = -\frac{n^2}{c^2} \omega_0^2 \vec{V}, \quad |\vec{k}|^2 = \left(n \frac{\omega_0}{c}\right)^2 = n^2 k_0^2$$

For the numerical wave an equivalent condition must hold:

$$\begin{aligned} \left(-j k_x \frac{\sin\left(k_x \frac{\Delta x}{2}\right)}{k_x \frac{\Delta x}{2}}\right)^2 + \left(-j k_y \frac{\sin\left(k_y \frac{\Delta y}{2}\right)}{k_y \frac{\Delta y}{2}}\right)^2 + \left(-j k_z \frac{\sin\left(k_z \frac{\Delta z}{2}\right)}{k_z \frac{\Delta z}{2}}\right)^2 \\ = \frac{n^2}{c^2} \left(-j \omega_0 \frac{\sin\left(\omega_0 \frac{\Delta t}{2}\right)}{\omega_0 \frac{\Delta t}{2}}\right)^2 \end{aligned}$$



# 1D Numerical Dispersion Artefact

Numerical wave solves Maxwell's equations if (refractive index  $n$ ):

$$\left(-j k_x \frac{\sin\left(k_x \frac{\Delta x}{2}\right)}{k_x \frac{\Delta x}{2}}\right)^2 + \left(-j k_y \frac{\sin\left(k_y \frac{\Delta y}{2}\right)}{k_y \frac{\Delta y}{2}}\right)^2 + \left(-j k_z \frac{\sin\left(k_z \frac{\Delta z}{2}\right)}{k_z \frac{\Delta z}{2}}\right)^2 = \frac{n^2}{c^2} \left(-j \omega_0 \frac{\sin\left(\omega_0 \frac{\Delta t}{2}\right)}{\omega_0 \frac{\Delta t}{2}}\right)^2$$

1D case,  $k_x = k_y = 0$ :

$$\left(\frac{\sin(k_z \Delta z/2)}{\Delta z/2}\right)^2 = \frac{n^2}{c^2} \left(\frac{\sin(\omega_0 \Delta t/2)}{\Delta t/2}\right)^2,$$

$$\sin^2\left(k_z \frac{\Delta z}{2}\right) = \left(\frac{n \Delta z}{c \Delta t}\right)^2 \sin^2\left(\omega_0 \frac{\Delta t}{2}\right),$$

$$\sin\left(\omega_0 \frac{\Delta t}{2}\right) = +\frac{c \Delta t}{n \Delta z} \sin\left(k_z \frac{\Delta z}{2}\right) \quad (\text{stability factor } S = \frac{c \Delta t}{n \Delta z}),$$

$$\omega_0 \Delta t = +2 \arcsin \left[ S \sin\left(k_z \frac{\Delta z}{2}\right) \right],$$

$$n \frac{\omega_0}{c} = \{n \Delta z = c \Delta t\} = k_z$$



# 1D Stability and Numerical Dispersion Artefact (refract. index $n$ )

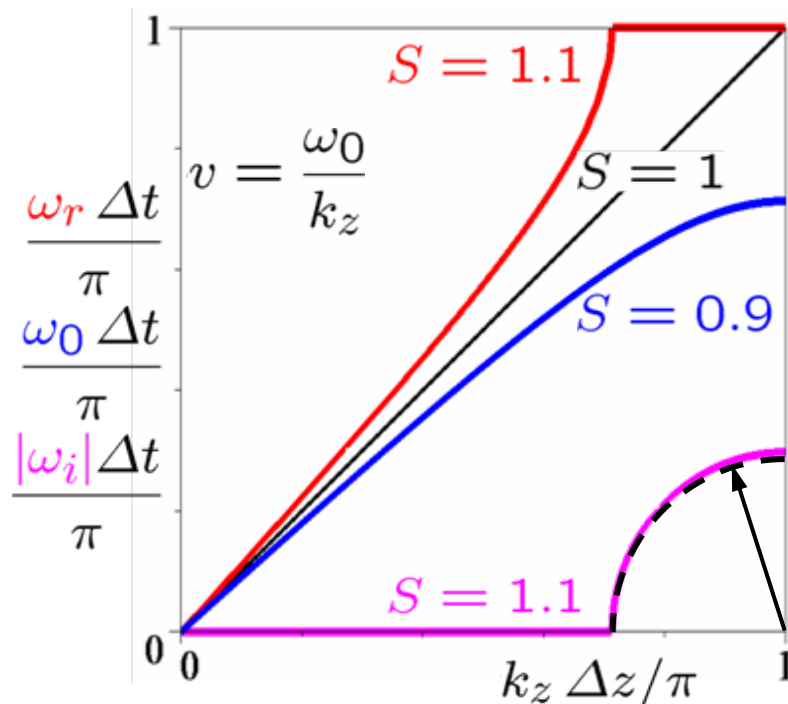
1D case,  $k_x = k_y = 0$ , stability factor (Courant number)  $S = \frac{c\Delta t}{n\Delta z}$ :

$$\sin\left(\omega_0 \frac{\Delta t}{2}\right) = S \sin\left(k_z \frac{\Delta z}{2}\right) \leq S \quad (0 \leq k_z \Delta z \leq \pi), \quad \omega_0 = \omega_r + j\omega_i$$

$$\omega_0 \Delta t = 2 \arcsin\left[S \sin\left(k_z \frac{\Delta z}{2}\right)\right]$$

Numerical 1D wave (increasing!):

$$\exp\left[+|\omega_i|t\right] \exp\left[j(\omega_r t - k_z z)\right]$$



Stability live (2D FDTD):  stbl  100Δn  unstbl



Dispersion live (2D FDTD):  0.999  S  0.1   $k_z \Delta z = \pi/2$



## 3D Stability and Numerical Dispersion

Numerical wave is solution of Maxwell's equations if

$$\sin\left(\omega_0 \frac{\Delta t}{2}\right) = \frac{c\Delta t}{n} \sqrt{\left(\frac{\sin\left(k_x \frac{\Delta x}{2}\right)}{\Delta x}\right)^2 + \left(\frac{\sin\left(k_y \frac{\Delta y}{2}\right)}{\Delta y}\right)^2 + \left(\frac{\sin\left(k_z \frac{\Delta z}{2}\right)}{\Delta z}\right)^2} \leq 1$$

or if :

$$S = S_{\text{FD2}} := \frac{c}{n} \frac{\Delta t}{\Delta l} = \frac{c\Delta t}{n} \sqrt{\left(\frac{1}{\Delta x}\right)^2 + \left(\frac{1}{\Delta y}\right)^2 + \left(\frac{1}{\Delta z}\right)^2} \leq 1$$

$\Delta l$  is maximum allowable spatial sampling interval for real propagation constant  $\vec{k} = 2\pi\vec{\kappa}$  (spatial frequency  $\kappa = n/\lambda$ ,  $k_s \frac{\Delta s}{2} = \frac{\pi}{2}$ ):

$$\kappa = \frac{n}{\lambda} = \sqrt{\xi^2 + \eta^2 + \zeta^2} = \sqrt{\left(\frac{1}{2\Delta x}\right)^2 + \left(\frac{1}{2\Delta y}\right)^2 + \left(\frac{1}{2\Delta z}\right)^2} = \frac{1}{2\Delta l}$$

Stability limit:

$$\Delta t = \frac{\Delta l}{c/n} = \frac{1}{2f_0} \frac{2\Delta l}{\lambda/n} = \frac{1}{2f_0} \quad \text{for} \quad \Delta l = \frac{\lambda/n}{2} = \frac{1}{2\kappa}$$



# 3D Plane Wave Dispersion and Accuracy — Choice of Step Sizes

**Stability**  $S = \frac{c}{n} \frac{\Delta t}{\Delta l}$ . Keep clear from dangerous  $S = 1$ , i. e., from

$$\Delta t = \frac{\Delta l}{c/n} = \frac{1}{2f_0} \frac{2\Delta l}{\lambda/n} = \frac{1}{2f_0} \quad \text{for} \quad \Delta l = \frac{\lambda/n}{2} = \frac{1}{2\kappa}. \quad \blacktriangleleft$$

Signal frequency  $f_0 = 1/T$  associated with spatial frequency  $\kappa = 1/(\lambda/n)$ . No temporal detail smaller than period  $T$ , no spatial detail smaller than medium wavelength  $\lambda/n$  may be resolved with this specific plane wave. The sampling recipe above yields an **exact representation**: No dispersion, but **risk of global or local instability**.

**Accuracy** Choose a safe  $S < 1$ . For low dispersion error  $< 1\%$  ( $\arcsin(x) \leq 1.01x$ ) choose  $k_s \frac{\Delta s}{2} \leq 0.16 \frac{\pi}{2} \ll \frac{\pi}{2}$  ( $s = x, y, z$ ), i. e.,

$$\omega_0 \Delta t = 2 \arcsin\left(S \frac{1}{2} k \Delta l\right) \approx S k \Delta l \leq S \times 0.16 \pi \quad (|\vec{k}| = k = 2\pi\kappa), \quad \blacktriangleleft$$

$$\omega_0 \approx \frac{c}{n} k \quad \text{for } S < 1 \text{ and } \Delta t \leq S \frac{0.16}{2f_0} \text{ or } \Delta l \leq \frac{0.16}{2\kappa}$$





## 3D Stability — Intuitive Explanation for $\Delta = \Delta x = \Delta y = \Delta z$

- A numerical value propagates  $\sqrt{3} \Delta \dots 3 \Delta$  per  $3 \Delta t$ , given the local nature of the spatial difference used in the Yee algorithm, i. e., at a speed  $\tilde{v} \geq \tilde{v}_\Delta = \frac{\sqrt{3} \Delta}{3 \Delta t} = \frac{1}{\sqrt{3}} \frac{\Delta}{\Delta t}$ .

Courant:  $S = \frac{c}{n} \frac{\Delta t}{\Delta l}$   
 $S = \frac{v_{\text{wrlld}}}{\tilde{v}}$

- If the dispersion of the numerical wave is chosen such that its phase velocity  $v > \tilde{v}$  exceeds the **slowest speed  $\tilde{v}_\Delta$  provided by the algorithm**, instability occurs.
- If the dispersion of the numerical wave leads to a phase velocity  $v < \tilde{v}$ , which is smaller than the **slowest algorithm speed  $\tilde{v}_\Delta$** , the computation is stable.
- For a good accuracy  $v \approx v_{\text{wrlld}}$  the step sizes

$$\Delta t \leq \frac{1}{2f_0}, \quad \Delta l = \left[ \left( \frac{1}{\Delta x} \right)^2 + \left( \frac{1}{\Delta y} \right)^2 + \left( \frac{1}{\Delta z} \right)^2 \right]^{-\frac{1}{2}} \leq \frac{1}{2\kappa} = \frac{\lambda/n}{2} = \frac{c/n}{2f_0}$$

must be significantly smaller than the Nyquist limit.



# Outline

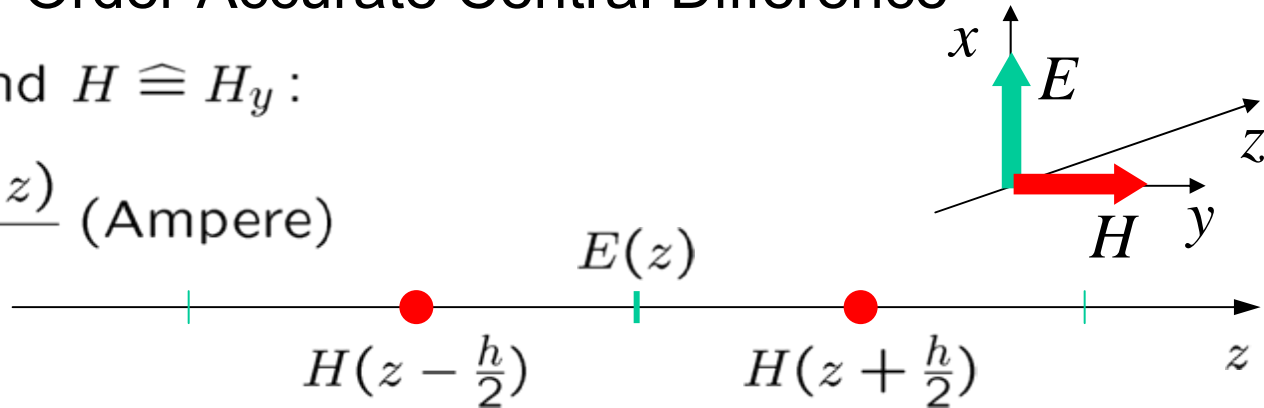
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## Second-Order Accurate Central Difference

1D wave  $E \cong E_x$  and  $H \cong H_y$ :

$$-\varepsilon \frac{\partial E(t, z)}{\partial t} = \frac{\partial H(t, z)}{\partial z} \quad (\text{Ampere})$$



Taylor series expansion:

$$H\left(z + \frac{h}{2}\right) = H(z) + \frac{h}{2} \frac{\partial H(z)}{\partial z} + \frac{1}{2!} \frac{h^2}{4} \frac{\partial^2 H(z)}{\partial z^2} + \frac{1}{3!} \frac{h^3}{8} \frac{\partial^3 H(z)}{\partial z^3} + \dots$$

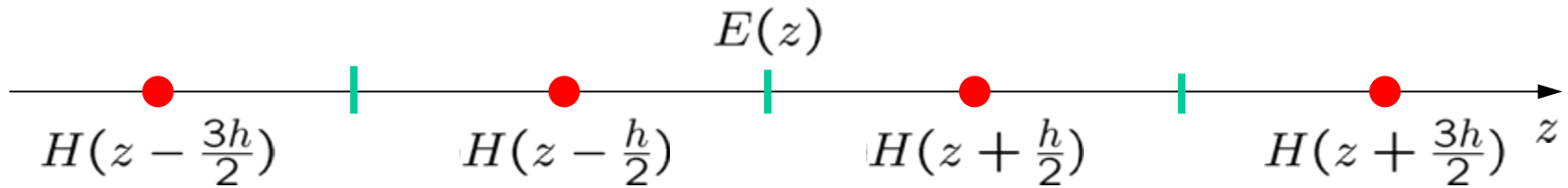
$$H\left(z - \frac{h}{2}\right) = H(z) - \frac{h}{2} \frac{\partial H(z)}{\partial z} + \frac{1}{2!} \frac{h^2}{4} \frac{\partial^2 H(z)}{\partial z^2} - \frac{1}{3!} \frac{h^3}{8} \frac{\partial^3 H(z)}{\partial z^3} + \dots$$

Second-order accurate central difference:

$$\frac{\partial H(z)}{\partial z} = \frac{1}{h} \left[ H\left(z + \frac{h}{2}\right) - H\left(z - \frac{h}{2}\right) \right] + \underbrace{\frac{h^2}{24} \frac{\partial^3 H(z)}{\partial z^3} + \dots}_{\mathcal{O}(h^2)}$$



# Fourth-Order Accurate Central Difference



Second-order accurate central difference:

$$27h \frac{\partial H(z)}{\partial z} \approx 27 \left[ H\left(z + \frac{h}{2}\right) - H\left(z - \frac{h}{2}\right) \right] + \underbrace{\frac{27h^3}{24} \frac{\partial^3 H(z)}{\partial z^3}}_{\text{substitute}}$$

Taylor series expansion:

$$\underbrace{H\left(z + \frac{3h}{2}\right)}_{-} \approx \cancel{H(z)} + \frac{3h}{2} \frac{\partial H(z)}{\partial z} + \cancel{\frac{9h^2}{8} \frac{\partial^2 H(z)}{\partial z^2}} + \frac{27h^3}{48} \frac{\partial^3 H(z)}{\partial z^3} + \dots$$

$$\underbrace{H\left(z - \frac{3h}{2}\right)}_{-} \approx \cancel{H(z)} - \frac{3h}{2} \frac{\partial H(z)}{\partial z} + \cancel{\frac{9h^2}{8} \frac{\partial^2 H(z)}{\partial z^2}} - \frac{27h^3}{48} \frac{\partial^3 H(z)}{\partial z^3} + \dots$$

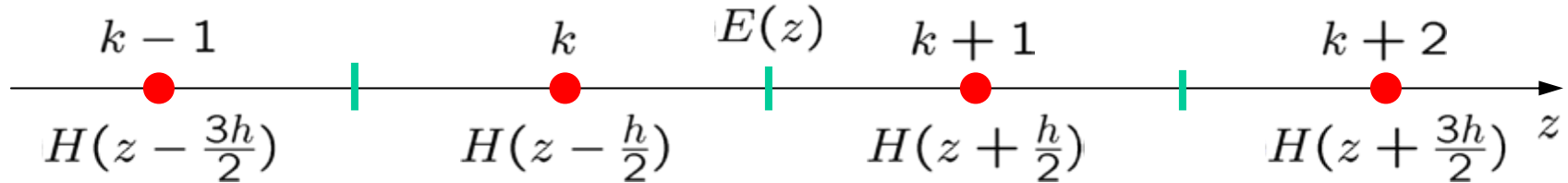
Fourth-order accurate central difference:

$$\frac{\partial H(z)}{\partial z} = \frac{27}{24h} \left[ H\left(z + \frac{h}{2}\right) - H\left(z - \frac{h}{2}\right) \right] + \frac{-1}{24h} \left[ H\left(z + \frac{3h}{2}\right) - H\left(z - \frac{3h}{2}\right) \right] + \mathcal{O}(h^4)$$



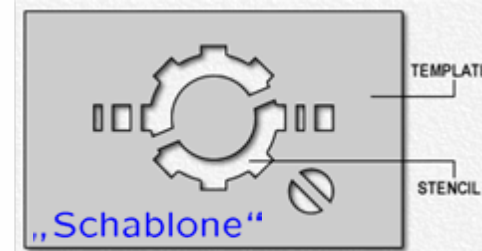
# High-Order Accurate Central Difference

1D wave  $E \hat{=} E_x$  and  $H \hat{=} H_y$ :  $-\varepsilon \frac{\partial E(t, z)}{\partial t} = \frac{\partial H(t, z)}{\partial z}$  (Ampere)



Second-order,  $L = 1$ :

$$\frac{\partial H(z)}{\partial z} = \frac{H_{k+1} - H_k}{\Delta z} + \mathcal{O}(\Delta z^2)$$



Fourth-order,  $L = 2$ :

$$\frac{\partial H(z)}{\partial z} = \frac{\frac{27}{24}H_{k+1} - \frac{27}{24}H_k - \frac{1}{24}H_{k+2} + \frac{1}{24}H_{k-1}}{\Delta z} + \mathcal{O}(\Delta z^4)$$

$(2L)$ th order, one-sided stencil length  $L$ , coefficients  $a_{l-1} = -a_{-l}$ :

$$\frac{\partial H(z)}{\partial z} = \frac{1}{\Delta z} \sum_{l=-L}^{L-1} a_l H_{k+1+l} + \mathcal{O}(\Delta z^{2L}), \quad \text{terms } H_{k+1-L}, \dots, H_{k+L}$$



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# Method of Weighted Residuals

Operator  $\mathcal{H}$  (e. g.,  $\mathcal{H} = \partial/\partial z + \partial/\partial t$ ), function  $G \rightsquigarrow$  **exact** solution:

$$\mathcal{H}F_{\text{ex}}(z, t) = G(z, t)$$

Seperable basis functions  $\phi_k(z) h_n(t) \rightsquigarrow$  residual error  $R(z, t)$ :

$$F(z, t) = \sum_{n=1}^N \sum_{k=1}^K c_{k,n} \phi_k(z) h_n(t), \quad R(z, t) := \mathcal{H}F(z, t) - G(z, t)$$

Weight functions  $\tilde{v}_k(z) \tilde{u}_n(t)$  force weighted- $R$  integrals to zero:

$$\iint R(z, t) \tilde{v}_k(z) \tilde{u}_n(t) dz dt \stackrel{!}{=} 0, \quad k = 1, 2, \dots, K, \quad n = 1, 2, \dots, N$$

System of  $K \times N$  (possibly nonlinear) equations for coefficients  $c_{k,n}$

- **Petrov-Galerkin**: spaces of  $\tilde{v}_k, \tilde{u}_n$  different from  $\phi_k, h_n$
- **Bubnov-Galerkin**:  $\tilde{v}_k = \phi_k^*, \tilde{u}_n = h_n^*$ ,
- **Method of moments**: injective (1:1) operator  $\tilde{v}_k \tilde{u}_n = \mathcal{M}(\phi_k^* h_n^*)$
- **Least squares**:  $\mathcal{M} = \mathcal{H}$
- **Finite elements**: compact-support  $\phi_k, h_n$ ;  $c_{k,n}$  (mesh nodes); **BG**
- **Collocation (point matching)**:  $\tilde{v}_k(z) = \delta(z - z_k), \tilde{u}_n(t) = \delta(t - t_n)$



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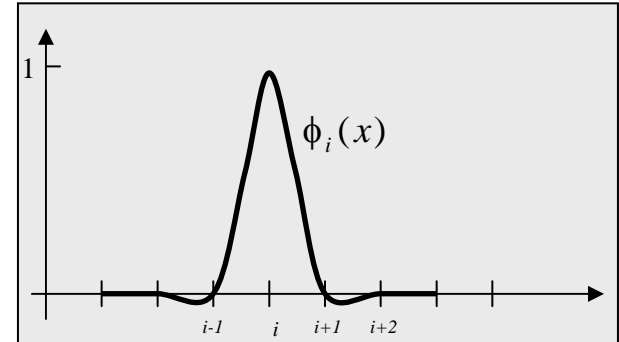
# Modelling: Wavelet-based FDTD Method

## Advantages FDTD:

Time domain solution available

(i.e. pulses / transients can be studied)

Readily extended to include nonlinear materials



## Disadvantages FDTD:

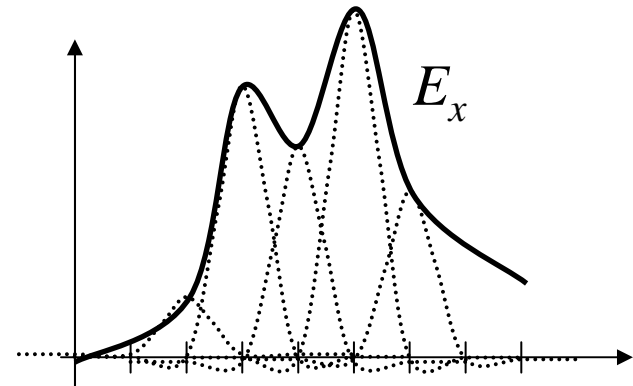
Much memory and time are needed for the computation



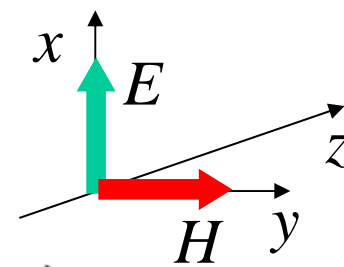
## Promise of interpolating wavelet basis:

Numerical dispersion decreased dramatically

⇒ Coarser grid (less memory)



# Collocation Method



1D wave  $E \hat{=} E_x$  and  $H \hat{=} H_y$ :

$$-\varepsilon \frac{\partial E(t, z)}{\partial t} = \frac{\partial H(t, z)}{\partial z} \text{ (Ampere), } -\mu \frac{\partial H(t, z)}{\partial t} = \frac{\partial E(t, z)}{\partial z} \text{ (Faraday)}$$

Spatial DD $p$  MRA & Haar scaling function  $\phi_h(t - n\Delta t) \hat{=} h_n(t)$ :

$$-\varepsilon \frac{\partial}{\partial t} E(t, z) = \sum_{k,n} E_{k,n}^{\phi^{J-1}} \phi_k^{J-1}(z) \underline{h_n(t)} \quad \left| \iint dz dt \tilde{\phi}_{k'}^{J-1}(z) h_{n'+\frac{1}{2}}(t) \right.$$

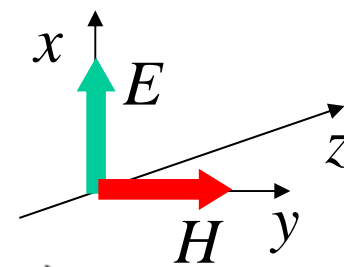
$$\begin{aligned} \text{scale } J \longrightarrow &= \sum_{k,n} E_{k,n}^{\phi^J} \phi_k^J(z) h_n(t) + \sum_{k,n} E_{k,n}^{\psi^J} \psi_k^J(z) h_n(t) \quad \begin{array}{c} \boxed{h_0(t)} \\ \text{---} \\ 0 \quad 1/\Delta t \end{array} \\ &= \sum_{k,n} E_{k,n}^{\phi^{J+3}} \phi_k^{J+3}(z) h_n(t) + \sum_{j=J}^{J+3} \sum_{k,n} E_{k,n}^{\psi^j} \psi_k^j(z) h_n(t) \end{aligned}$$

Spatially and temporarily interleaved  $H(t, z)$ ,  $\cdot(t)$  and  $\cdot(z)$  omitted:

$$\frac{\partial}{\partial z} H(t, z) = \sum_{k,n} H_{k+\frac{1}{2}, n+\frac{1}{2}}^{\phi^{J+3}} \underline{\phi_{k+\frac{1}{2}}^{J+3}} h_{n+\frac{1}{2}} + \sum_{j=J}^{J+3} \sum_{k,n} H_{k, n+\frac{1}{2}}^{\psi^j} \underline{\psi_{k+\frac{1}{2}}^j} h_{n+\frac{1}{2}}$$



# Collocation Method



1D wave  $E \hat{=} E_x$  and  $H \hat{=} H_y$ :

$$-\varepsilon \frac{\partial E(t, z)}{\partial t} = \frac{\partial H(t, z)}{\partial z} \text{ (Ampere), } -\mu \frac{\partial H(t, z)}{\partial t} = \frac{\partial E(t, z)}{\partial z} \text{ (Faraday)}$$

Spatial DD $p$  MRA & Haar scaling function  $\phi_h(t - n\Delta t) \hat{=} h_n(t)$ :

$$E(t, z) = \sum_{k,n} E_{k,n}^{\phi^{J-1}} \phi_k^{J-1}(z) h_n(t) \quad \left| \iint dz dt \tilde{\phi}_{k'}^J(z) h_{n'+\frac{1}{2}}(t) \right. \quad \blacktriangleright$$

$$-\varepsilon \frac{\partial}{\partial t} E(t, z) = \sum_{k,n} E_{k,n}^{\phi^J} \phi_k^J(z) \underline{h_n(t)} + \sum_{k,n} E_{k,n}^{\psi^J} \psi_k^J(z) \underline{h_n(t)} \quad \begin{array}{c} \boxed{h_0(t)} \\ \text{---} \\ 0 \quad 1 \quad t/\Delta t \end{array}$$

Spatially and temporarily interleaved  $H(t, z)$ ,  $\cdot(t)$  and  $\cdot(z)$  omitted:

$$\frac{\partial}{\partial z} H(t, z) = \sum_{k,n} H_{k+\frac{1}{2},n+\frac{1}{2}}^{\phi^J} \underline{\phi_{k+\frac{1}{2}}^J} h_{n+\frac{1}{2}} + \sum_{k,n} H_{k,n+\frac{1}{2}}^{\psi^J} \underline{\psi_{k+\frac{1}{2}}^J} h_{n+\frac{1}{2}}$$



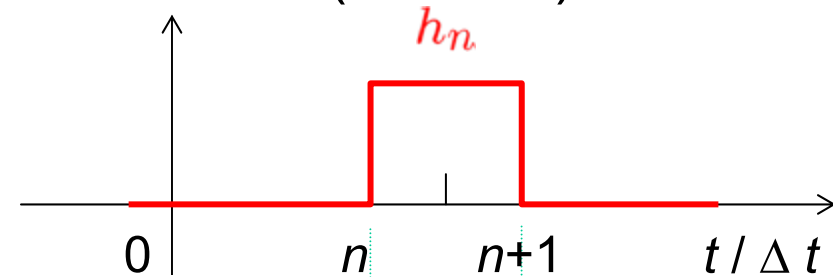
# Force Residual Error to Zero — LHS (Scale $J$ )

Inner product of time functions:

$$T = \int_{-\infty}^{+\infty} \frac{1}{\Delta t} \frac{dh_n(t)}{dt} h_{n'+\frac{1}{2}}(t) dt$$

$$= \int_{-\infty}^{+\infty} [\delta(t - n\Delta t) - \delta(t - (n+1)\Delta t)] h_{n'+\frac{1}{2}}(t) dt / \Delta t$$

$$= h([n - (n' + \frac{1}{2})]\Delta t) - h([(n+1) - (n' + \frac{1}{2})]\Delta t) / \Delta t$$

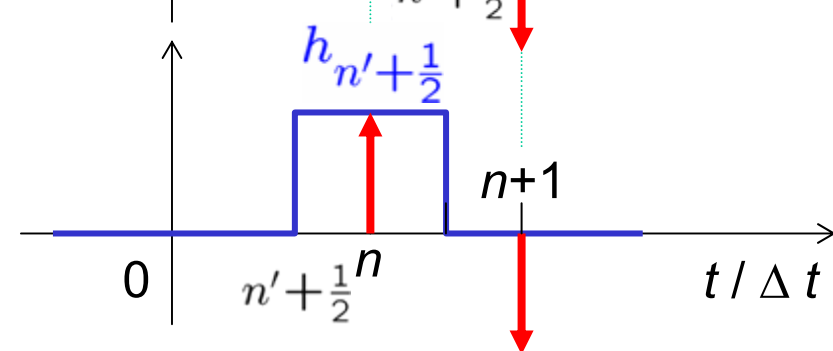
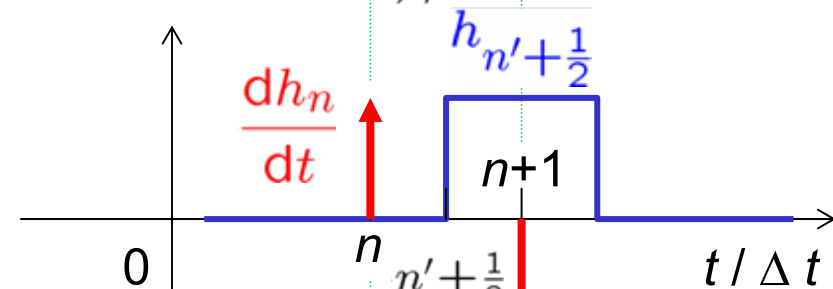


$$T = (\delta_{n n'+1} - \delta_{n n'}) / \Delta t$$

Inner product of spatial functions:

$$\Sigma = \int_{-\infty}^{+\infty} \phi_k^J(z) \tilde{\phi}_{k'}^J(z) dt = \delta_{k k'}$$

$$\int_{-\infty}^{+\infty} \psi_k^J(z) \tilde{\phi}_{k'}^J(z) dt = 0$$



$$\text{LHS} = -\varepsilon_{k'} \frac{E_{k',n'+1}^{\phi^J} - E_{k',n}^{\phi^J}}{\Delta t}$$



# Force Residual Error to Zero — RHS (Scale $J$ )

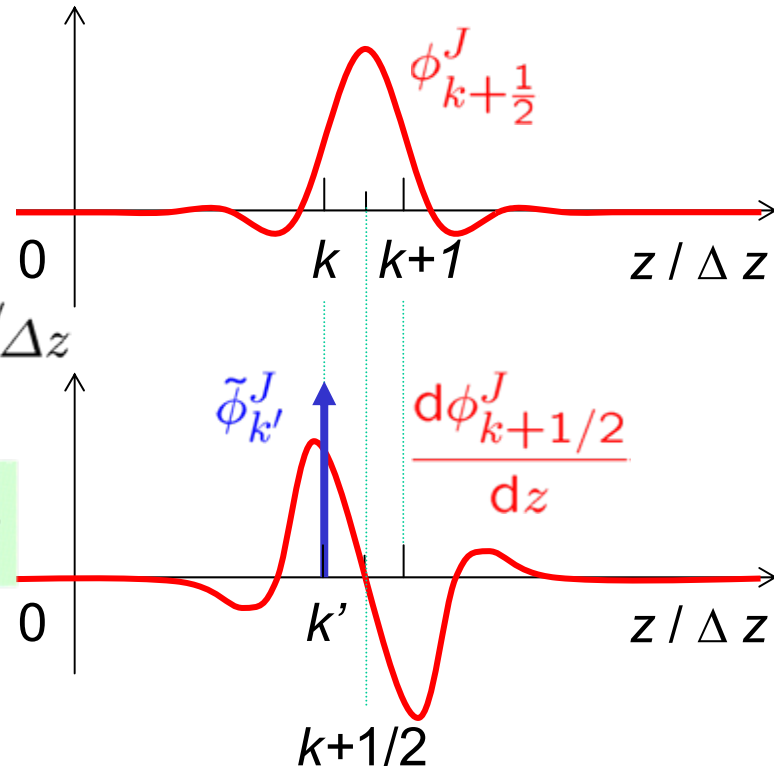
Inner product of spatial functions:

$$\Sigma = \int_{-\infty}^{+\infty} \frac{1}{\Delta z} \frac{d\phi_{k+1/2}^J(z)}{dz} \tilde{\phi}_{k'}^J(z) dz$$

$$= \int_{-\infty}^{+\infty} \frac{d\phi_{k+1/2}^J(z)}{dz} \delta(z - k' \Delta z) dz / \Delta z$$

$$\Sigma = \frac{1}{\Delta z} \left. \frac{d\phi_{k+1/2}^J(z)}{dz} \right|_{k' \Delta z} = a_l^{\phi^J \phi^J} / \Delta z$$

$l = k - k'$



Inner product of time functions:

$$T = \int_{-\infty}^{+\infty} \frac{1}{\sqrt{\Delta t}} h_{n+1/2}(t) \frac{1}{\sqrt{\Delta t}} h_{n'+1/2}(t) dt = \delta_{nn'}$$

$$-\varepsilon_{k'} \frac{E_{k',n'+1}^{\phi^J} - E_{k',n'}^{\phi^J}}{\Delta t} = \frac{\sum_l a_l^{\phi^J \phi^J} H_{k'+1/2+l, n'+1/2}^{\phi^J} + \sum_l a_l^{\psi^J \phi^J} H_{k'+1/2+l, n'+1/2}^{\psi^J}}{\Delta z}$$



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# Comparison of Wavelet DDP FDTD and Standard FDTD

$$-\varepsilon_{k'} \frac{E_{k',n'+1}^{\phi^J} - E_{k',n'}^{\phi^J}}{\Delta t} = \frac{\sum_l a_l^{\phi^J \phi^J} H_{k'+\frac{1}{2}+l, n'+\frac{1}{2}}^{\phi^J} + \sum_l a_l^{\psi^J \phi^J} H_{k'+\frac{1}{2}+l, n'+\frac{1}{2}}^{\psi^J}}{\Delta z}$$

$$a_l^{\phi^J \phi^J} = \left. \frac{d\phi_{k+\frac{1}{2}}^J(z)}{dz} \right|_{z=k'\Delta z} = \left. \frac{d\phi^J(z)}{dz} \right|_{z=-(l+\frac{1}{2})\Delta z} \quad \text{for } l = k - k', \quad -L \leq l \leq L - 1$$

**Wavelet FDTD:** Scaling function only ( $a_l^{\psi^J \phi^J} = 0$ ),  $l = -L, \dots, L-1$ , one-sided stencil size  $L = 1$ , new subscripts ( $k' = k + \frac{1}{2}$ ,  $n' = n - \frac{1}{2}$ ):

$$-\varepsilon_{k+\frac{1}{2}} \frac{E_{k+\frac{1}{2}, n+\frac{1}{2}}^{\phi^J} - E_{k+\frac{1}{2}, n-\frac{1}{2}}^{\phi^J}}{\Delta t} = \frac{a_0^{\phi^J \phi^J} H_{k+1, n}^{\phi^J} + a_{-1}^{\phi^J \phi^J} H_{k, n}^{\phi^J}}{\Delta z}$$

**Standard FDTD** with central differences ( $H_z|_{i, \cdot, k+\frac{1}{2}}^n = 0$ ):

$$-\varepsilon_{i, j+\frac{1}{2}, k+\frac{1}{2}} \frac{E_x|_{i, j+\frac{1}{2}, k+\frac{1}{2}}^{n+\frac{1}{2}} - E_x|_{i, j+\frac{1}{2}, k+\frac{1}{2}}^{n-\frac{1}{2}}}{\Delta t} = \frac{H_y|_{i, j+\frac{1}{2}, k+1}^n - H_y|_{i, j+\frac{1}{2}, k}^n}{\Delta z}$$



# 2D Stability and Numerical Dispersion for Higher-Order FDTD

Numerical wave is solution of Maxwell's equations if

$$\sin\left(\omega_0 \frac{\Delta t}{2}\right) = \frac{c\Delta t}{n} \sqrt{\left(\sum_{l=0}^{L-1} a_l \frac{\sin\left(k_x \frac{\Delta x}{2}(2l+1)\right)}{\Delta x}\right)^2 + \left(\sum_{l=0}^{L-1} a_l \frac{\sin\left(k_z \frac{\Delta z}{2}(2l+1)\right)}{\Delta z}\right)^2} \leq 1$$

or if:

$$S_{\text{HFD}2L} = \left(\sum_{l=0}^{L-1} |a_l|\right) \underbrace{\frac{c\Delta t}{n} \sqrt{\left(\frac{1}{\Delta x}\right)^2 + \left(\frac{1}{\Delta z}\right)^2}}_{S_{\text{FD}2} \text{ for } \Delta t = (\Delta t)_{\text{FD}2}} \leq 1$$

Higher-order scheme for  $S_{\text{HFD}2L} = S_{\text{FD}2}$ :  $\Delta t = (\Delta t)_{\text{FD}2} / \left(\sum_{l=0}^{L-1} |a_l|\right)$

Stencil	$a_0$	$a_1$	$a_2$	$a_3$	$a_4$	$a_5$	$a_6$	$a_7$	$\sum_l  a_l $
HFD4	1.125	-0.042	0.	0.	0.	0.	0.	0.	1.1667
D2/DD2	1.229	-0.094	0.010	0.	0.	0.	0.	0.	1.3333
HFD6	1.172	-0.065	0.005	0.	0.	0.	0.	0.	1.2417
D4/DD4	1.311	-0.156	0.042	-0.009	0.001	0.000	-0.000	0.	1.5181



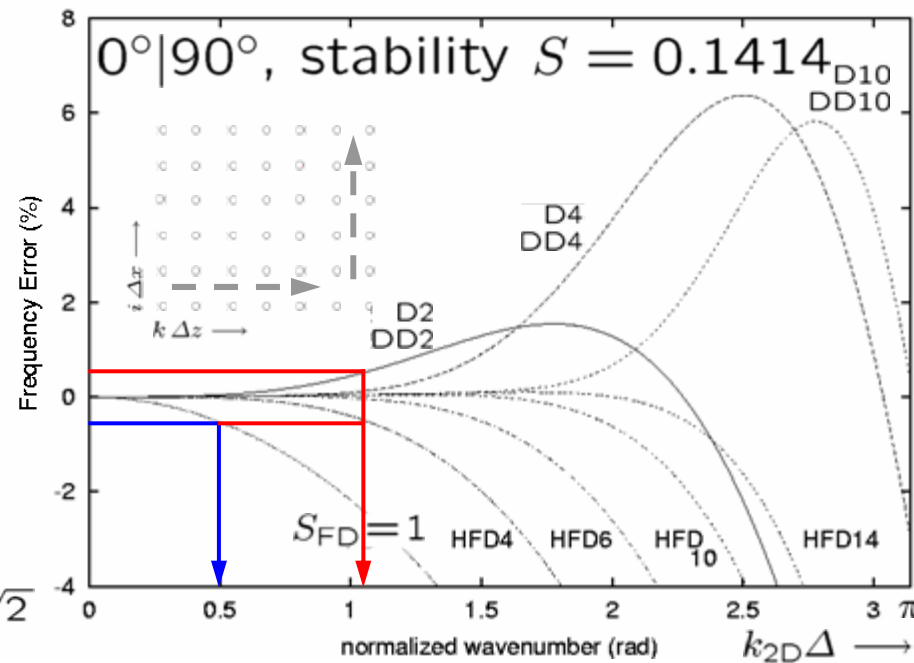
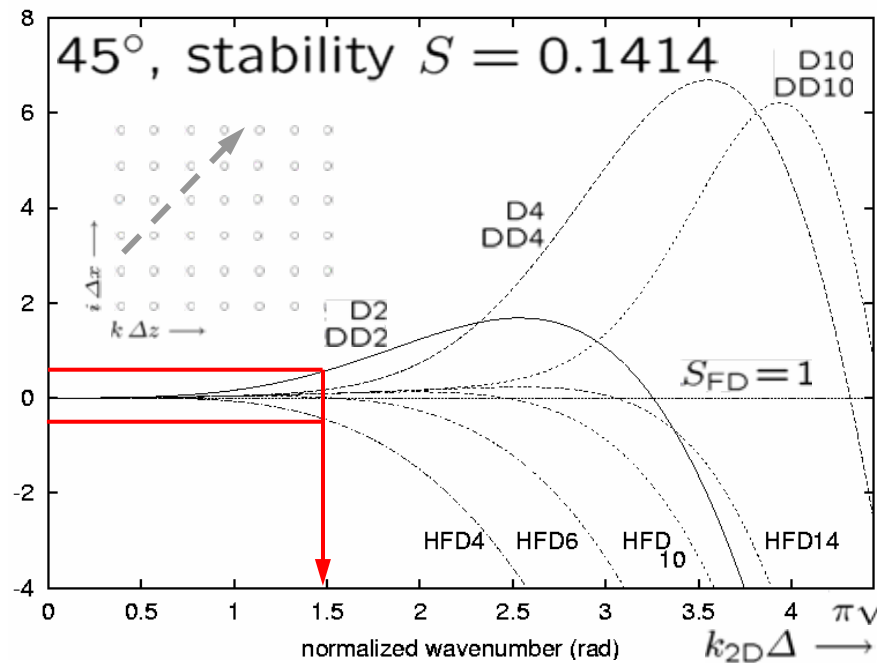


# 2D Numerical Dispersion Error. Stability: $S = \frac{c\Delta t}{n\Delta} \sqrt{D} \leq 1$

**FD:** Standard 2<sup>nd</sup>-order finite-differences of accuracy order  $\mathcal{O}(\Delta^2)$

**HFD<sub>q</sub>:** Higher-order FD,  $L=q/2$ , support  $[-L, L]$ , accuracy  $\mathcal{O}(\Delta^q)$

**D<sub>p</sub>, DD<sub>p</sub>:** Scaling fct, order  $L=2p-1$ , compact support  $[-L, L]$



Stencil	$a_0$	$a_1$	$a_2$	$a_3$	$a_4$	$a_5$	$a_6$	$a_7$	$\sum_l  a_l $
HFD4	1.125	-0.042	0.	0.	0.	0.	0.	0.	1.1667
D2/DD2	1.229	-0.094	0.010	0.	0.	0.	0.	0.	1.3333
HFD6	1.172	-0.065	0.005	0.	0.	0.	0.	0.	1.2417
D4/DD4	1.311	-0.156	0.042	-0.009	0.001	0.000	-0.000	0.	1.5181

Same stencil- $L$   
Same accuracy



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# Summary of Wavelet FDTD

## Pros:

- Difference operators basically broader (more coefficients), but detail coefficients may be truncated if fields do not change much in certain regions (less coefficients)  $\rightsquigarrow$  self-adaptiv
- Static use of  $\phi_k^J(t)$  and  $\psi_k^j(t)$  for  $0 \leq j \leq J$  in a predetermined high-resolution region of space  $\rightsquigarrow$  simple meshing

## Requirements:

- **Orthogonality** of expansion (synthesizing) functions  $\phi$ ,  $\psi$  and test (analyzing) functions  $\tilde{\phi}$ ,  $\tilde{\psi}$  to avoid solving a system of equations for each time step in a collocation setting
- **Vanishing moments** of wavelets up to order  $p$  assure rapid convergence and effective truncation of expansion with  $\mathcal{O}(s^{2+1})$  ◀
- **Compact support** to avoid truncation of difference operators (spurious modes)
- **Symmetry** of  $\phi$ ,  $\psi$  and  $\tilde{\phi}$ ,  $\tilde{\psi}$ , so that mirroring carries over to respective coefficients (but unsymm.  $D_p$  with symm. stencil!)



# Wavelet Families and their Properties

Property	morl	mexh	meyr	haar	dbN	symN	coifN	biorNr.Nd	rbioNr.Nd	gaus	dmey	cgau	cmor	fbsp	shan
Crude	•	•								•		•	•	•	•
Infinitely regular	•	•	•							•		•	•	•	•
Arbitrary regularity					•	•	•	•	•						
Compactly supported orthogonal				•	•	•	•								
Compactly supported biorthogonal								•	•						
Symmetry	•	•	•	•				•	•	•	•	•	•	•	•
Asymmetry					•										
Near symmetry						•	•								
Arbitrary number of vanishing moments					•	•	•	•	•						
Vanishing moments for $\phi$							•								
Existence of $\phi$			•	•	•	•	•	•	•						
Orthogonal analysis			•	•	•	•	•								
Biorthogonal analysis			•	•	•	•	•	•	•						
Exact reconstruction	≈	•	•	•	•	•	•	•	•	•	≈	•	•	•	•
FIR filters				•	•	•	•	•	•		•				
Continuous transform	•	•	•	•	•	•	•	•	•	•					
Discrete transform			•	•	•	•	•	•	•		•				
Fast algorithm				•	•	•	•	•	•		•				
Explicit expression	•	•		•					For splines	For splines	•		•	•	•

Misiti, M.; Misiti, Y.; Oppenheim, G.; Poggi, J.-M.: Wavelet toolbox user's guide, v. 3.1 (MATLAB release 2006b). Natick (MA): The Mathworks Inc. 2006. Page 6-90 ff. [http://www.mathworks.com/access/helpdesk/help/pdf\\_doc/wavelet/wavelet\\_ug.pdf](http://www.mathworks.com/access/helpdesk/help/pdf_doc/wavelet/wavelet_ug.pdf)



# Summary of Wavelet Methods

Daubechies wavelets  $D_p$  and Bubnov-Galerkin:

- ✓  $\phi$ ,  $\psi$  and  $\tilde{\phi}$ ,  $\tilde{\psi}$  are orthogonal
- Moments of  $\psi$  up to order  $p$  vanish, *could* be self-adaptive
- ✓ Compact support
- ✓ No symmetry, but stencil coefficients are symmetric

Deslauries-Dubuc interpolating wavelets  $DD_p$  and collocation:

- ✓  $\phi$ ,  $\psi$  and  $\tilde{\phi}$ ,  $\tilde{\psi}$  are biorthogonal
- Moments of  $\psi$  do *not* vanish, self-adaptiveness doubtful
- ✓ Compact support
- ✓ Symmetry
- $\phi^{DD_p}$ -stencil coefficients for collocation identical to those with  $\phi^{D_p}$  and Bubnov-Galerkin: **Methods equivalent!**



# Summary of Results, And What Should Be Done Further

## Results:

- **Equivalence** of  $\phi^{Dp}$ -Bubnov-Galerkin and  $\phi^{DDp}$ -collocation
- **Low dispersion** for  $\phi^{Dp}/\phi^{DDp}$  (reduced memory and run time)  
— but this *may* be equivalent to HFD6
- **Equivalence between  $\phi$  and  $\psi$**  via wavelet transform  $\rightsquigarrow$  all dispersion findings also valid for possible MRA

## What should be done?

- Chose higher-order time difference operator (relatively simple)
- Try MRA with Daubechies  $\phi^{Dp}$  and  $\psi^{Dp}$
- Chose or invent other wavelets with vanishing moments, observing other requirements



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# Dispersive and Nonlinear Media

## Lossy and dispersive materials:

- More realistic models (Debye, Lorentz)
- Wave propagation in a plasma

## Nonlinear modeling:

- High-power pulses (Kerr, Raman, self-steepening)
- Design of switches
- Design of wavelength converters (e. g., FWM)

## Problems and solutions:

- Stability issues
- ABC so far not well established
- PML applicable, also to higher-order FD methods
- Implemented with auxiliary differential equations (ADE)

Fujii, M.; Tahara, M; Sakagami, I; Freude, W.; Russer, P.: High-order FDTD and [auxiliary differential equation](#) formulation of optical pulse propagation in 2D Kerr and Raman nonlinear dispersive media. IEEE J. Quantum Electron. 40 (2004) 175–182

Fujii, M.; Omaki, N.; Tahara, M.; Sakagami, I.; Poulton, C.; Freude, W.; Russer, P.: Optimization of nonlinear [dispersive APML ABC](#) for the FDTD analysis of optical solitons. IEEE J. Quantum Electron. 41 (2005) 448–454

Fujii, M.; Koos, C.; Poulton, C.; Sakagami, I.; Leuthold, J.; Freude, W.: A simple and rigorous verification technique for [nonlinear FDTD](#) algorithm by optical parametric four-wave mixing. Microwave and Optical Technol. Lett. 48 (2006) 88–91

Fujii, M; Koos, C.; Poulton, C.; Leuthold, J.; Freude, W.: Nonlinear FDTD analysis and experimental validation of [four-wave mixing](#) in InGaAsP/InP racetrack micro-resonators. IEEE Photon. Technol. Lett. 18 (2006) 361–363





# FDTD Analysis of Nonlinear and Dispersive Media

## Optical Kerr effect

polarization:  $P_K(t) = \varepsilon_0 \chi_K^{(3)} E^3(t)$   $\Rightarrow$  finite difference eq.

$\uparrow$  const. nonlinear susceptibility

## Lorentz dispersion

polarization:  $\tilde{P}_L(\omega) = \tilde{\chi}_L(\omega) \tilde{E}(\omega) = \frac{\varepsilon_0 \Delta \varepsilon_L \omega_L^2}{\omega_L^2 + 2j\delta_L \omega - \omega^2} \tilde{E}(\omega)$

$\Downarrow \mathcal{F}^{-1}$

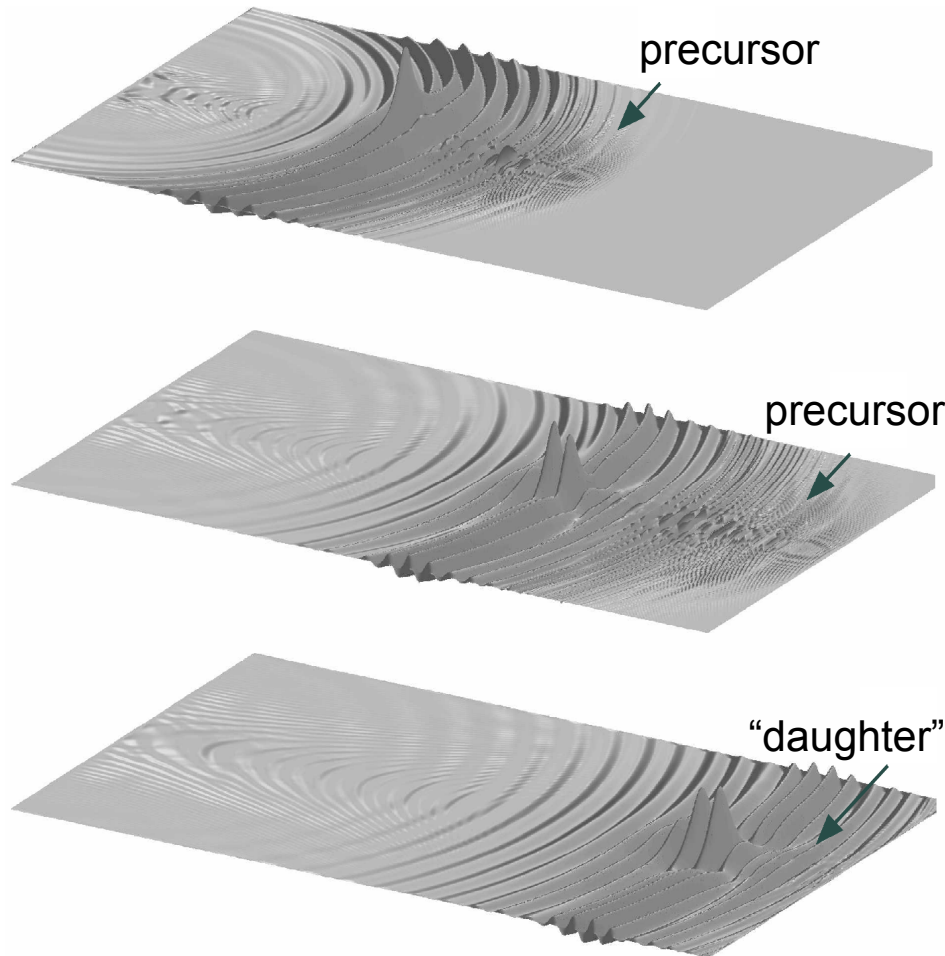
$$\omega_L^2 P_L(t) + 2\delta_L \frac{dP_L(t)}{dt} + \frac{d^2 P_L(t)}{dt^2} = \varepsilon_0 \Delta \varepsilon_L \omega_L^2 E(t)$$

$\Rightarrow$  finite difference eq.

$\Rightarrow$  solved with Yee's leapfrog algorithm



# Nonlinear 2D Wavelet FDTD Method for Kerr / Raman Medium



Dual processor version:

Electric field of the 2D pulse in the Kerr and the Raman medium taken at times

$t = 270, 360$  and  $450$  fs, from top.

High-order DD4  $\phi$ -scheme with resolution  $\Delta x = \Delta z = 0.04 \mu\text{m}$ .

REQUIRED COMPUTATIONAL RESOURCES FOR THE 2D ANALYSES

Scheme	Space resolution ( $\mu\text{m}$ )	Time step (fs)	Memory (MB)	CPU time (minutes)
FDTD	0.02	0.0472	666	201
DD <sub>4</sub>	0.02	0.00943	671	1162
DD <sub>4</sub>	0.04	0.0189	193	146
DD <sub>4</sub>	0.08	0.0377	60	20



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# Example for Silicon On Insulator (SOI) Nanophotonic Device

## Challenges for numerical modelling:

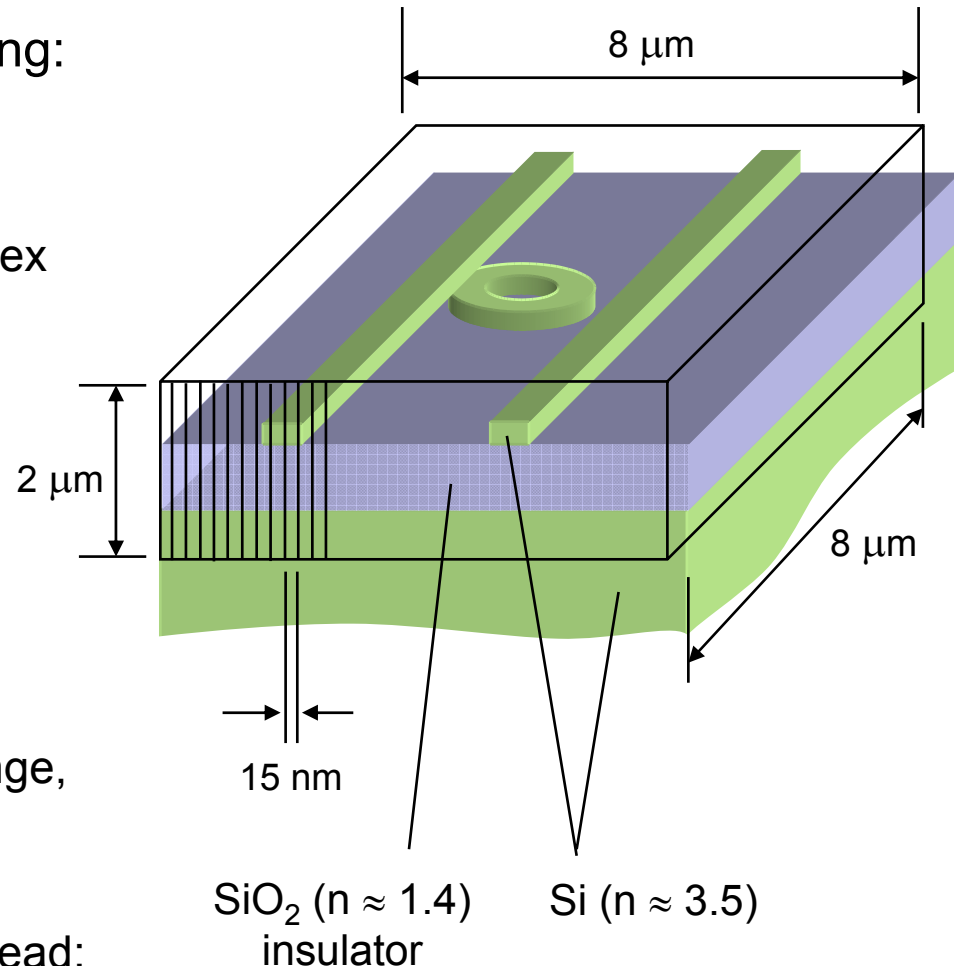
- High-contrast

1. Large changes in refractive index prevent approximations  
⇒ vectorial optics needed

- Numerically “large”

2. Typical problem size  
 $8 \mu\text{m} \times 8 \mu\text{m} \times 2 \mu\text{m}$   
( $16 \lambda \times 16 \lambda \times 4 \lambda$ )
3. Grid size in 15 nm ( $\sim \lambda / 30$ ) range, essential for reliable results  
⇒ Memory requirements for standard FDTD > 2 GB. Instead:

**Wavelet FDTD with memory savings**



# Parallel Computing Efficiency with 3D DD4-Scaling FDTD Code

IBM RS/6000 SP-SMP, distributed memory  
256 CPU, 1 GB RAM each  
375 MHz clock rate

Computation region:  
 $8.3 \times 7.1 \times 2.3 \mu\text{m}^3$ ,  
various gridsizes

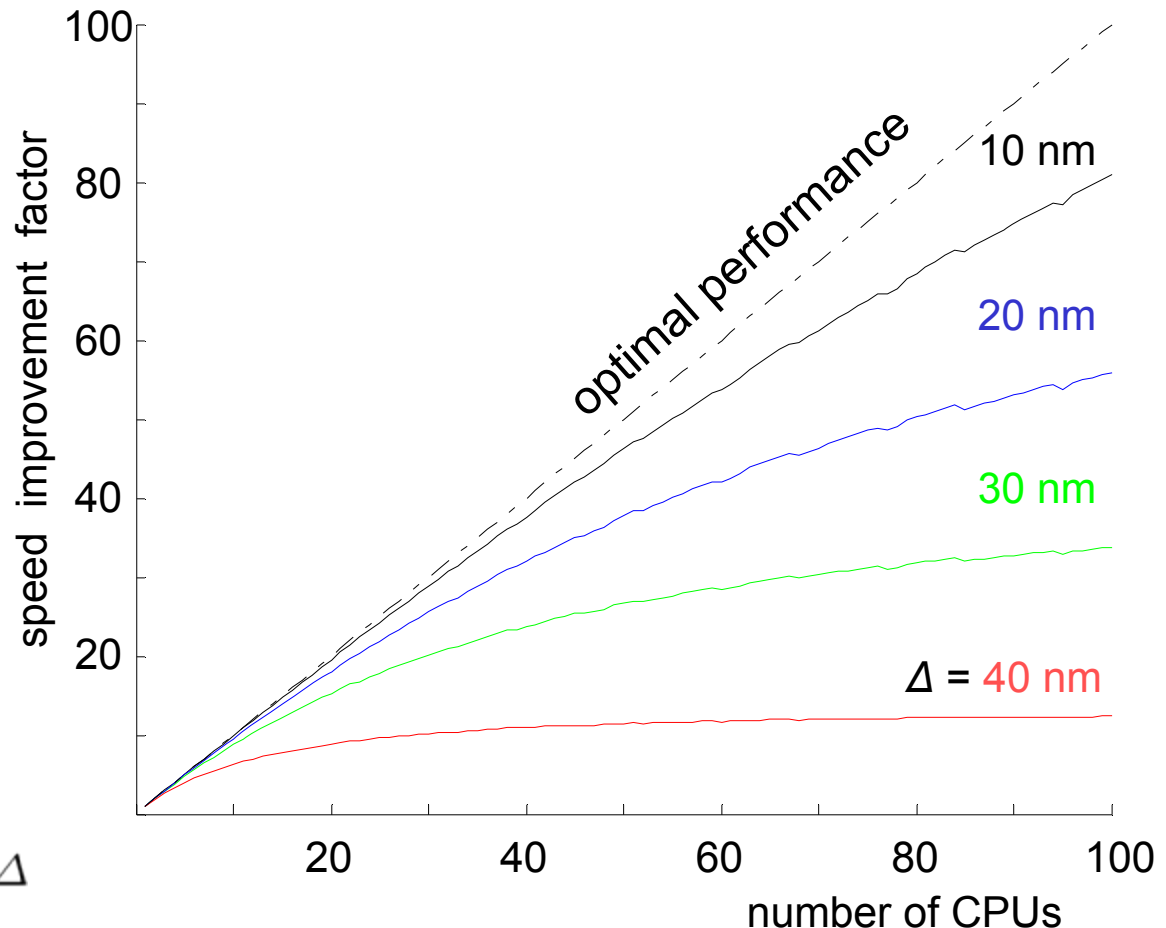
Timespan:  
250 fs

Computation times:

1 **PC** CPU: 110 h  
1 // CPU: 450 h ( $\times 4$ )  
100 // CPU: 5.5 h ( $\div 20$ )

DD4 scaling function:

$S = 0.2; \Delta x = \Delta y = \Delta z = \Delta$

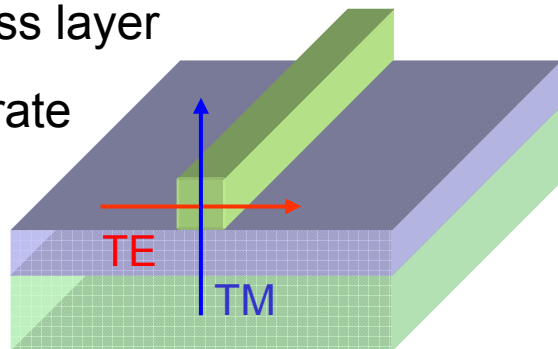


# Silicon Nanostrip Waveguide with Sidewall Roughness

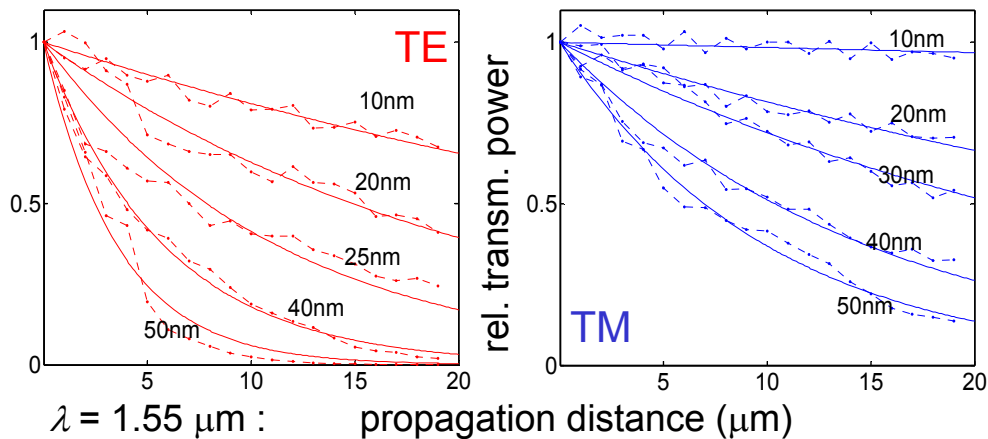
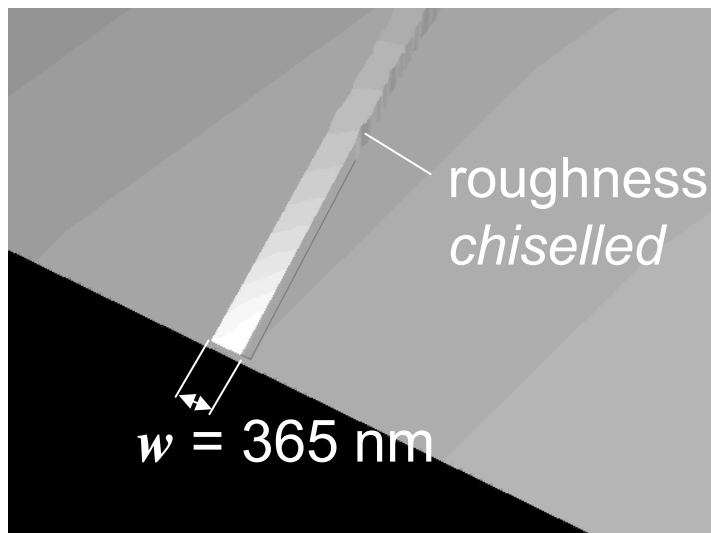
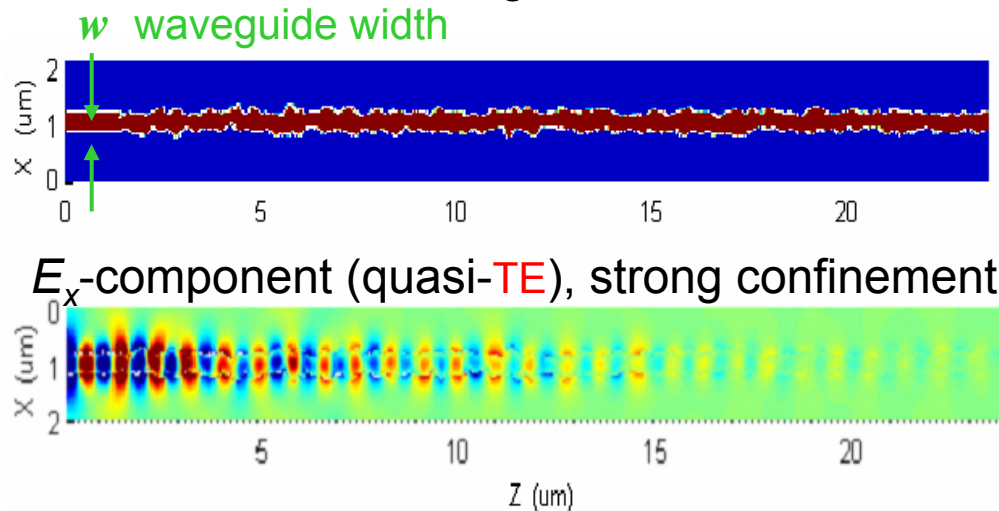
Straight  $0.365 \times 0.365 \mu\text{m}^2$  Si WG, side walls with RMS roughness  $\sigma = 10 \dots 50 \text{ nm}$

thick glass layer

Si substrate



DD4 scaling fct, parallel cluster:  
 $S=0.2$ ;  $\Delta x, \Delta y, \Delta z = 10, 20, 50 \text{ nm}$



Poulton, C. G.; Koos, C.; Fujii, M.; Pfrang, A.; Schimmel, Th.; Leuthold, J.; Freude, W.: Radiation modes and roughness loss in high index-contrast waveguides. IEEE J. Sel. Topics Quantum Electron. 12 (2006) 1306–1321



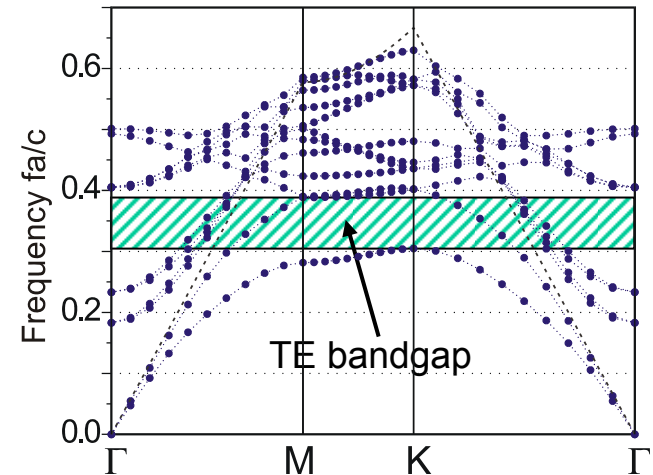
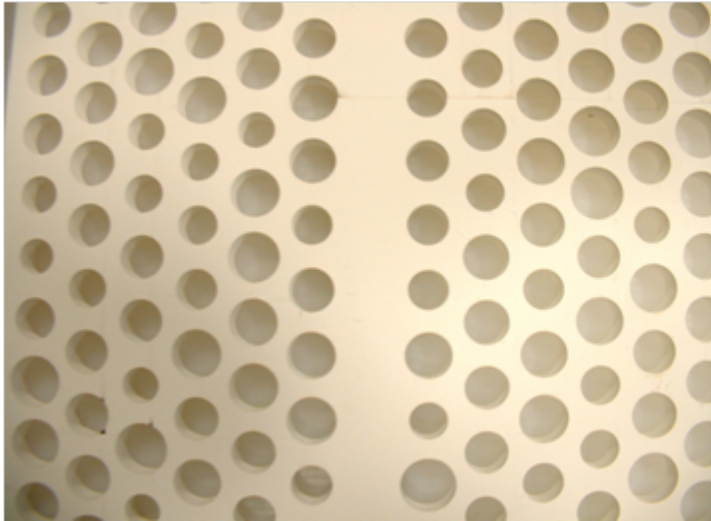
# Outline

- Wavelets
  - What are they good for?
- Finite-differences in time-domain
  - Yee's leapfrog algorithm
  - Numerical dispersion, stability and accuracy
  - Higher-order finite-differences
  - Method of weighted residuals: Collocation
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# PC W1-WG with Radius Disorder

10% radius disorder



Band diagram for PC  
without waveguide

Plane wave method,  
RSoft Bandsolve

Model for Silicon PC with substrate

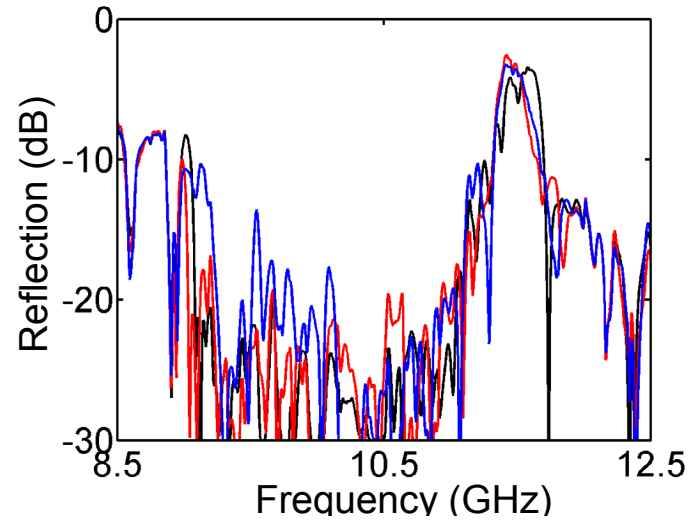
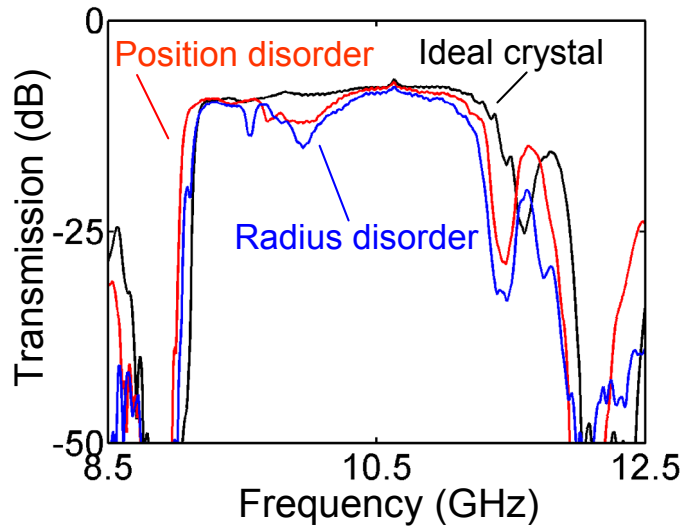
$$a = 0.52 \mu\text{m}, r/a = 0.35, h/a = 0.52$$

$$n = 3.16, n_{\text{Sub}} = 1.53$$

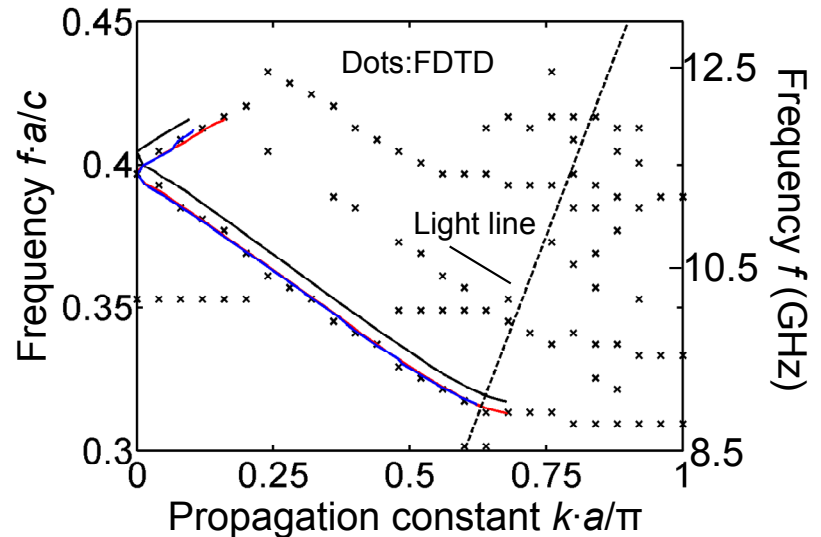




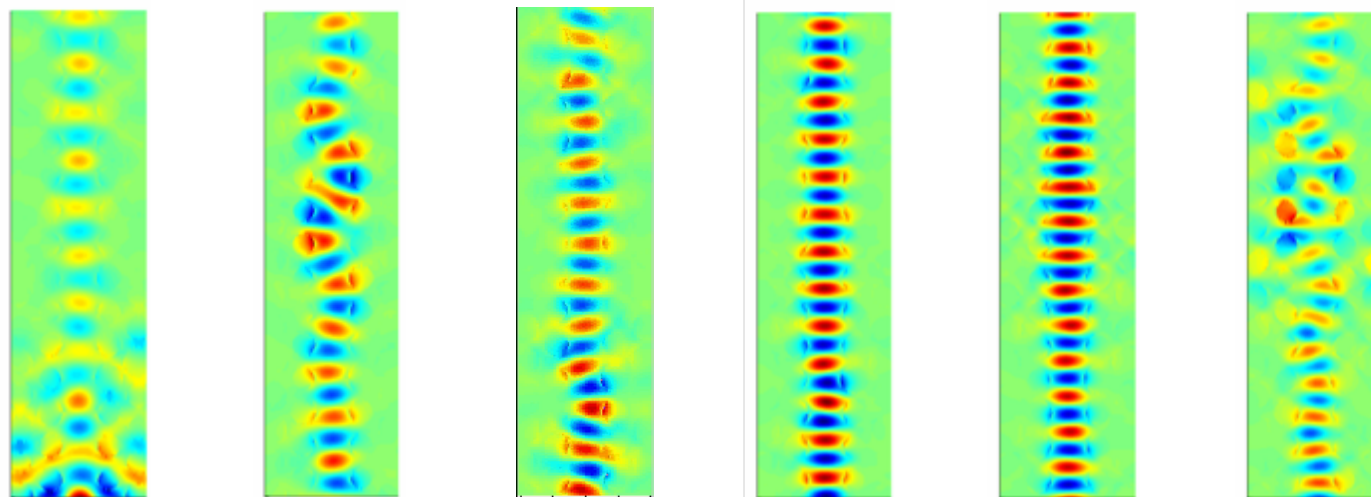
# PC W1-WG With Radius and Position Disorder



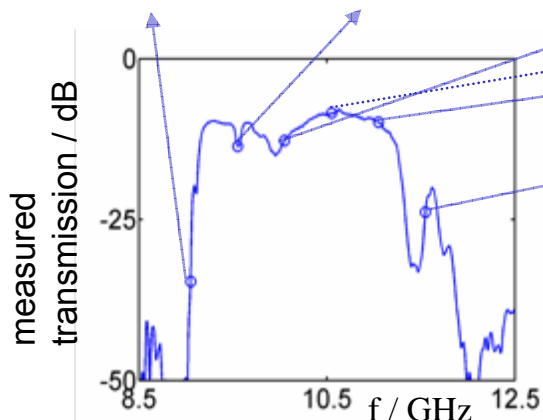
Measurements in upscaled  
microwave model



# TE ( $E_x$ ) Field of W1 Defect Waveguide with 10 % Radius Disorder



9.05 GHz      9.55 GHz      10.05 GHz      10.55 GHz      11.05 GHz      11.55 GHz



10% radius disorder

## Parallel DD4 FDTD:

48 CPU, 3 000 min  
 $\Delta x, y, z = 2 \text{ nm}$  (uniform)  
 $\rightarrow 23.7 \text{ Mio}$  mesh cells!

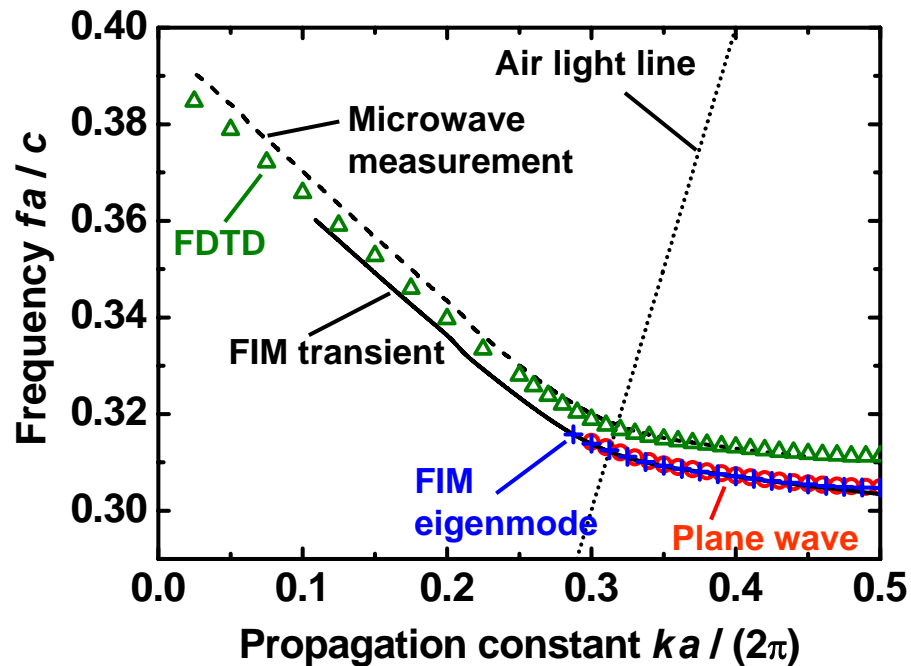
Duration: 92 pulse widths

Brosi, J.-M.; Leuthold, J.; Freude, W.: Microwave-frequency experiments validate optical simulation tools and demonstrate novel dispersion-tailored photonic crystal waveguides. IEEE J. Lightw. Technol. (2007) (submitted)

Brosi, J.-M.; Freude, W.; Leuthold, J.; Petrov, A. Yu.; Eich, M.: 'Broadband slow light in a photonic crystal line defect waveguide', Technical Digest OSA Topical Meeting on Slow and Fast Light (SL'06), Washington (DC), USA, 23–26 July 2006, Paper MD6



# Slow-Light PC Waveguide: Band diagram



Model for Silicon membrane

$$a = 0.45 \mu\text{m}, n = 3.16$$

$$r/a = 0.25, h/a = 0.6$$

**FDTD:** RSoft Fullwave, unit cell with phase cond., ~90 min/point

**FIM Eigenmode:** CST Microwave Studio, unit cell with phase cond., ~1,5 min/point

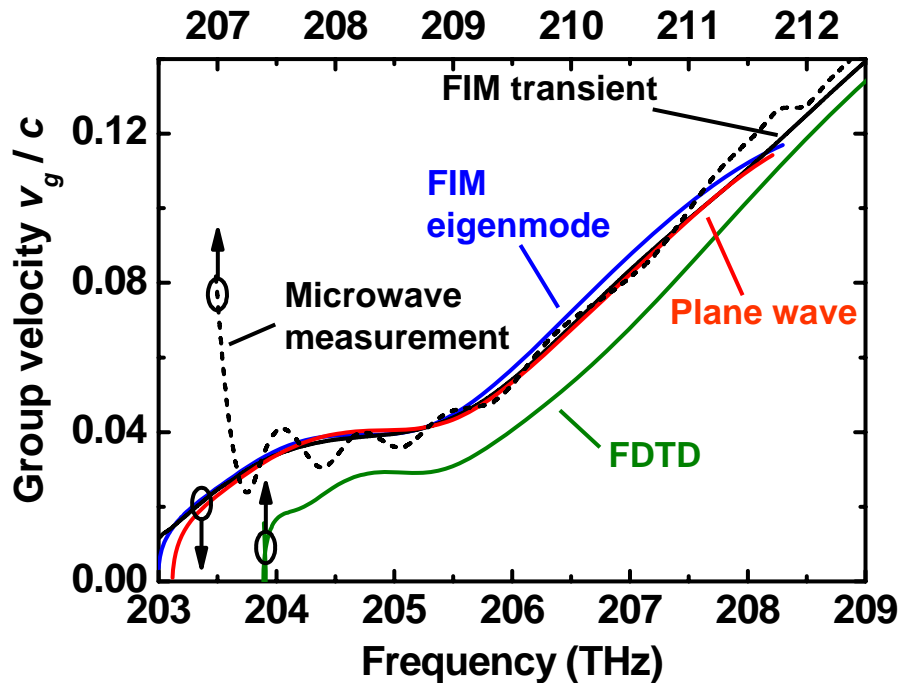
**FIM Transient:** CST Microwave Studio, full structure with pulse exc., ~5 hrs

**Plane Wave:** MIT MPB, periodically repeated in all directions, ~18 min/point

**Microwave Measurement:** Upscaled microwave model



# Slow-Light PC Waveguide: Group Velocity



Curves derived from band diagram data

**FIM eigenmode, FIM transient and plane wave method agree very well**

**Results verified with microwave experiments  
(Frequency offset because of material tolerances)**



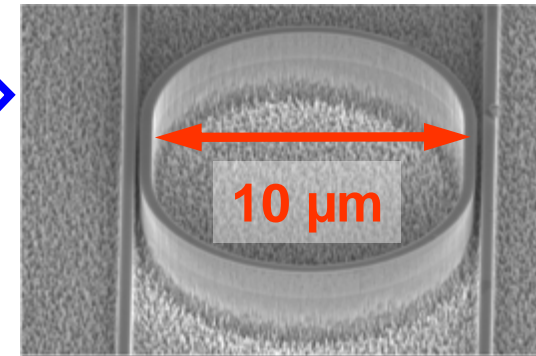
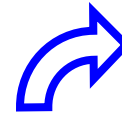
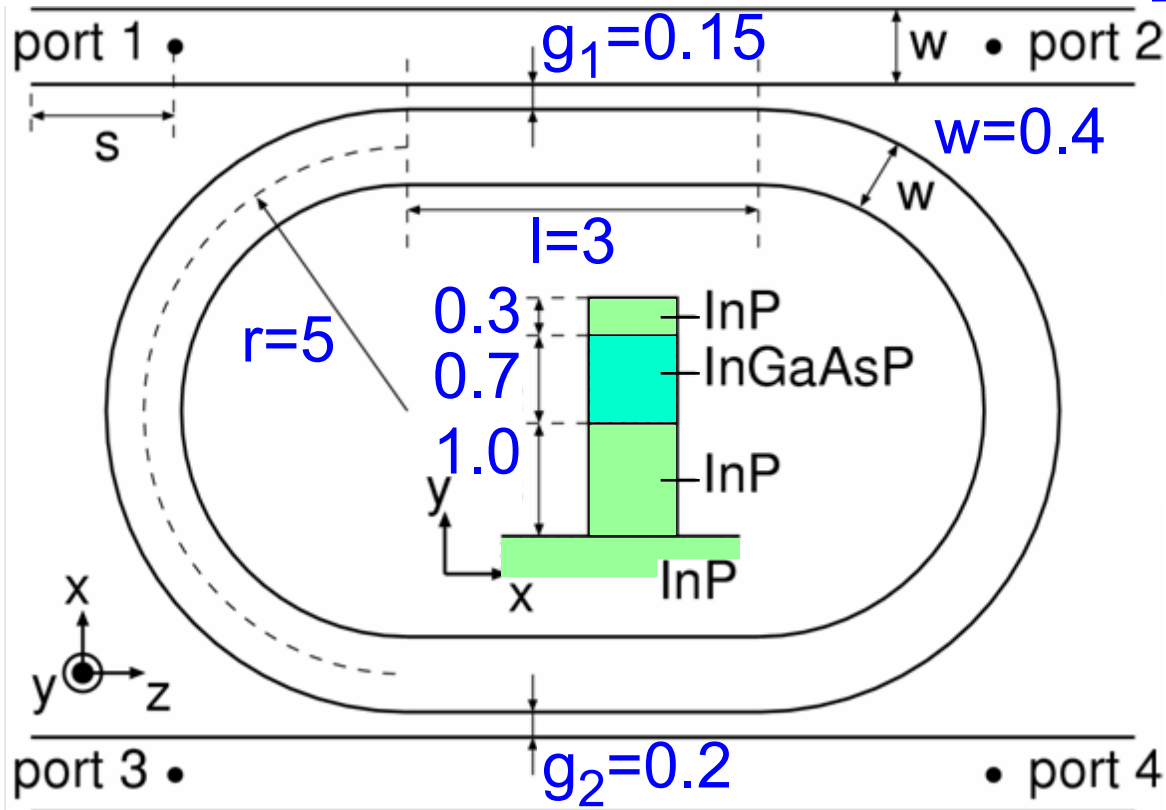
# Outline

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# Linear and Nonlinear InP/InGaAsP Micro-Resonator

## Resonator structure (2D and 3D)



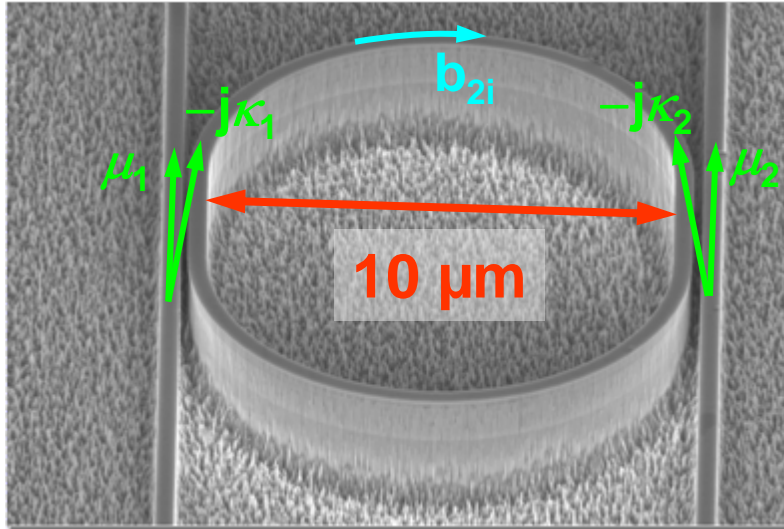
- substrate: InP
- InGaAsP:  $n = 3.42$   
 $\chi^{(3)} = 3.8 \times 10^{18} \text{ m}^2/\text{V}^2$
- InP:  $n = 3.17$   
at  $1.55 \mu\text{m}$
- for 2D analysis:  
 $n = 3.34$  (slab WG)

(all units:  $\mu\text{m}$ )



$b_2$

# InP/InGaAsP Micro-Resonator and FWM



$a_1$

$b_3$

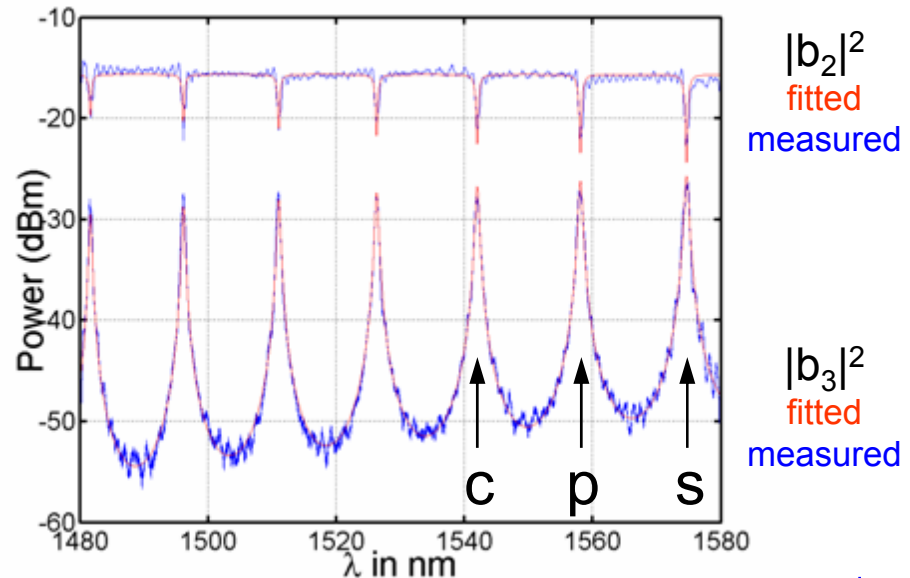
Evaluation of the coupling parameters from transmission fit:

$$\begin{aligned} \kappa_1 &= 0.24 & \rho &= 0.92 \\ \kappa_2 &= 0.13 & FE &= 2.16 \end{aligned}$$

Conversion efficiency improved by resonant field enhancement:

$$\eta_{\text{FWM}} = FE_p^4 FE_s^2 FE_c^2 \gamma_{\text{re}}^2 \left| L_{\text{rt, eff}} \right|^2 P_p^2$$

$$FE = \frac{b_{2i}}{a_1} = \frac{\kappa_1}{1 - \rho\mu_1\mu_2}$$



# FWM in Ring Resonators

FWM conversion efficiency is strongly improved by resonant field enhancement:

$$\eta_{\text{FWM}} = FE_p^4 FE_s^2 FE_c^2 \gamma_{\text{re}}^2 \left| L_{\text{rt, eff}} \right|^2 P_p^2$$

$$FE = \frac{b_{2i}}{a_1} = \frac{\kappa_1}{1 - \rho\mu_1\mu_2}$$

Field enhancement factor

$$\gamma_{\text{re}} \sim \text{Re} \left\{ \iint \left( \chi^{(3)} : \vec{E}\vec{E}\vec{E} \right) \cdot \vec{E}^* \, dx \, dy \right\}$$

Nonlinear Parameter of the ring waveguide

$$L_{\text{rt, eff}} \approx L_{\text{rt, geom}} \times \rho^2$$

Effective round-trip length for nonlinear interaction

$$L_{\text{rt, geom}}$$

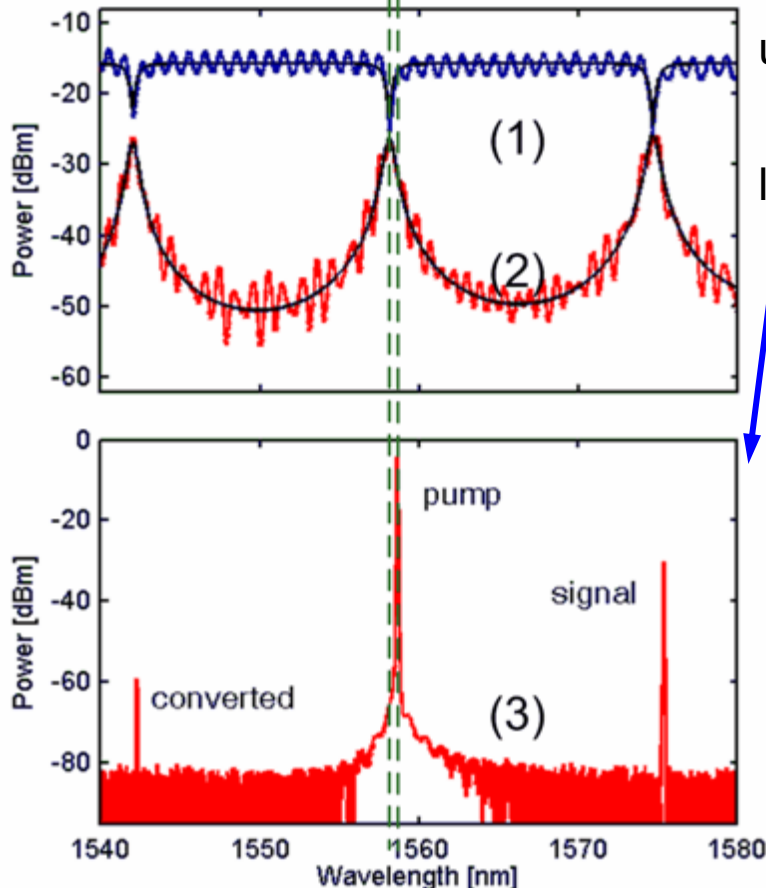
Geometrical round-trip length of the ring resonator





# FWM Experiment

Shift of resonance comb due to heating by pump power



upper part: Linear transmission characteristics

lower part: Output spectrum for nonlinear experiment

**FWM conversion efficiency:**

**$\eta = -32.5$  dB for  $P_p = 13.6$  dBm**  
(measured, pump power on chip)

To our knowledge the highest FWM-conversion efficiency measured in passive microrings

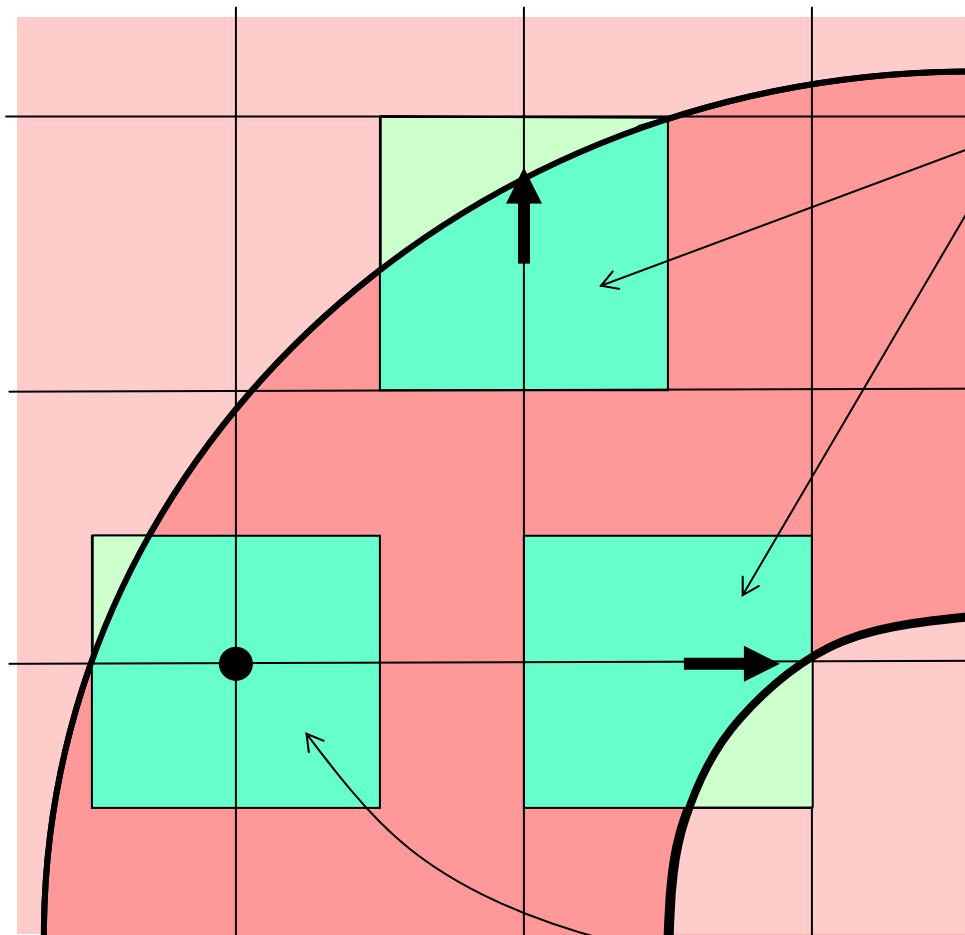
Absil et. al., Optics Letters 25, No. 8, April 2000:  
“Wavelength conversion in GaAs microring resonators”  
measured efficiency:  $-44.6$  dB  
predicted efficiency:  $-8.5$  dB



# Effective Dielectric Constant Technique Reduces Staircasing

N.Kaneda, et.al, IEEE MTT, vol.45, No.9, Sep. 1997

S.Yu, R.Mitra, IEEE MWCL, vol.11, No.1, Jan. 2001



For E fields having components normal to the interface:

$\epsilon_{\text{eff}}$ : FD method of Laplace equation

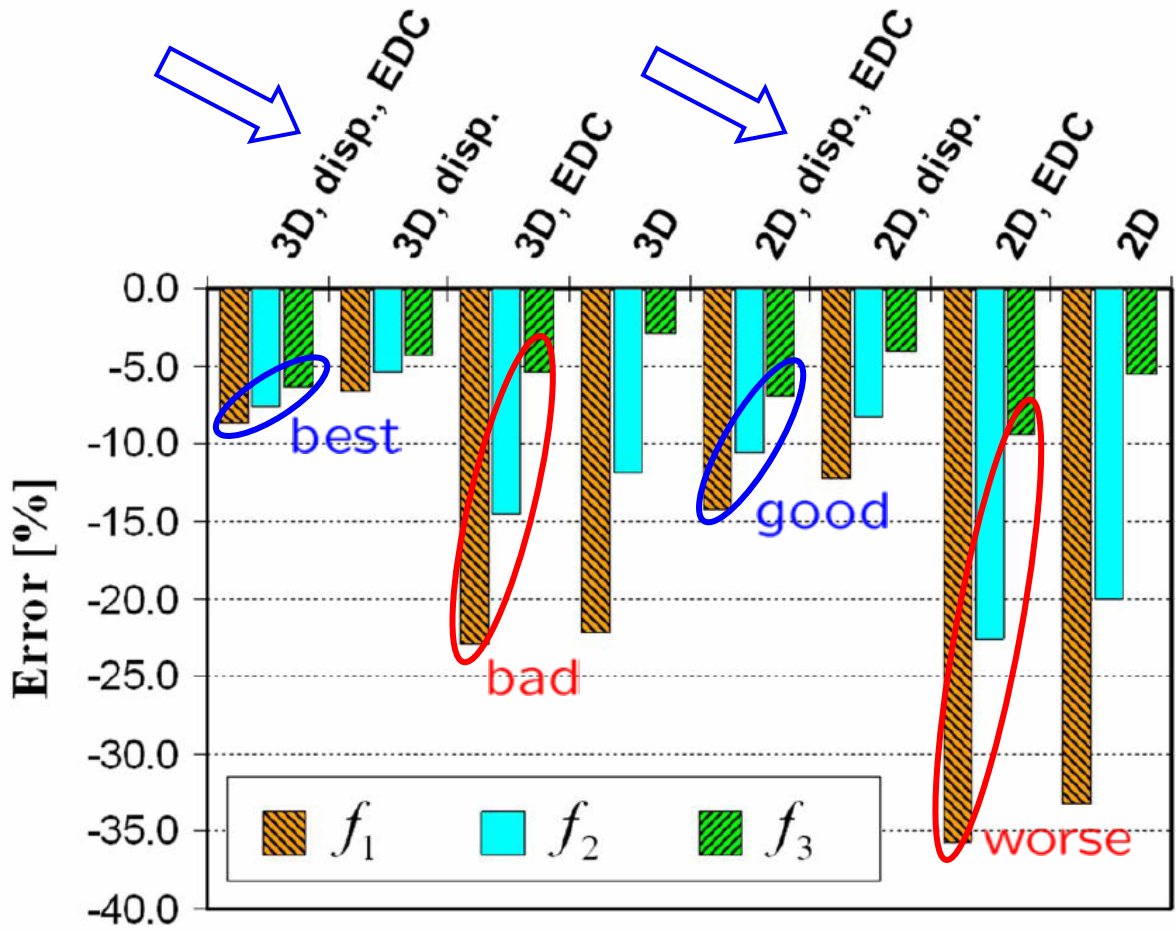
For E fields tangential to the interface:

**Tangential BC**  
(E is continuous)

$\epsilon_{\text{eff}}$ : geometric average



# Measured and Calculated Resonance Frequencies, Linear FDTD



Standard FDTD:

- 3D, 2D<sub>eff. index</sub>
- mat. dispersion
- EDC effective  $\epsilon$

Parameters

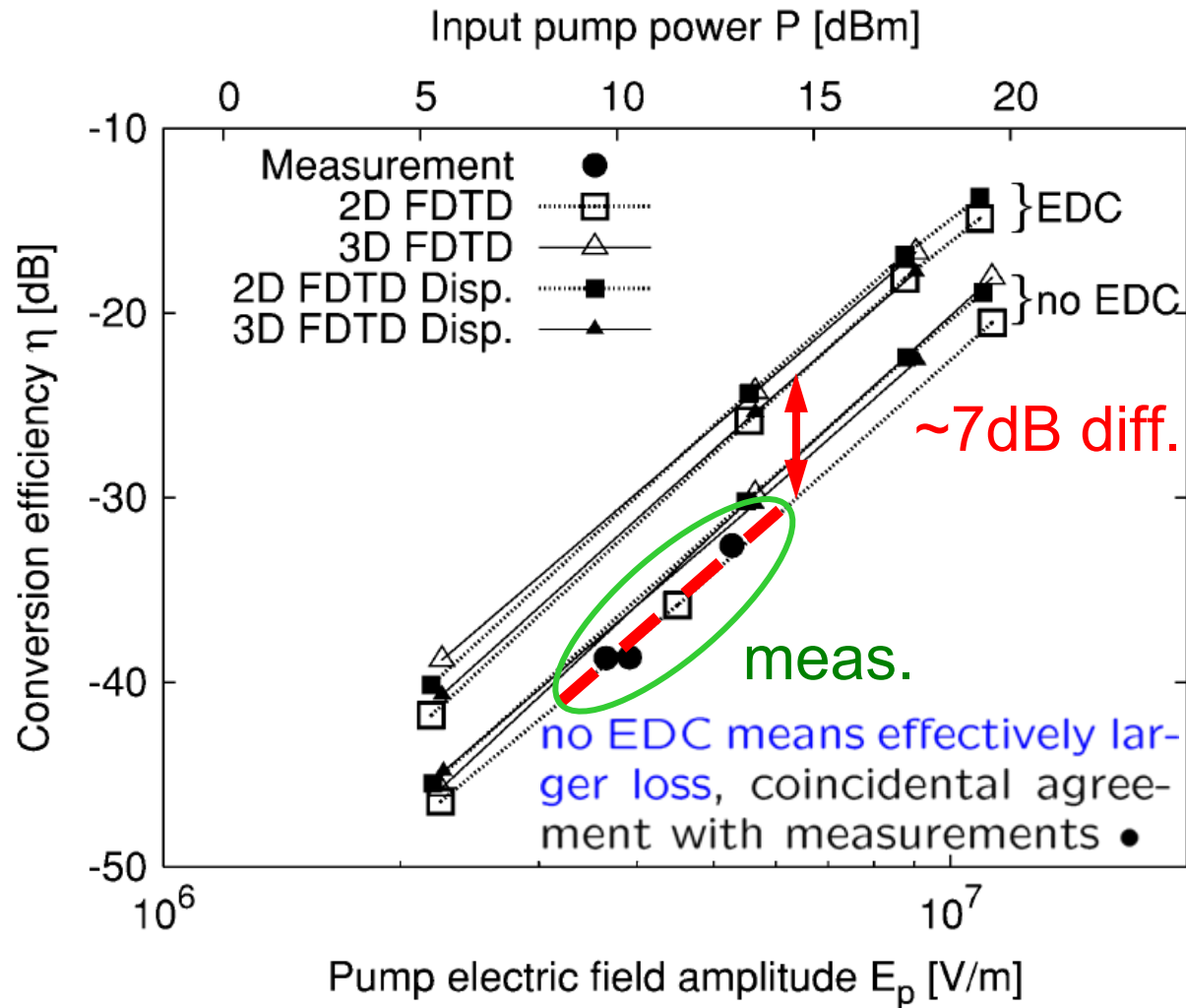
- $S = 0.8$
- $\Delta t = 0.0465$  fs
- $\Delta x = \Delta z = 25$  nm ( $\Delta y = 100$  nm)
- parallel cluster HP XC6000
- 3D/2D
- 64/32 CPU
- 4 GB / 0.5 GB,
- 28 h / 2.5 h

Comparison of measured and calculated resonance frequencies for the different FDTD implementations. The bars indicate the deviation of the resonance frequencies  $f_1$ ,  $f_2$ , and  $f_3$  in percent of the averaged measured FSR  $\Delta f^{(\text{meas})} = 2.016$  THz.

Koos, C.; Fujii, M.; Poulton, C. G.; Steingrueber, R.; Leuthold, J.; Freude, W.: FDTD-modelling... JQE 42 (2006) 1215–1223



# Measured and Calculated FWM Efficiency, Nonlinear Std FDTD



## Nonlinear std FDTD

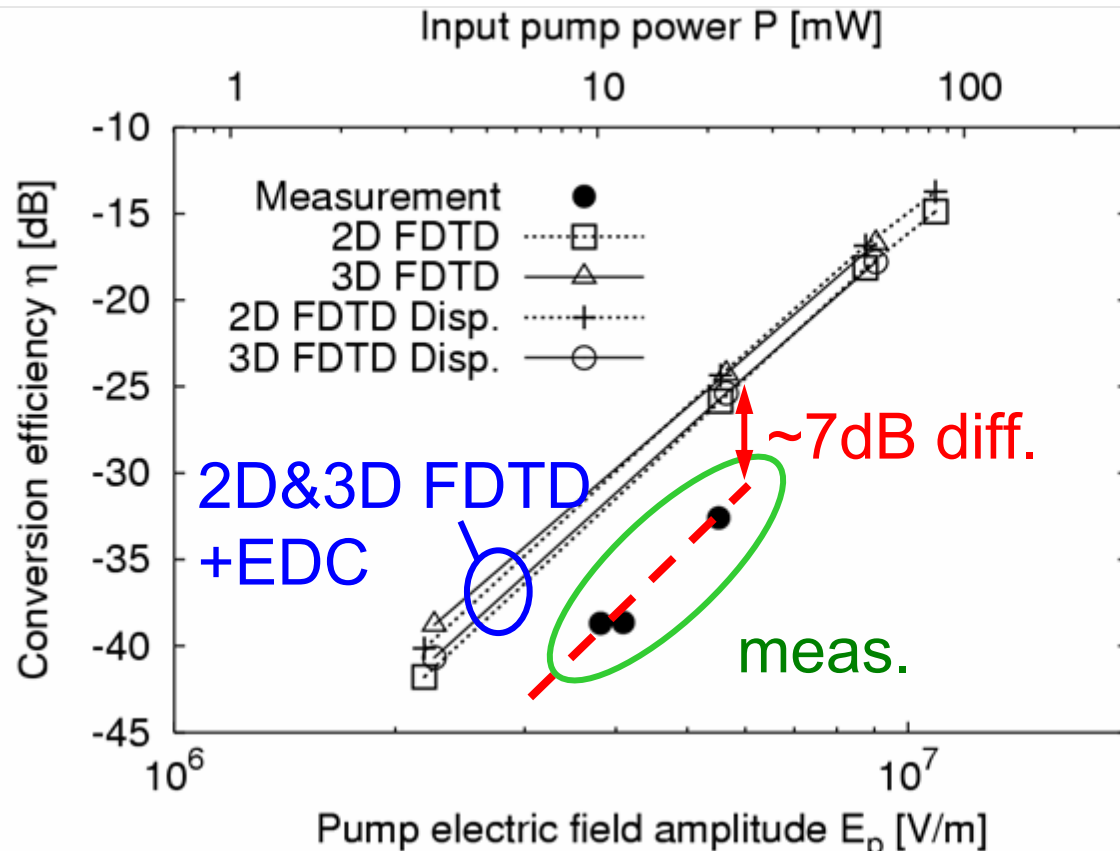
- 3D, 2D<sub>eff. index</sub>
- mat. dispersion
- EDC effective  $\epsilon$

## Parameters

- $S = 0.8$
- $\Delta t = 0.0465$  fs
- $\Delta x = \Delta z = 25$  nm  
( $\Delta y = 100$  nm)
- parallel cluster  
HP XC6000  
3D/2D  
64/32 CPU  
4 GB / 0.5 GB,  
28 h / 2.5 h



# Measured and Calculated FWM Efficiency, Nonlinear Std FDTD



## Nonlinear std FDTD

- 3D, 2D<sub>eff. index</sub>
- mat. dispersion
- EDC effective  $\epsilon$

## Parameters

- $S = 0.8$
- $\Delta t = 0.0465$  fs  
 $\Delta x = \Delta z = 25$  nm  
 $(\Delta y = 100$  nm)
- parallel cluster  
 HP XC6000  
 3D/2D  
 64/32 CPU  
 4 GB / 0.5 GB,  
 28 h / 2.5 h



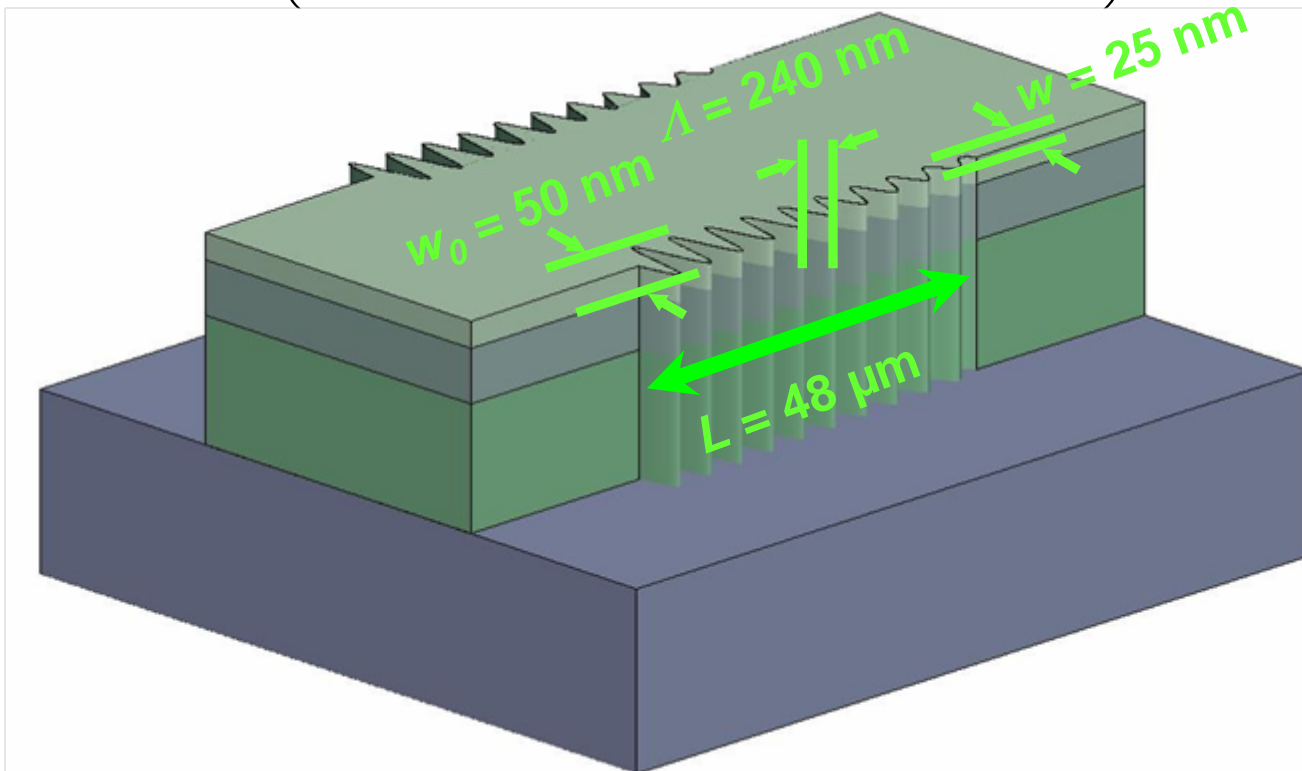
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# Simulation of Stopband-Tapered WBG Using NL 2D Std FDTD

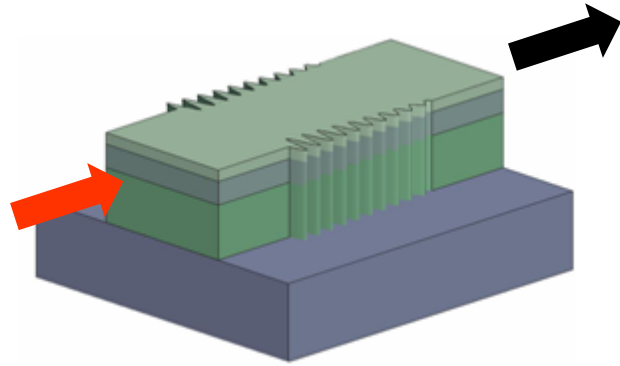
$$w = w_0 \left( -0.4856 \left( \frac{z}{L} \right)^3 + -0.0009 \left( \frac{z}{L} \right)^2 + 1 \right) \sin(2\pi z / \Lambda)$$



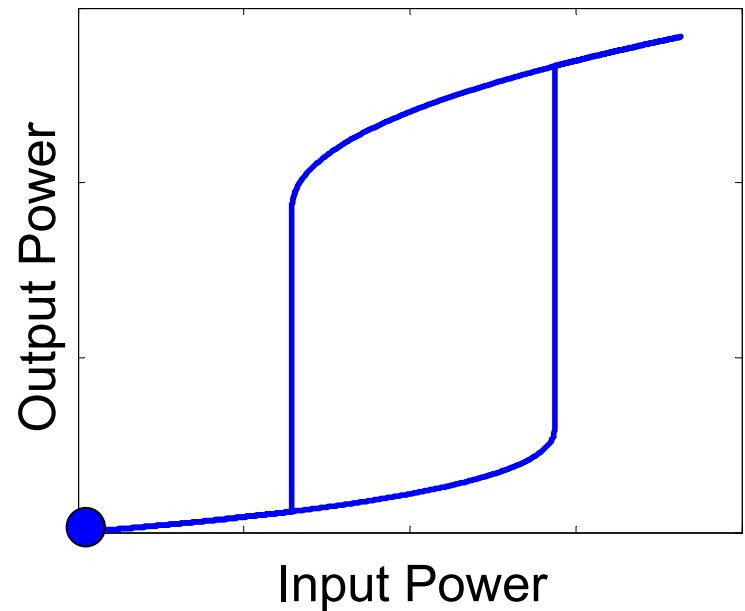
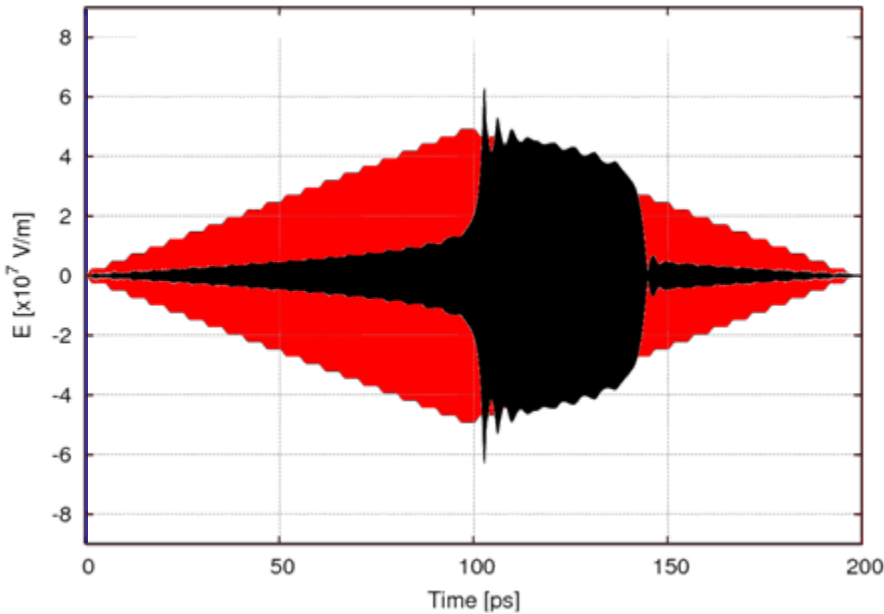
The sidewall modulation amplitude  $w$  is varied from 50 nm to 25 nm for the total grating with length  $L = 48 \mu\text{m}$  and period  $\Lambda = 0.24 \mu\text{m}$  (200 periods).



# Bistability of Stopband-Tapered WBG Using NL 2D Std FDTD

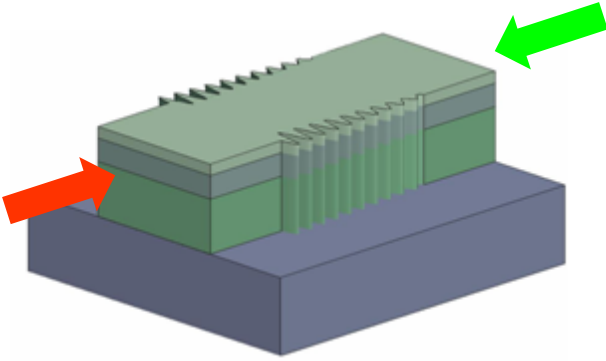


Bistable behaviour in tapered sidewall corrugated structure



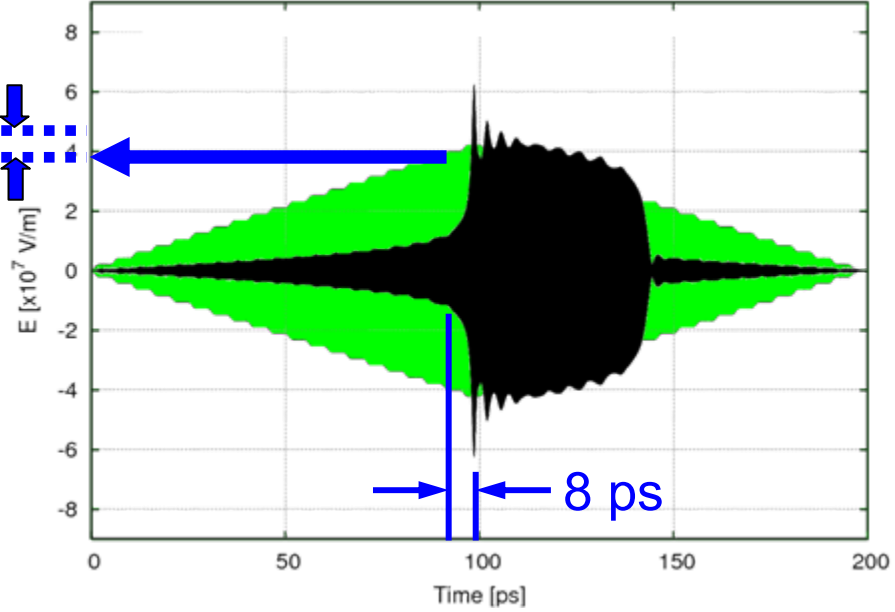
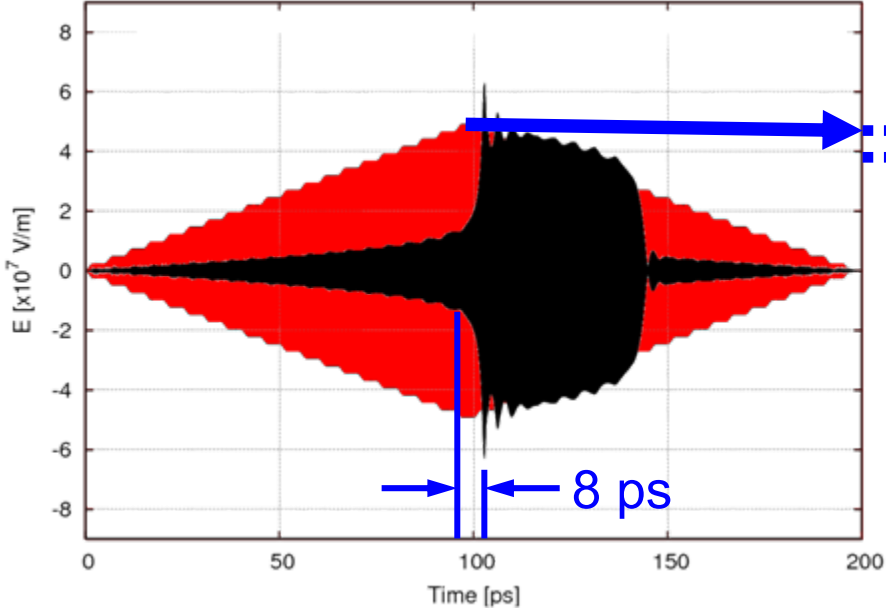


# Directionality of Stopband-Tapered WBG Using NL 2D Std FDTD



Isolator behaviour in tapered sidewall-corrugated structure

Different up-switching thresholds for **LTR** & **RTL** case



Fujii, M.; Maitra, A.; Poulton, C.; Leuthold, J.; Freude, W.: Non-reciprocal transmission and Schmitt trigger operation in strongly modulated asymmetric WBGs. Opt. Expr. 14 (2006) 12782–12793



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# Summary

## Results:

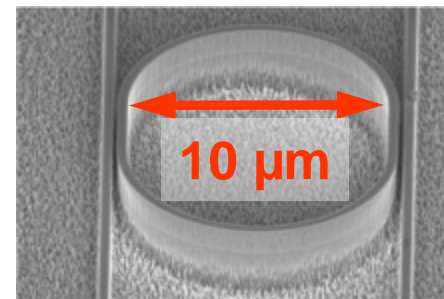
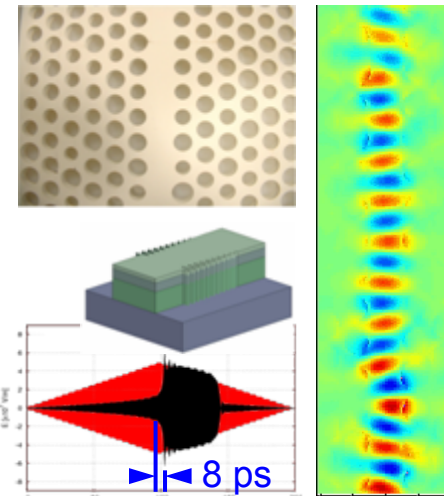
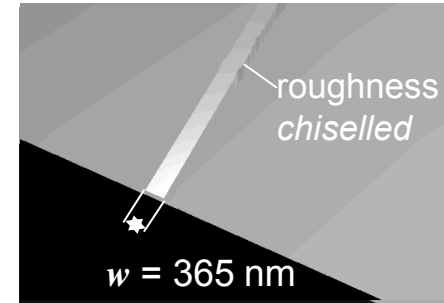
- Low dispersion for  $\phi^{Dp} / \phi^{DDp}$  (reduced memory and run time) — *might* be like HFD6
- Equivalence between  $\phi$  and  $\psi$  via wavelet transform  $\rightsquigarrow$  all dispersion findings also valid for possible MRA

## Examples (+ NL DD4 Raman/Kerr):

- Waveguide roughness (DD4 scal fct FDTD)
- Slow light ( $0.04c$ ) in PhC with disorder (DD4)
- FWM in a microring (NL FDTD)
- NL Bragg grating (NL FDTD)

## What should be done?

- Chose higher-order time difference operator (relatively simple)
- Try MRA with Daubechies  $\phi^{Dp}$  and  $\psi^{Dp}$



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- Summary and **further reading (57 citations)**



# Further Reading (1/10)

## Guided-wave modeling

- [1] Huang, W. P. (Ed.): Methods for simulation of guided-wave optoelectronic devices: Part II: Waves and interactions. In: PIER 11 — Progress in Electromagnetic Research, Chief Editor: J. A. Kong. Cambridge (MA): EMW Publishing 1995
- [2] Scarmozzino, R.; Gopinath, A.; Pregla, R.; Helfert, S.: Numerical techniques for modeling guided-wave photonic devices. IEEE J. Sel. Topics Quantum Electron. 6 (2000) 150–162

## General FDTD methods

- [3] Yee, K. S.: Numerical solution of initial boundary value problems involving Maxwells equations in isotropic media. IEEE Trans. Antennas Propag. AP-14 (1966) 302–307
- [4] Taflove, A.; Hagness, S. C.: Computational electrodynamics: The finite-difference time-domain method, 2. and 3. Ed. Boston: Artech House 2000 and 2005

## Large-scale FDTD modeling

- [5] Yu, W.; Liu, Y.; Su, T.; Hunag, N.-T.; Mittra, R.: A robust parallel conformal finite-difference time-domain processing package using the MPI library. IEEE Antennas Propag. Mag. 47 (2005) 39–59
- [6] Poulton, C. G.; Koos, C.; Fujii, M.; Pfrang, A.; Schimmel, Th.; Leuthold, J.; Freude, W.: Radiation modes and roughness loss in high index-contrast waveguides. IEEE J. Sel. Topics Quantum Electron. 12 (2006) 1306–1321



## Further Reading (2/10)

### FDTD Sources: Charging — Sinusoidal — Pulsed modal hard

- [7] Taflove, A.; Hagness, S. C.: Computational electrodynamics: The finite-difference time-domain method, 2. Ed. Boston: Artech House 2000. Sect. 5.4 p. 182 ff.; Sect. 5.10 p. 227 ff. (3. Ed. Boston: Artech House 2005)

### FDTD staircasing error mitigation

- [8] Fujii, M.; Lukashevich, D.; Sakagami, I.; Russer, P.: Convergence of FDTD and wavelet-collocation modeling of curved dielectric interface with the effective dielectric constant technique. IEEE Microw. Compon. Lett. 13 (2003) 469–471

### Replacing explicit Yee leapfrogging by alternating direction implicit scheme

- [9] Rao, H.; Scarmozzino, R.; Osgood, Jr., R. M.: An improved ADI-FDTD method and its application to photonic simulations IEEE Photon. Technol. Lett. 14 (2002) 477–479



# Further Reading (3/10)

## Stability according to the Courant-Friedrichs-Lewy condition

- [10] Lakshmikantham, V.; Trigiante, D.: Theory of difference equations: Numerical methods and applications. Boston: Academic Press 1988
- [11] Krumpholz, M.; Katehi, L. P. B.: MRTD: new time-domain schemes based on multiresolution analysis. IEEE Trans. Microwave Theory Tech. 44 (1996) 555–571
- [12] Tentzeris, E. M.; Robertson, R. L.; Harvey, J. F.; Katehi, L. P. B.: Stability and dispersion analysis of Battle-Lemarié-based MRTD schemes. IEEE Trans. Microwave Theory Tech. 47 (1999) 1004–1013
- [13] Taflove, A.; Hagness, S. C.: Computational electrodynamics: The finite-difference time-domain method, 2. Ed. Boston: Artech House 2000. Stability of fourth-order accurate differences: Sect. 4.9.2 pp. 143–147
- [14] Fujii, M.; Hoefer, W. J. R.: Interpolating wavelet collocation method of time dependent Maxwell's equations: characterization of electrically large optical waveguide discontinuities. J. Comp. Phys. 186 (2003) 666–689



# Further Reading (4/10)

## General wavelet theory

- [15] Gabor, D.: Theory of communication. J. Inst. Elec. Eng. 93 (1946) 429–457 (*cited after Ref. [187] in [7]*)
- [16] Deslauriers, G.; Dubuc, S.: Symmetric iterative interpolation processes. Constr. Approx. 5 (1989) 49–68
- [17] Goedecker, S.: Wavelets and their application for the solution of partial differential equations in physics. Lausanne: Presses Polytechniques et Universitaires Romandes 1998
- [18] Sarkar, T. K.; Su, C.: A tutorial on wavelets from an electrical engineering perspective, Part 1: Discrete wavelet techniques. IEEE Antennas Prop. Mag. 40 (1998) No. 5 49–70
- [19] Sarkar, T. K.; Su, C.; Adve, R.; Salazar-Palma, M.; Garcia-Castilloz, L.; Boix, R. R.: A tutorial on wavelets from an electrical engineering perspective, Part 2: The continuous case. IEEE Antennas Prop. Mag. 40 (1998) No. 6 36–49
- [20] Louis, A. K.; Maaß, P.; Rieder, A.: Wavelets. Theorie und Anwendungen, 2. Ed. Stuttgart: Teubner 1998 (in German)
- [21] Mallat, S.: A wavelet tour of signal processing, 2. Ed. San Diego: Academic Press 1999
- [22] Misiti, M.; Misiti, Y.; Oppenheim, G.; Poggi, J.-M.: Wavelet toolbox user's guide, v. 3.1 (MATLAB release 2006b). Natick (MA): The Mathworks Inc. 2006  
[http://www.mathworks.com/access/helpdesk/help/pdf\\_doc/wavelet/wavelet\\_ug.pdf](http://www.mathworks.com/access/helpdesk/help/pdf_doc/wavelet/wavelet_ug.pdf)





# Further Reading (5/10)

## Wavelets for solving partial differential equations

- [23] Lakshmikantham, V.; Trigiante, D.: Theory of difference equations: Numerical methods and applications. Boston: Academic Press 1988
- [24] Bacry, E.; Mallat, S.; Papanicolaou, G.: A wavelet based spacetime adaptive numerical method for partial differential equations. Math. Model. Numer. Anal. 26 (1992)
- [25] Beylkin, G.: On wavelet-based algorithms for solving differential equations. In 'Wavelets: Mathematics and Applications', edited by J. J. Benedetto and M. W. Frazier. Boca Raton: CRC Press 1994. Pages 449–466
- [26] Harten, A.: Adaptive multiresolution schemes for shock computations. J. Comput. Phys. 115 (1994) 319–338
- [27] Goedecker, S.: Wavelets and their application for the solution of partial differential equations in physics. Lausanne: Presses Polytechniques et Universitaires Romandes 1998



# Further Reading (6/10)

## Wavelet FDTD leads to low dispersion and coarser grids

- [28] Fujii, M.; Hofer, W. J. R.: Application of wavelet-galerkin method to electrically large optical waveguide problems. IEEE MTT-S Int. Microwave Symposium Digest, vol. TU3E-3 (2000) 239–242
- [29] Fujii, M.; Hofer, W. J. R.: Dispersion of time domain wavelet Galerkin method based on Daubechies compactly supported scaling functions with three and four vanishing moments. IEEE Microwave Guided Wave Lett., 10 (2000) 125–127
- [30] Fujii, M.; Hofer, W. J. R.: Time-domain wavelet Galerkin modeling of two-dimensional electrically large dielectric waveguides. IEEE Trans. Microwave Theory Tech. 49 (2001) 886–892

## Wavelet FDTD staircasing error mitigation

- [31] Fujii, M.; Lukashevich, D.; Sakagami, I.; Russer, P.: Convergence of FDTD and wavelet-collocation modeling of curved dielectric interface with the effective dielectric constant technique. IEEE Microw. Compon. Lett. 13 (2003) 469–471

## Multiscale analysis

- [32] Marrone, M.; Mittra, R.: A theoretical study of the stability criteria for hybridized FDTD algorithms for multiscale analysis. IEEE Trans. Antennas Propagat. 52 (2004) 2158–2166



## Further Reading (7/10)

### Linear wavelet FDTD

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## Boundary conditions

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# Further Reading (10/10)

## Periodic boundary conditions

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# End of Heraeus Presentation

Additional material: Periodic boundary conditions



# Periodic Boundary Conditions (PBC)

PBC equivalently describe an infinite structure, comprising the elementary domain (“unit cell” size  $D_x, D_y, D_z$ ), which is translated and thereby infinitely repeated in all directions with a so-called lattice vector  $\vec{R} = n_x D_x \vec{e}_x + n_y D_y \vec{e}_y + n_z D_z \vec{e}_z$  (integers  $n_{x,y,z}$ ).

Bloch’s theorem (Floquet theorem in microwaves) states that in a periodic medium the eigenmodes have the property:

$$\begin{aligned}\vec{E}_{\vec{k}}(\vec{r}) &= \vec{u}_{\vec{k}}(\vec{r}) e^{-j\vec{k}\cdot\vec{r}}, & \vec{H}_{\vec{k}}(\vec{r}) &= \vec{v}_{\vec{k}}(\vec{r}) e^{-j\vec{k}\cdot\vec{r}} \\ \vec{u}_{\vec{k}}(\vec{r} + \vec{R}) &= \vec{u}_{\vec{k}}(\vec{r}), & \vec{v}_{\vec{k}}(\vec{r} + \vec{R}) &= \vec{v}_{\vec{k}}(\vec{r})\end{aligned}$$

**Consequence** (assume  $n_{x,y,z} = 1$  for primitive vector  $\vec{R}$ ):

$$\Rightarrow \vec{E}_{\vec{k}}(\vec{r} + \vec{R}) = \vec{E}_{\vec{k}}(\vec{r}) e^{-j\vec{k}\cdot\vec{R}}, \quad \Rightarrow \vec{H}_{\vec{k}}(\vec{r} + \vec{R}) = \vec{H}_{\vec{k}}(\vec{r}) e^{-j\vec{k}\cdot\vec{R}}$$

Only if  $k_x D_x, k_y D_y, k_z D_z$  happen to equal  $2\pi$ , then the phasors  $\vec{E}_{\vec{k}}(\vec{r} + \vec{R}), \vec{H}_{\vec{k}}(\vec{r} + \vec{R})$  do *not* experience a phase shift.

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# Periodic Boundary Conditions — Analytic Signal

Analytic signal  $\underline{\Psi}(t, \vec{r}) = \Psi(t, \vec{r}) + j\Psi_i(t, \vec{r})$  with real part  $\Psi(t, \vec{r})$  and imaginary part  $\Psi_i(t, \vec{r})$ . Has causal spectrum  $\underline{\Psi}(f < 0) = 0$ .

Hilbert transform connects real and imaginary parts (spatial dependency dropped, principal value integral  $\mathcal{P}f$ ):

$$\Psi_i(t) = -\frac{1}{\pi} \mathcal{P} \int_{-\infty}^{+\infty} \frac{\Psi(t')}{t' - t} dt', \quad \Psi(t) = \frac{1}{\pi} \mathcal{P} \int_{-\infty}^{+\infty} \frac{\Psi_i(t')}{t' - t} dt'$$

Plane wave, real  $A$  and  $\vec{k} = k_x \vec{e}_x + k_y \vec{e}_y + k_z \vec{e}_z$ ,  $|\vec{k}|^2 = (n\omega_0/c)^2$ :  
 $\underline{\Psi}(t, \vec{r}) = A \exp[j(\omega_0 t - \vec{k} \cdot \vec{r})] = \underbrace{A \cos(\omega_0 t - \vec{k} \cdot \vec{r})}_{\Psi(t, \vec{r})} + j \underbrace{A \sin(\omega_0 t - \vec{k} \cdot \vec{r})}_{\Psi_i(t, \vec{r})}$

$\underline{\Psi}(t, \vec{r}) = A(t, \vec{r}) \exp[j(\omega_0 t - \vec{k} \cdot \vec{r})]$  represents actual real signal  $\Psi(t, \vec{r})$ . Average frequency  $f_0 = \omega_0/(2\pi)$ .

“Slowly” varying complex amplitude  $A(t, \vec{r}) \rightarrow \tilde{\Psi}(f, \vec{r})$  concentrated at  $f = \pm f_0$ , no spectral overlap at  $f = 0$ .



# Periodic Boundary Conditions — Formulation

Analytic signal  $\underline{\Psi}(t, \vec{r}) = A(t, \vec{r}) \exp[j(\omega_0 t - \vec{k} \cdot \vec{r})]$ . Real part  $\Psi(t, \vec{r}) = E_q(t, x, y, z), H_q(t, x, y, z)$  ( $q = x, y, z$ ) stands for any field component.

PBC and propagation in  $z$ -direction only for simplification:

$$\underline{\Psi}(t, z + D_z) = \underline{\Psi}(t, z) \exp(-j k_z D_z)$$

Wrap-around update:

$$\underline{\Psi}(t, z) = \underline{\Psi}(t, z + D_z) \exp(j k_z D_z)$$

Ingenious idea by Ko and Mittra (1993): Propagate real signal  $\Psi(t, \vec{r})$  and imaginary signal  $\Psi_i(t, z)$  with parallel FDTD runs. Excitation of unit cell by  $A(t, z) = |A(t, z)| \exp[j \varphi(t, z)]$  at  $z = z_0$ :

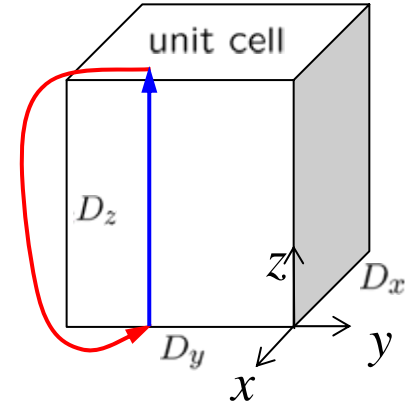
$$\Psi(t, z_0) = |A(t, z_0)| \cos[\omega_0 t - k_z z_0 + \varphi(t, z_0)]$$

$$\Psi_i(t, z_0) = |A(t, z_0)| \sin[\omega_0 t - k_z z_0 + \varphi(t, z_0)]$$



# Periodic Boundary Conditions — Procedure

- Excite unit cell at  $z = 0$  with  $\Psi(t, z), \Psi_i(t, z)$  of modulated plane wave (spectrum near  $f = f_0$ ).
- Propagate initial fields according to FDTD update equations until boundary  $z = D_z + \Delta z$ .
- Calculate update fields at  $z = 0$  by:



$$\Psi(t, 0) = \Re \{ [\Psi(t, D_z + \Delta z) + j\Psi_i(t, D_z + \Delta z)] \exp(j k_z D_z) \},$$

$$\Psi_i(t, 0) = \Im \{ [\Psi(t, D_z + \Delta z) + j\Psi_i(t, D_z + \Delta z)] \exp(j k_z D_z) \}$$

- Repeat updating and wrapping until stationary state is reached.
- Fourier transform stationary time sequence and analyze for resonances (eigenfrequencies)  $f_\ell$  belonging to chosen  $k_z$ .
- The spatial distributions associated with the various  $f_\ell$  are the eigenfields or modes of the periodic arrangement.

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